

2012

WURTH ELECTRONICS

EMC Seminar

Introduction to Concepts and
Techniques TORONTO

Agenda

Magnetic Materials / Inductors

Common Mode Filters

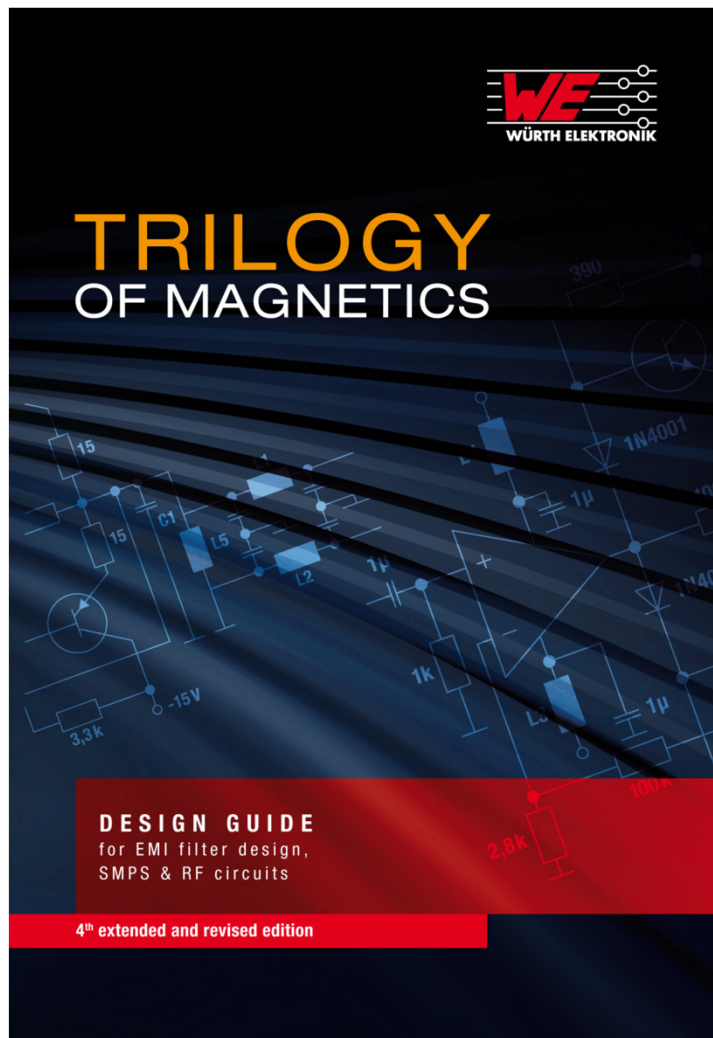
Chip Bead Ferrites

Shields

TRILOGY OF MAGNETICS



Supplement Material to Today's Seminar

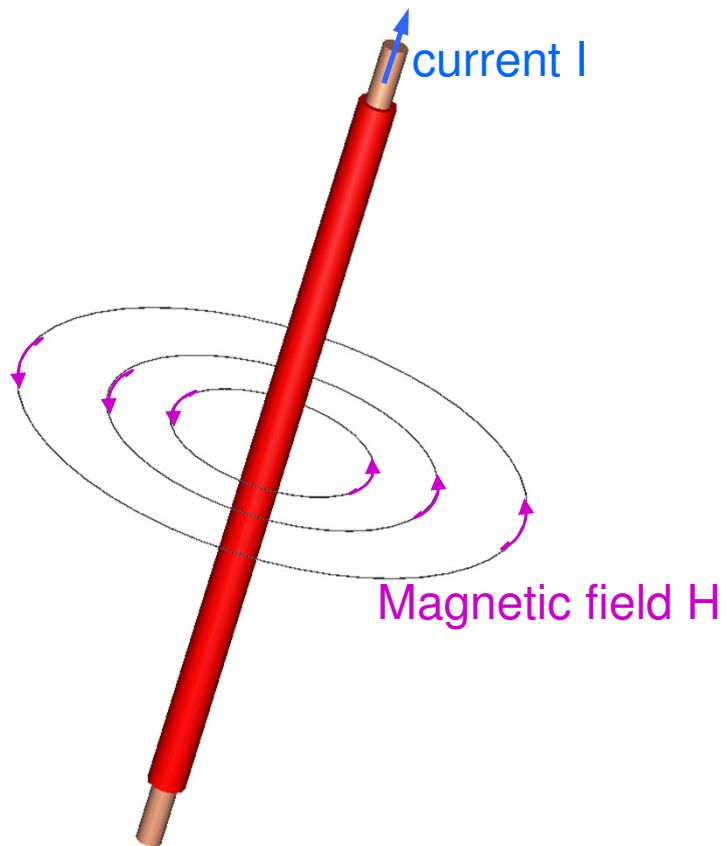


- 1. Basic Principles of Magnetics
 - 2. Magnetics Components
 - 3. Filter Circuits
 - 4. Applications
- concrete examples on more than 300 pages

MAGNETIC MATERIALS AND INDUCTORS

The Magnetic Field (H)

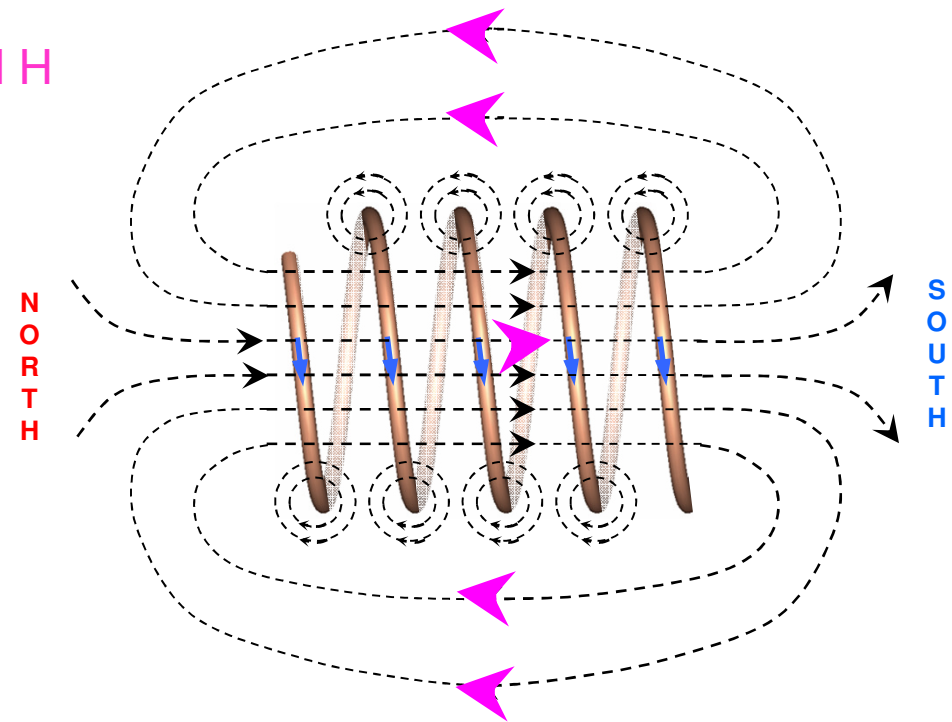
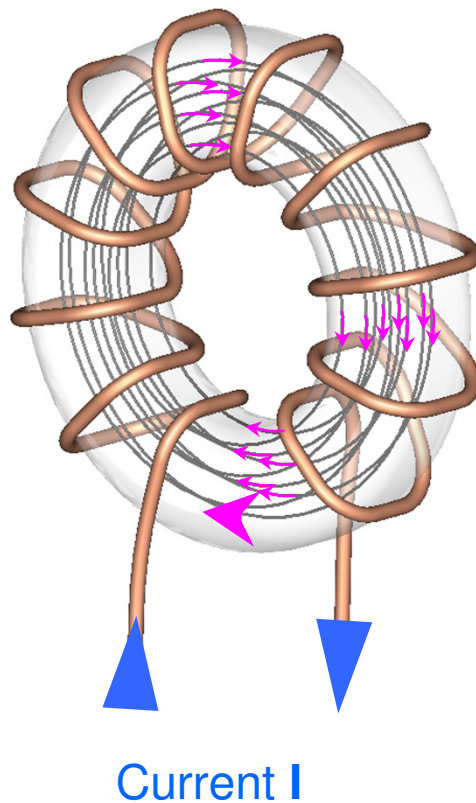
When current flows through a wire it generates a magnetic field



Field model



The Inductor and its Magnetic Field



Many EMC solutions consist of **inductive** components

Materials Used to Make Inductors



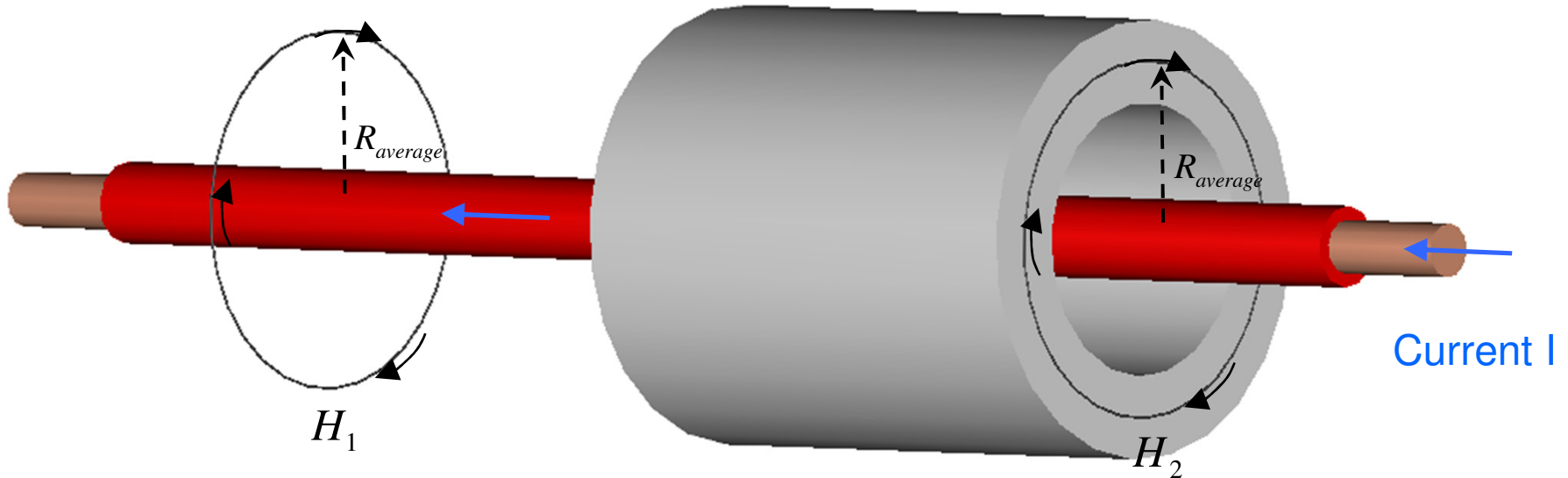
Iron Alloys

Fe + Other
materials

Ferrites

Ni
+ Zn
Mn

Magnetic H and B Fields



$$H_1 = H_2 = H = \frac{I}{2 \cdot \pi \cdot R_{average}}$$

$$B_1 \begin{matrix} \neq \\ ? \\ = \end{matrix} B_2$$

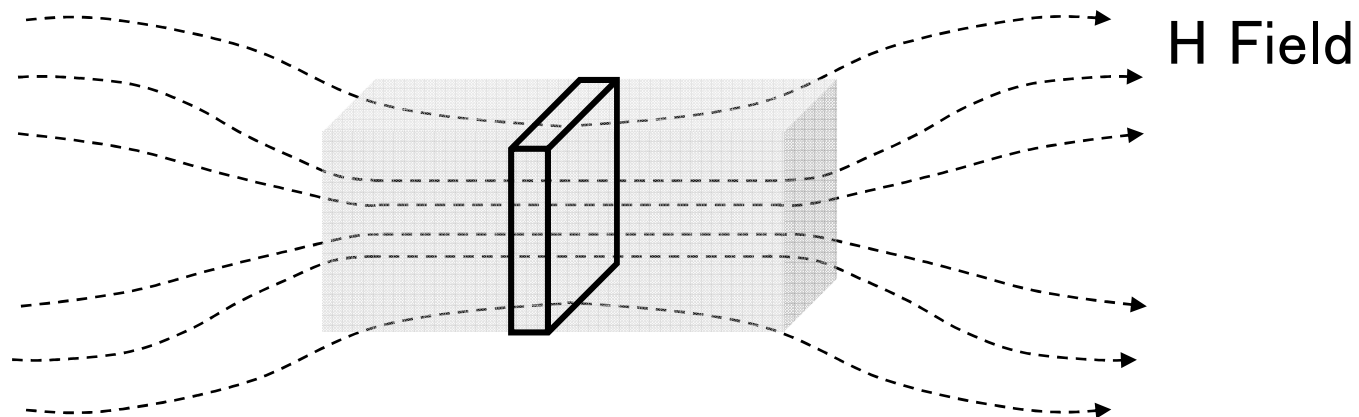
$$B = \mu_o H$$

Permeability



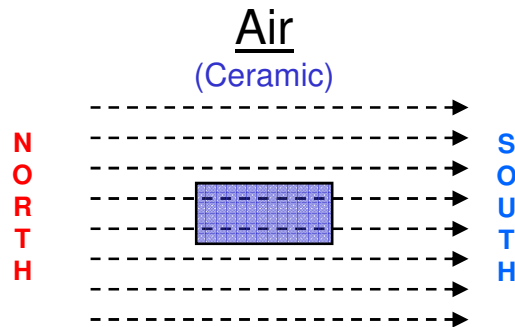
Permeability

Describes the capacity of concentration of the magnetic flux in the material
It is a factor of the energy needed to magnetize a material (i.e. magnetically align dipoles)

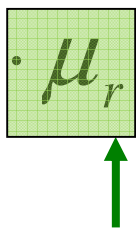


$$B = \mu_0 H$$

Permeability

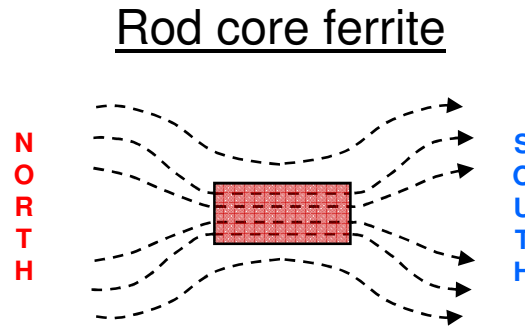


Induction in air:


$$B = \mu_0 \cdot \mu_r \cdot H$$


Linear function because $\mu_r = 1$

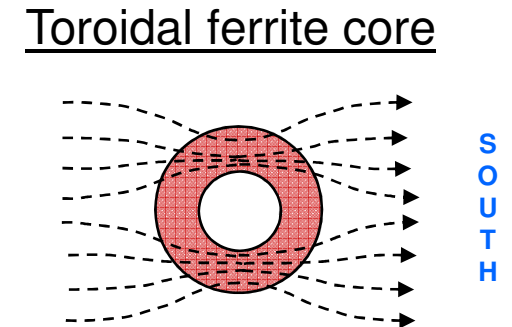
$$B = \mu_0 \cdot H$$



Induction in Ferrite:

$$B = \mu_0 \cdot \mu_r \cdot H$$


Relative permeability is non-linear



Material

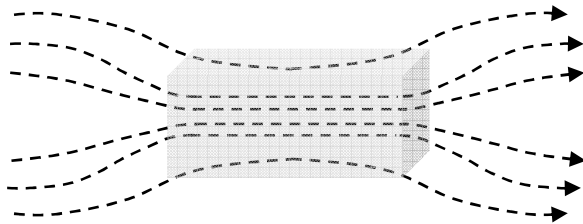
Frequency

Temperature

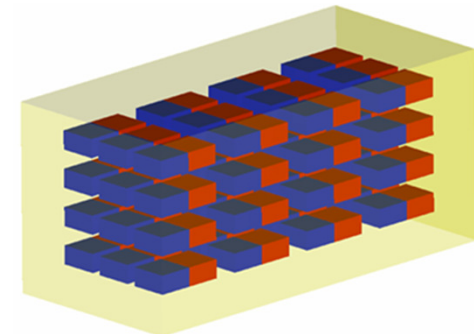
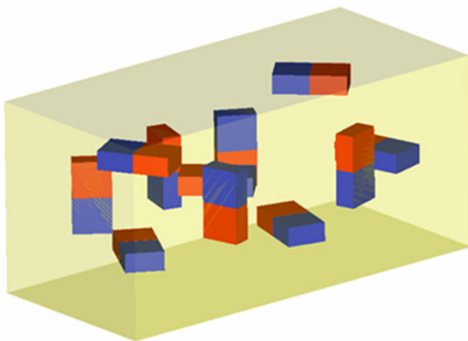
Current

Pressure

Magnetic Materials



H Field



Non-magnetic

Soft-magnetic

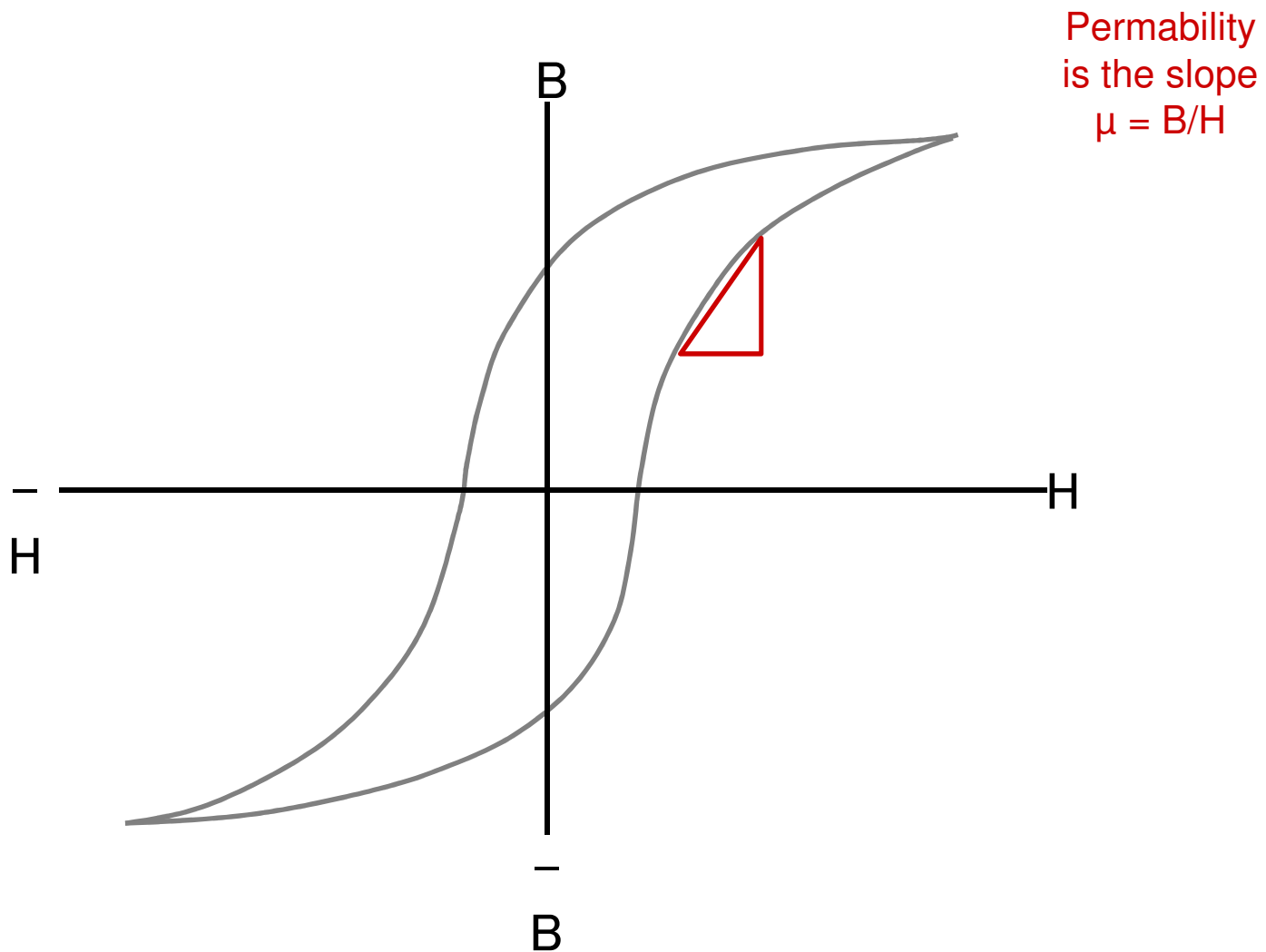
Hard-magnetic

Insulators

Ferrites

Fe Alloys

Hysteresis Graph



Soft Ferrites (Have low losses at high frequencies)



- NiZn and MnZn ferrites can be easily magnetized and demagnetized **without dissipating much energy** (hysteresis loss)
- Hysteresis loss is a component of the total core loss.
- Another component of total core loss are Eddy currents which leak useful magnetizing energy
- MnZn has a higher permeability and saturation induction than NiZn
- NiZn has higher resistivity than MnZn. (NiZn works well above 1 MHz)



Hard Ferrites (Have high losses in general)

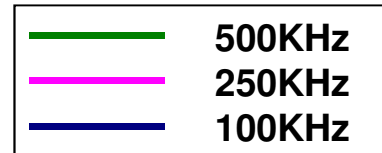


- Hard ferrites cannot be easily magnetized and demagnetized. They need a substantial amount of energy change the direction of the magnetization (high hysteresis loss)
- Iron-based ferrites have the general formula $MO-Fe_2O_3$ where M is a divalent ion such as Fe, Ni, Cu, Mg, Mn, Co, Zn or Li.
- They are great conductors of magnetic flux well and have a high magnetic permeability
- They store (handle) stronger magnetic field than soft ferrites
- Relatively, they have low resistivity

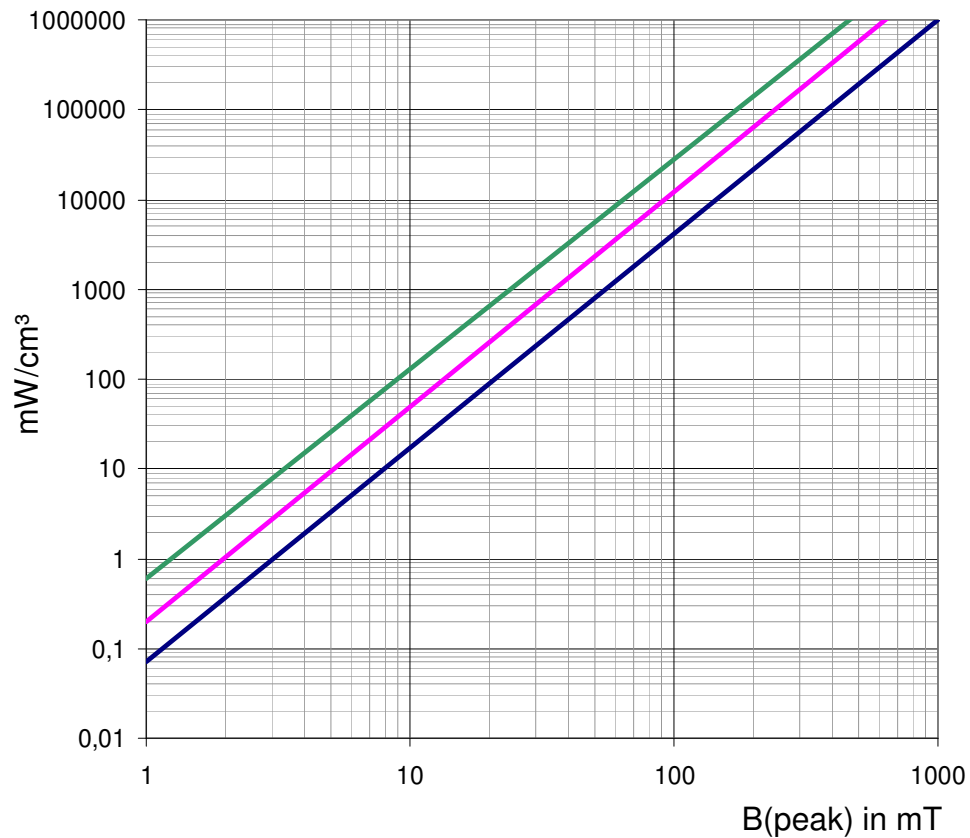


Core losses

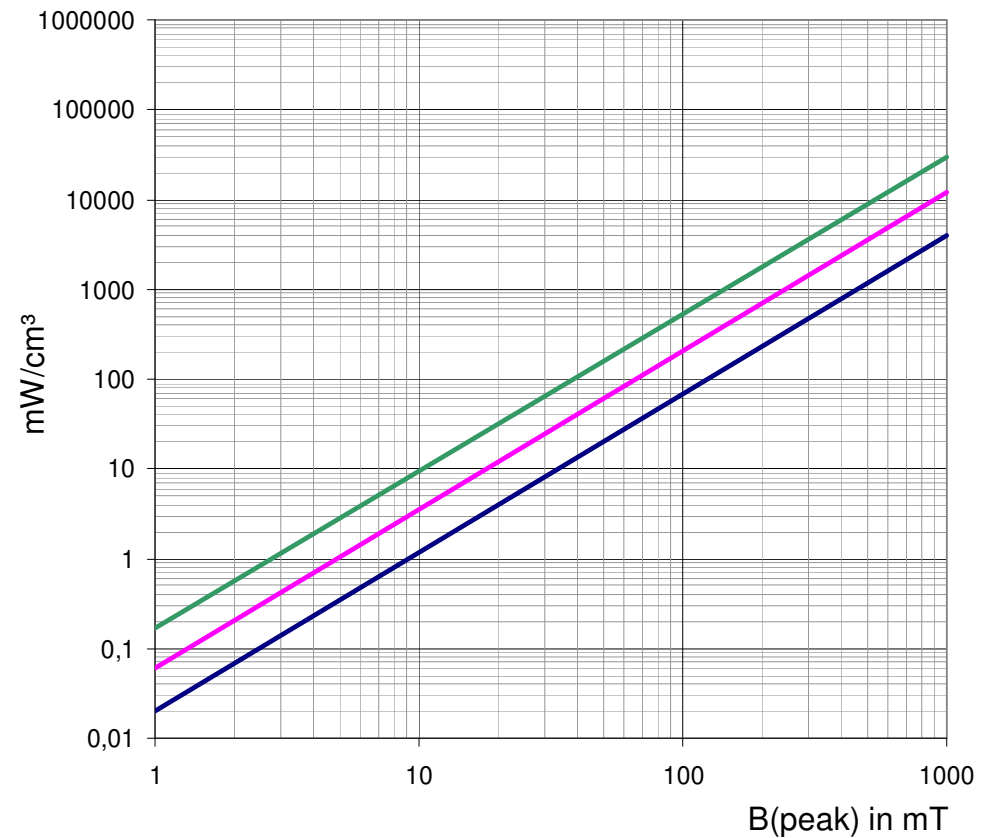
Comparing materials



Iron powder



NiZn

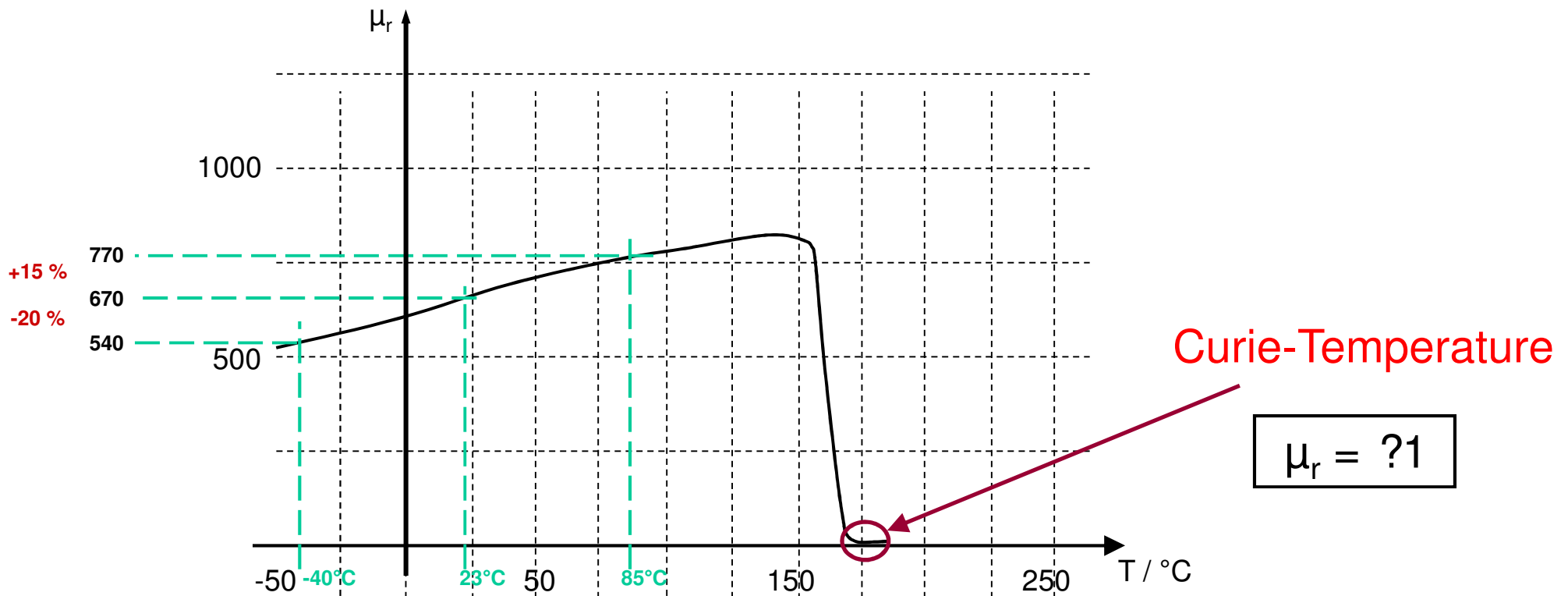


CORE GAP, INDUCTOR IMPEDANCE, TEMPERATURE PERFORMANCE AND SHIELDING

Permeability's dependence on temperature

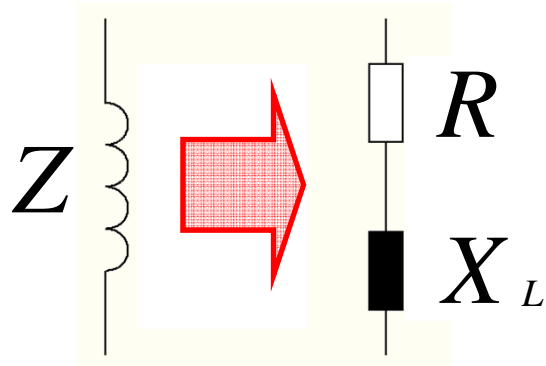
$T \uparrow \rightsquigarrow$ thermal motion $\uparrow \rightsquigarrow$ degree of order \downarrow

\rightsquigarrow Alignment of elementary magnets 



Complex Permeability

Equivalent Circuit diagram



$$Z = \sqrt{R^2 + X_L^2}$$

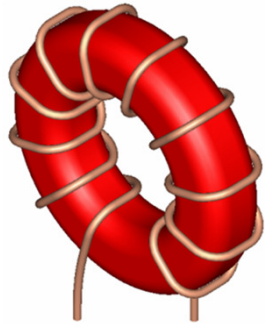
XL is called the reactance (a measure of storing ability)

R is called the resistance (a measure of dissipating ability)

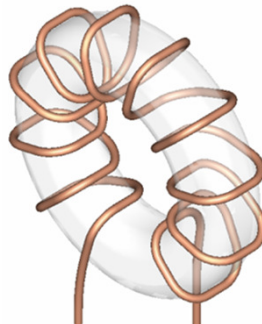
XL is purely inductive and hence stores and releases energy without any loss

R is purely lossy and hence it dissipates energy

Permeability – complex permeability



=



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Impedance of winding on
with core material

=

Impedance of winding
w/o core material

•

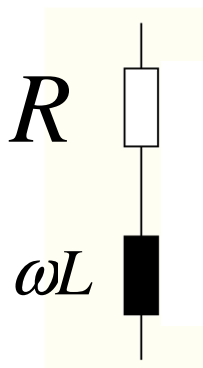
core material

↓
 \underline{Z}

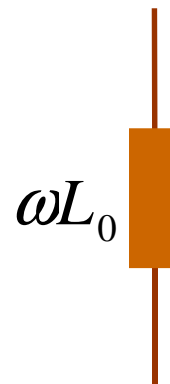
=

↓
 $j \omega L_0$

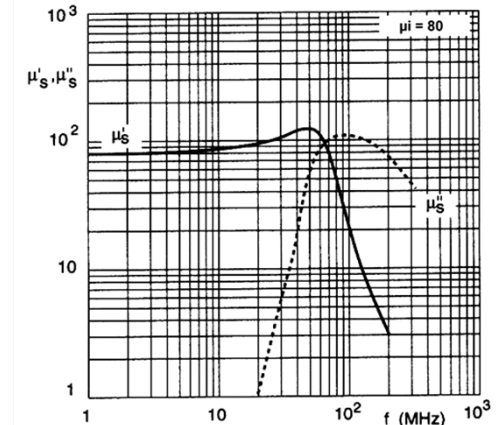
↓
 $(\mu' - j\mu'')$



=

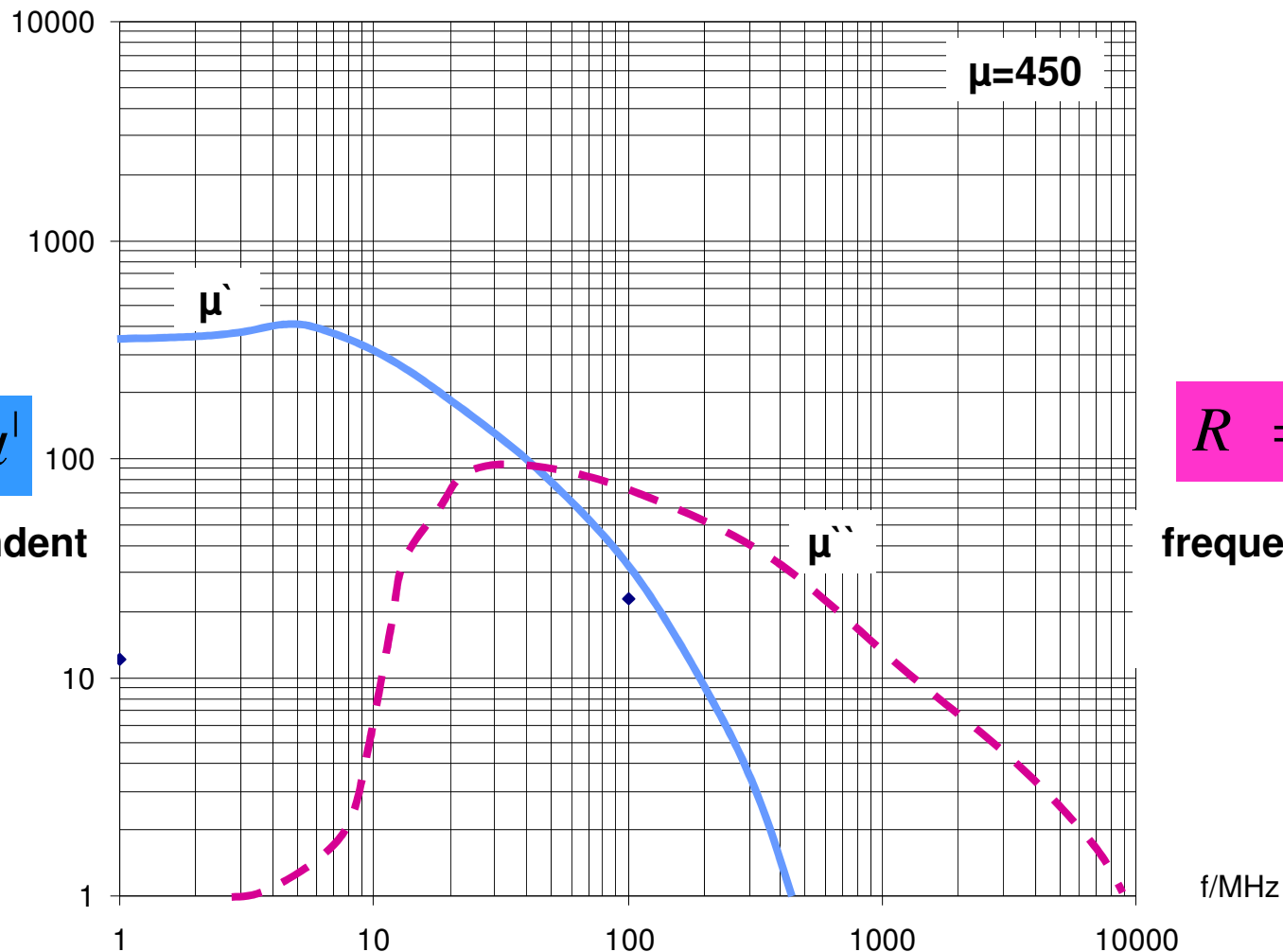


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Permeability – complex permeability

$$\underline{Z} = j\omega L_0 (\mu' - j\mu'') = R + jX$$



$$X_L = j\omega L_0 \mu'$$

**frequency dependent
inductive part**

(Energy storage)

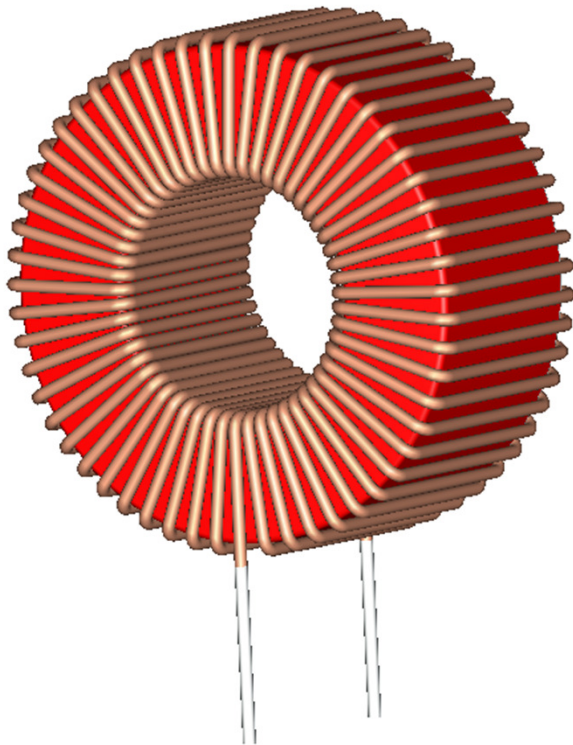
$$R = \omega L_0 \mu''$$

**frequency dependent
core losses**

(Eddy current losses)

Inductance and Core Material Geometry

Toroidal Inductor



$$L = \frac{(\mu_0 * \mu_r * A_{\text{eff}} * N^2)}{l_{\text{eff}}}$$

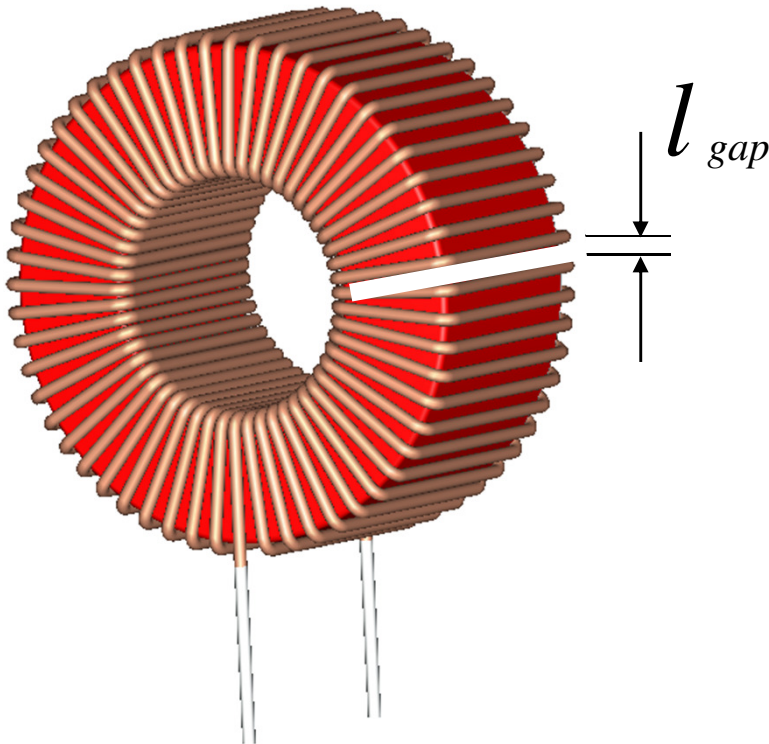
$$\mu_0 = 4 \cdot \pi \cdot 10^{-7}$$

μ_r = relative permeability
 N = No. of turns
 A_{eff} = effective magnetic area
 l_{eff} = effective magnetic length

Inductance and Core Material Geometry

Toroidal Inductor with a gap

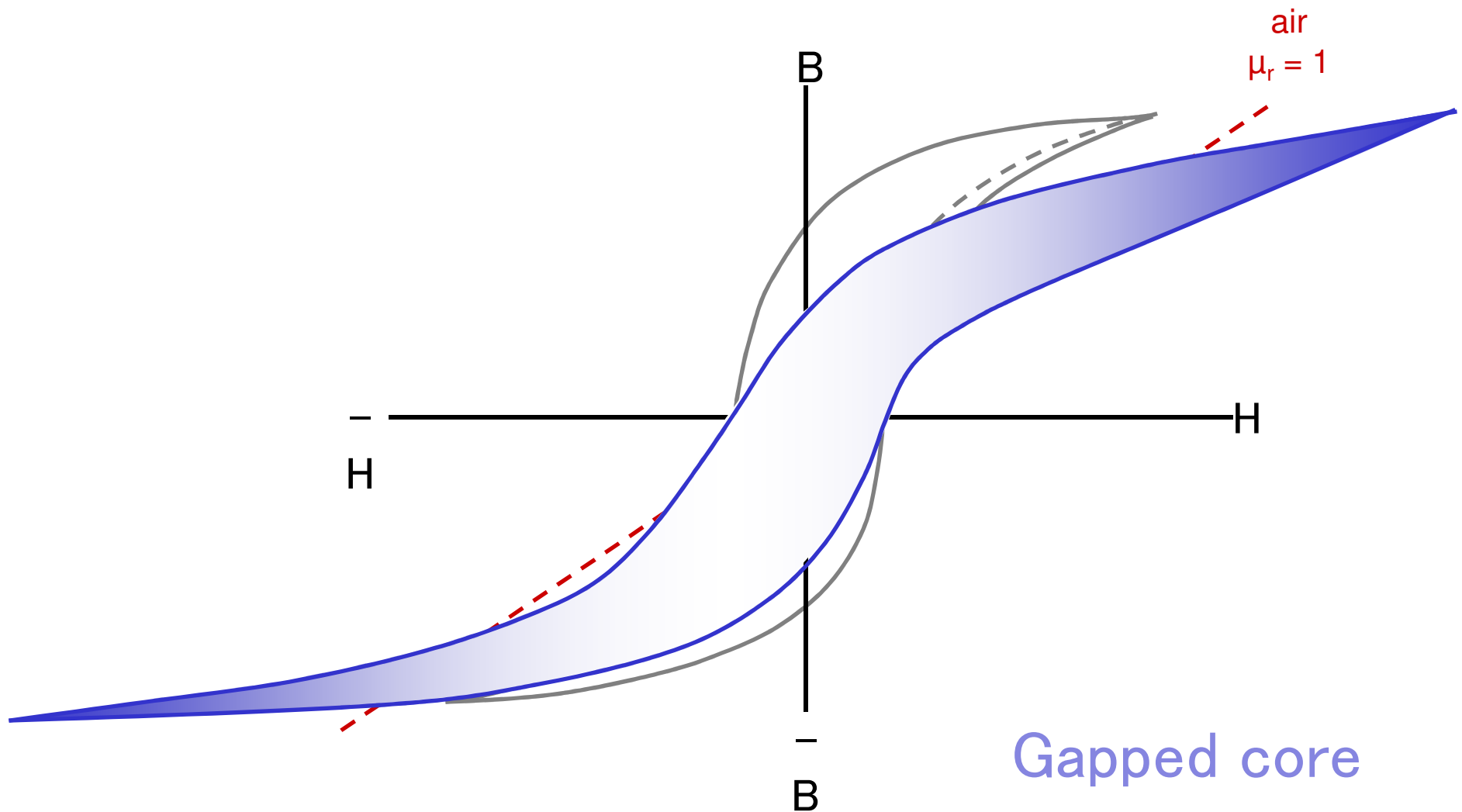
Adding an air gap will increase the effective magnetic length



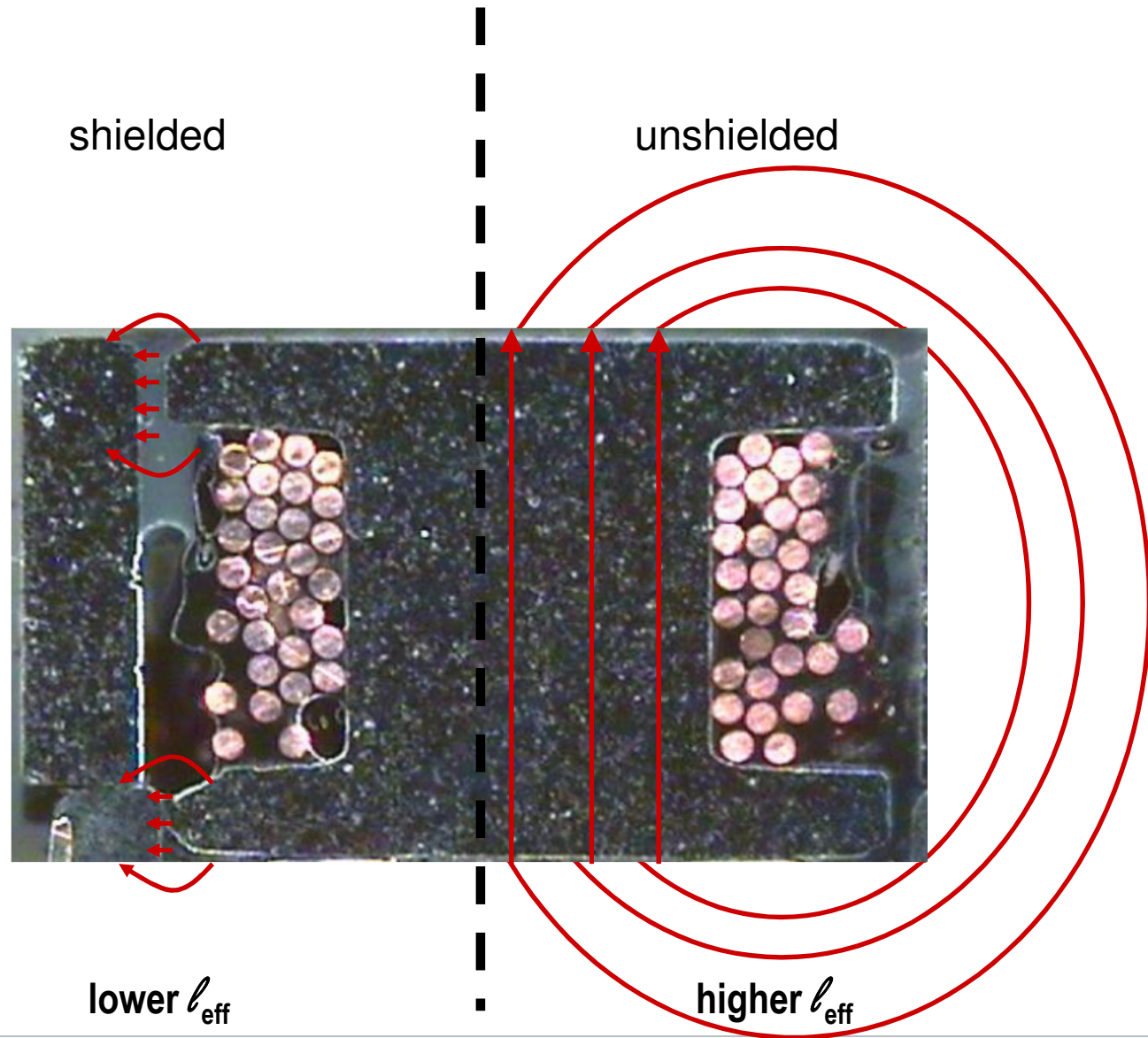
$$L = \frac{(\mu_0 * \mu_r * A_{\text{eff}} * N^2)}{l_{\text{eff}}}$$

$$l'_{\text{eff}} \sim l_{\text{eff}} + l_{\text{gap}} \mu_r$$

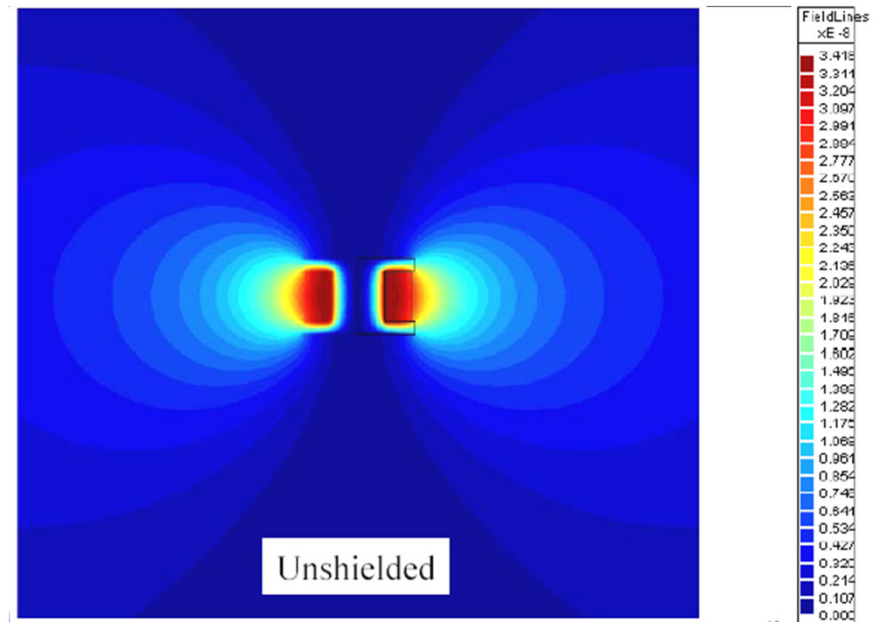
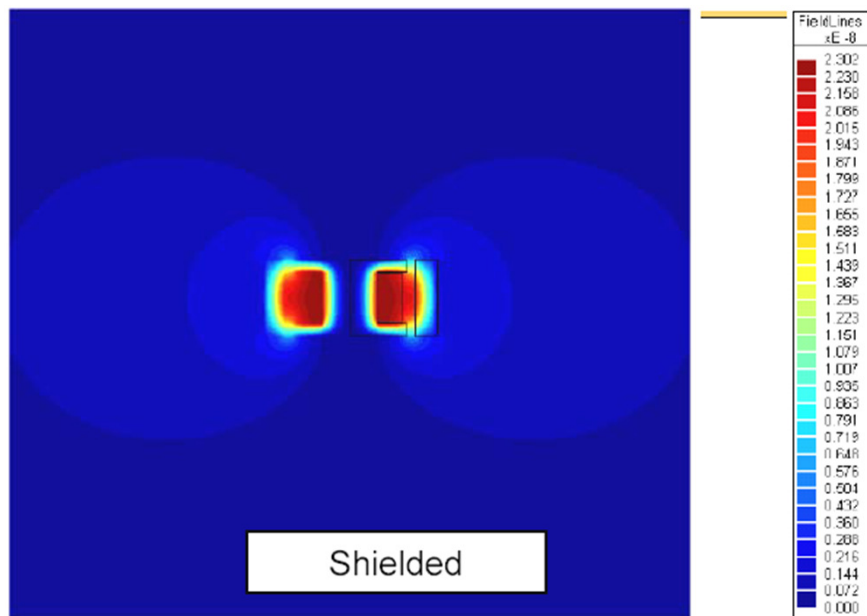
Un-gapped Cores Vs. Gapped cores



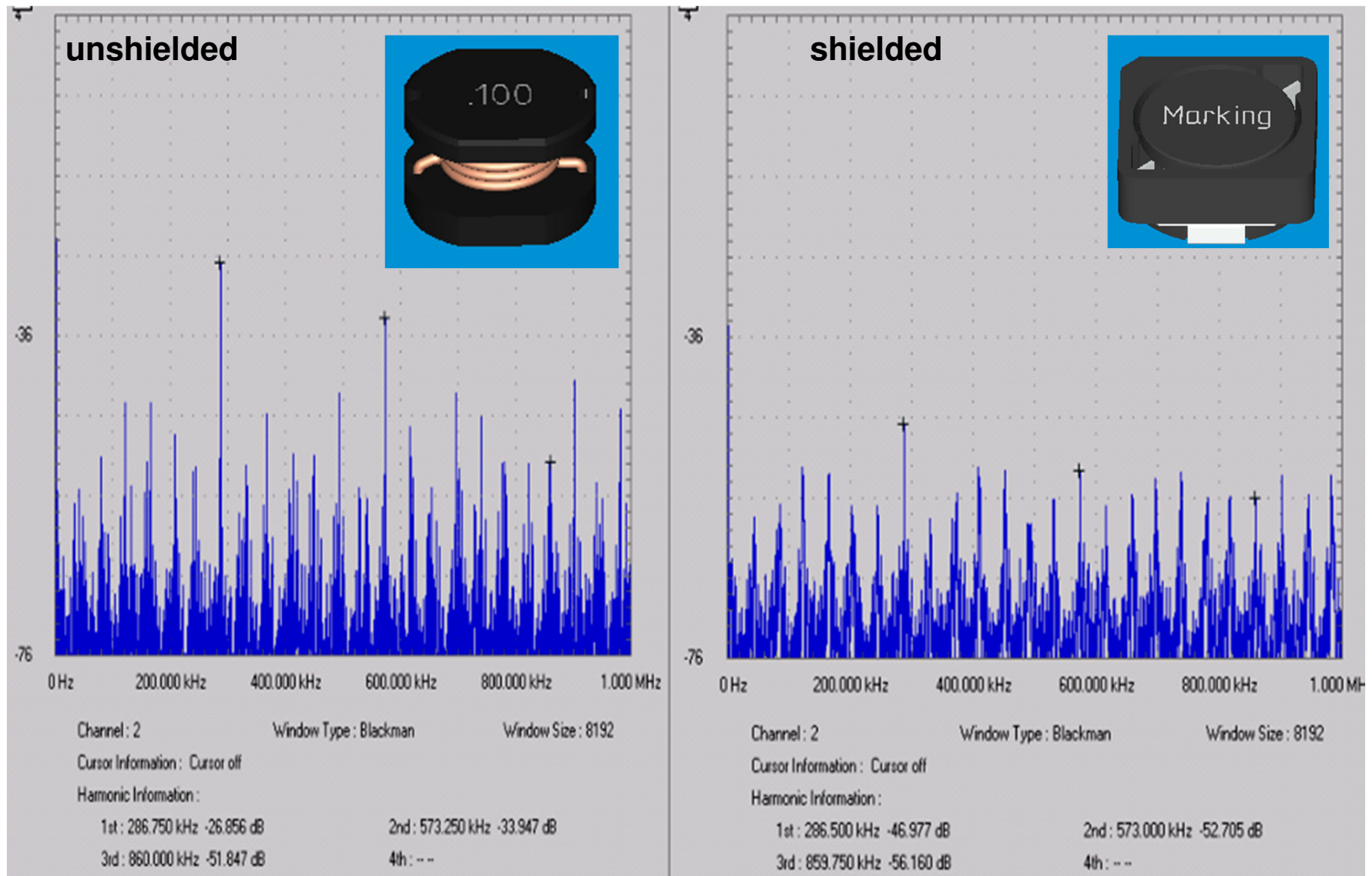
Shielded Vs. Unshielded



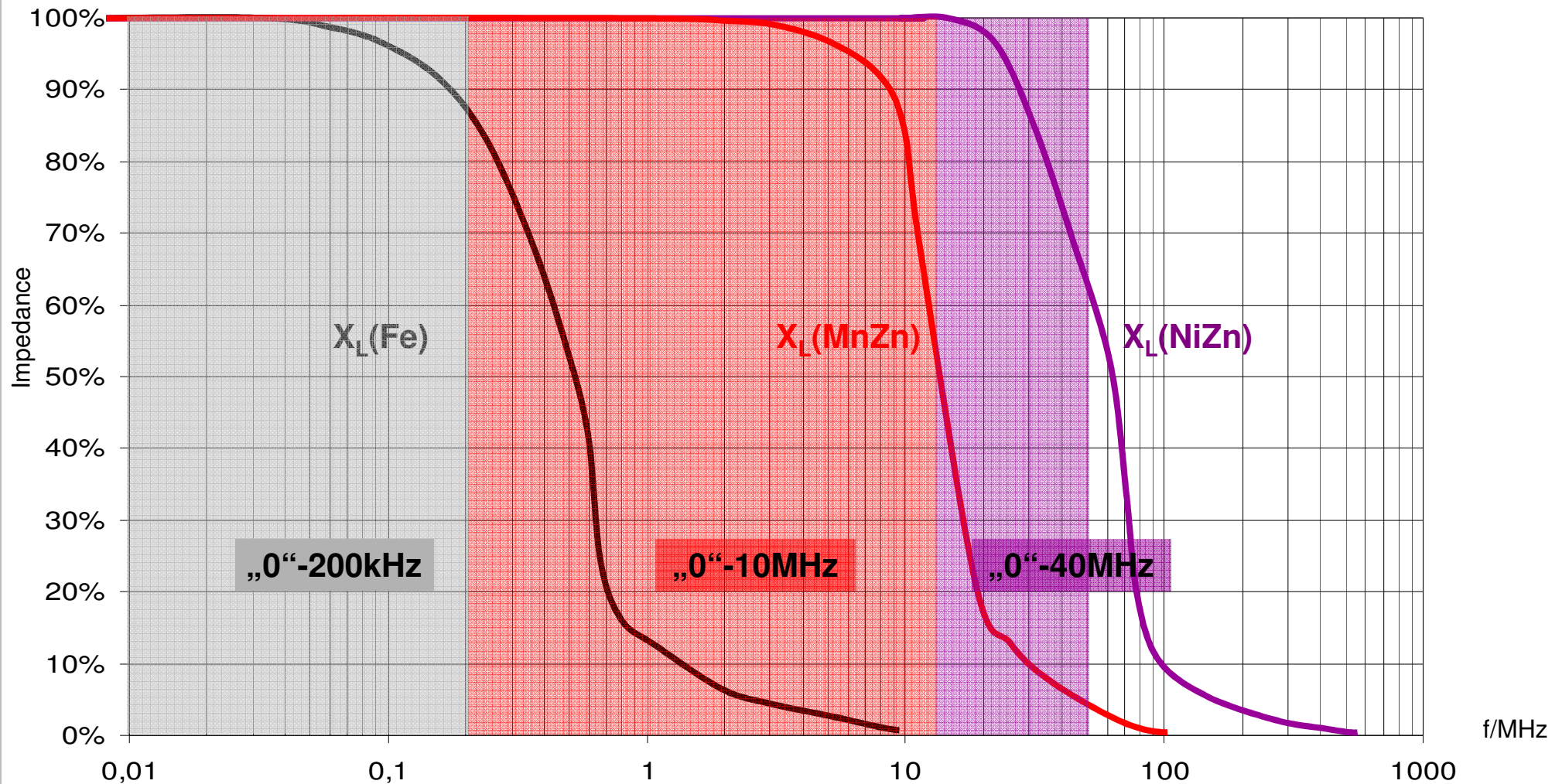
Stray field - Shielded vs. Unshielded



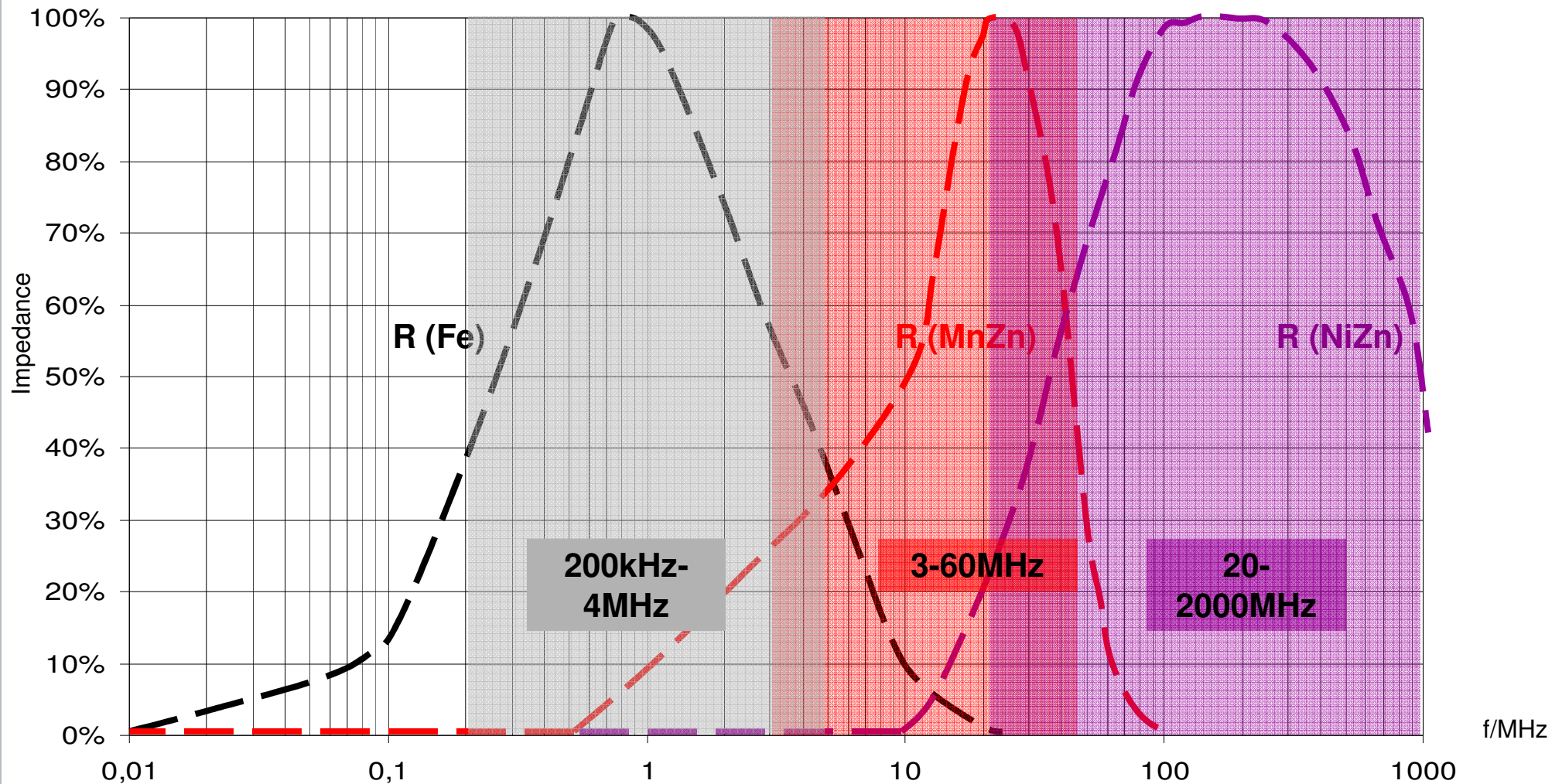
stray field - shielded vs. unshielded



Applications (Inductors for Storage)



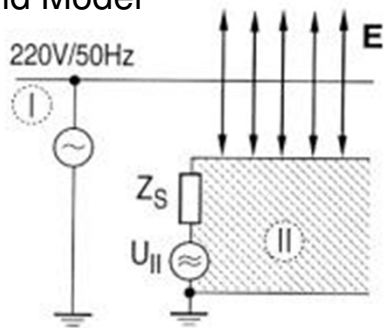
Applications (Inductors for Filtering)



Types of Noise Coupling Mechanisms

Capacitive Coupling

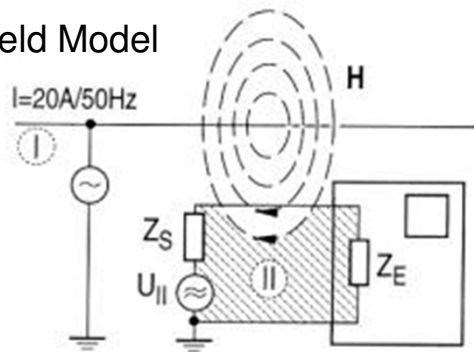
Field Model



Evident in cases where components or traces are at distance of $1/10 \lambda$ (wavelength) of the **signal frequency**

Inductive Coupling

Field Model



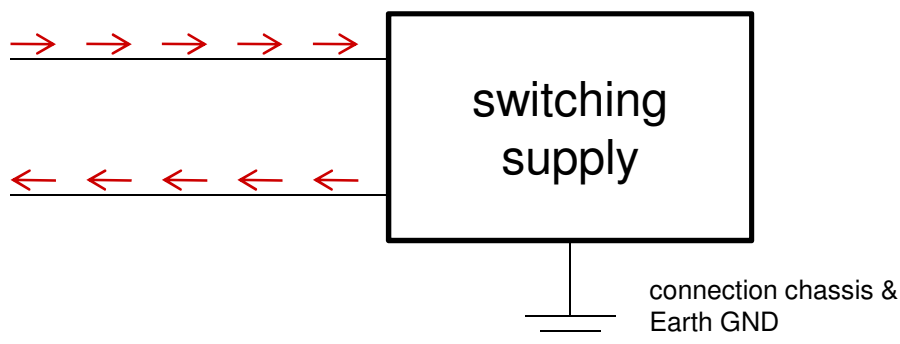
Evident in cases where components or traces are at a distance of $1/4 \lambda$ (wavelength) of the **noise signal**

COMMON MODE CHOKES

Types of Noise Signals

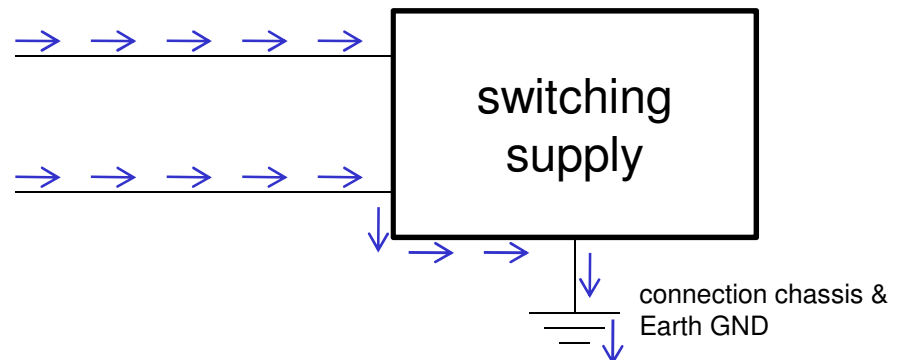
Differential–Mode signal

- Noise flows into one line and exits through another
- Independent from GND



Common–Mode signal

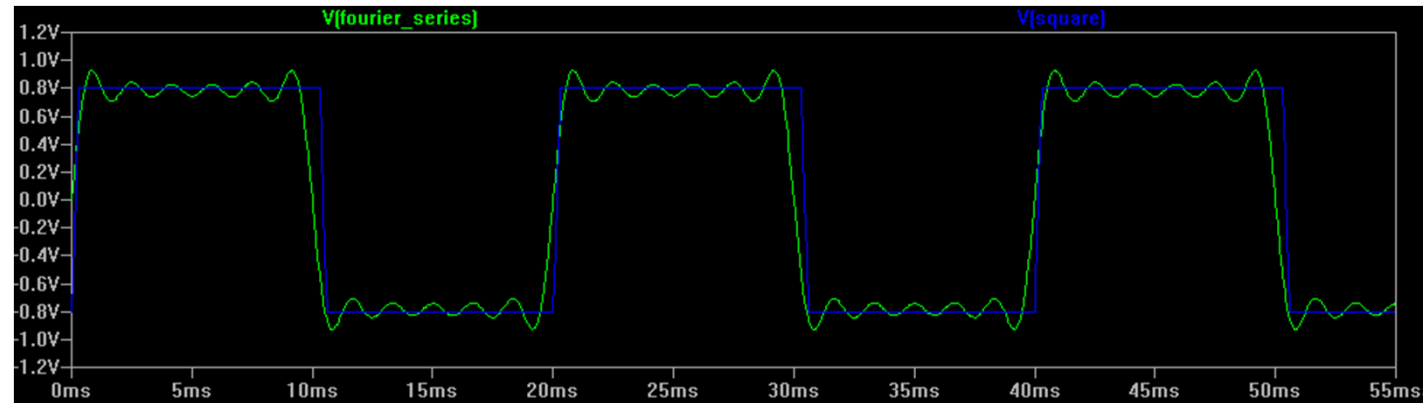
- Noise flows along both lines in the same direction
- returns by some parasitics path through system GND



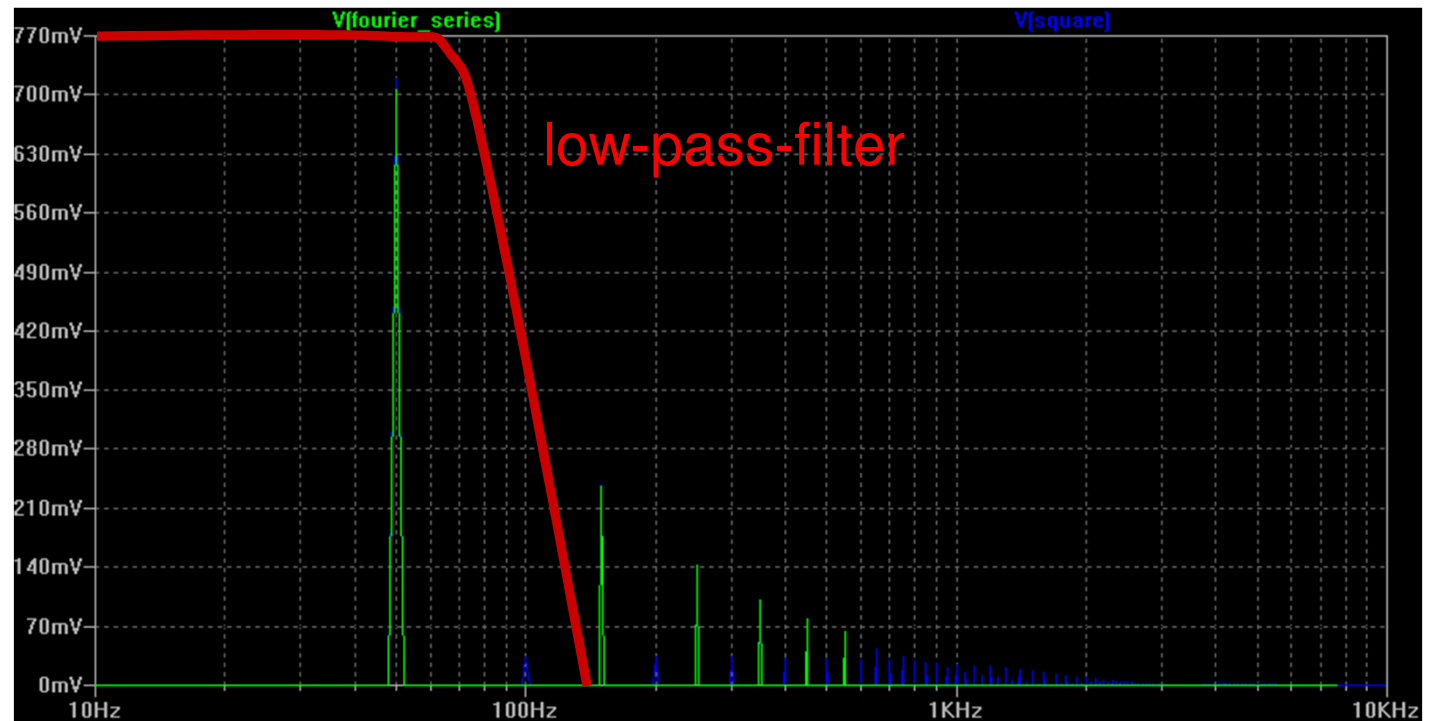
Common mode choke - advantages



time based



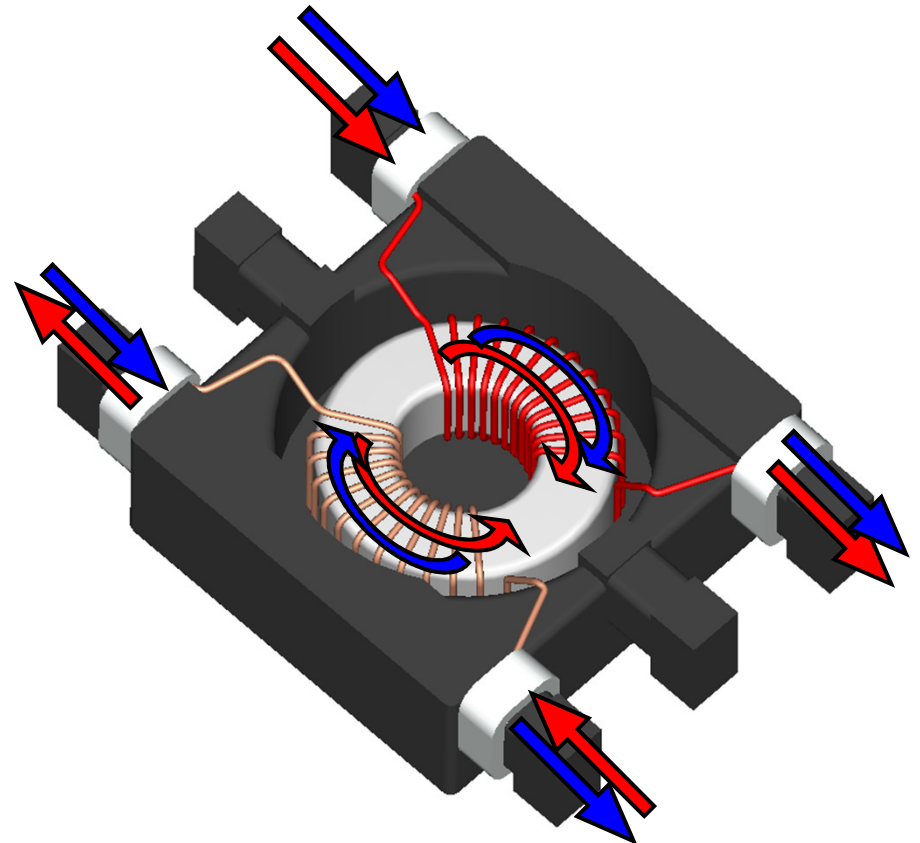
frequency based



Common Mode Filter

Reduction of noise

- From device to environment
- From environment to device



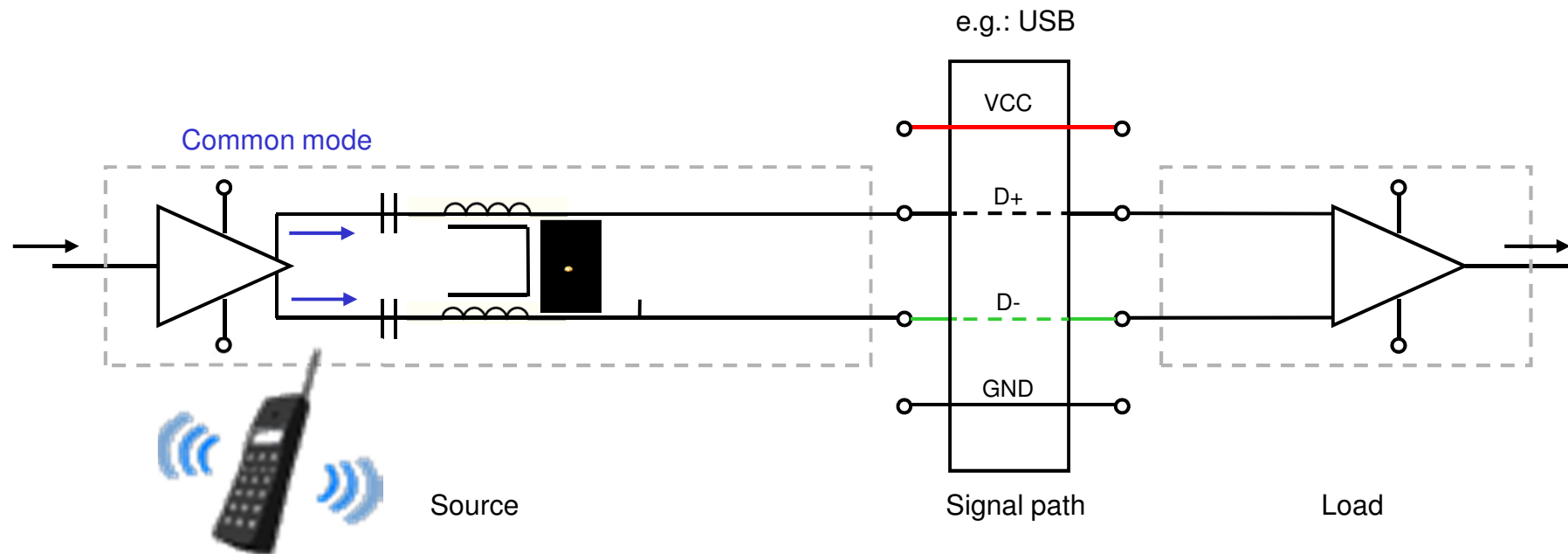
Conclusion:

- “Almost” no influencing of the signal
→ Differential mode
- High attenuation of noise

→ Common mode

Common Mode Filter – Signal theories

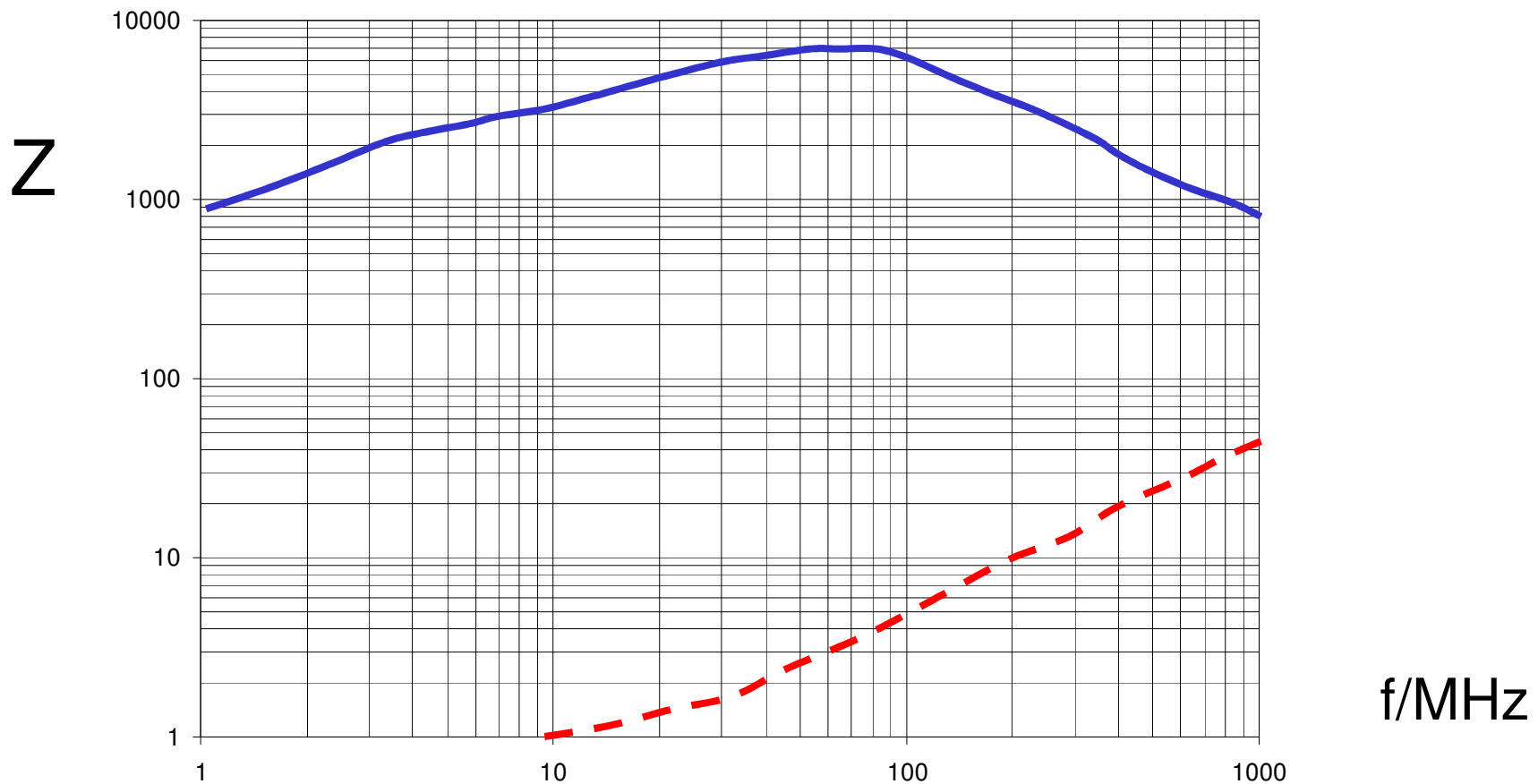
Filtering



Common Mode Filter



- The **Common Mode** Impedance attenuates just the **noise**



- The **Differential mode**-Impedance will also attenuate the **signal**

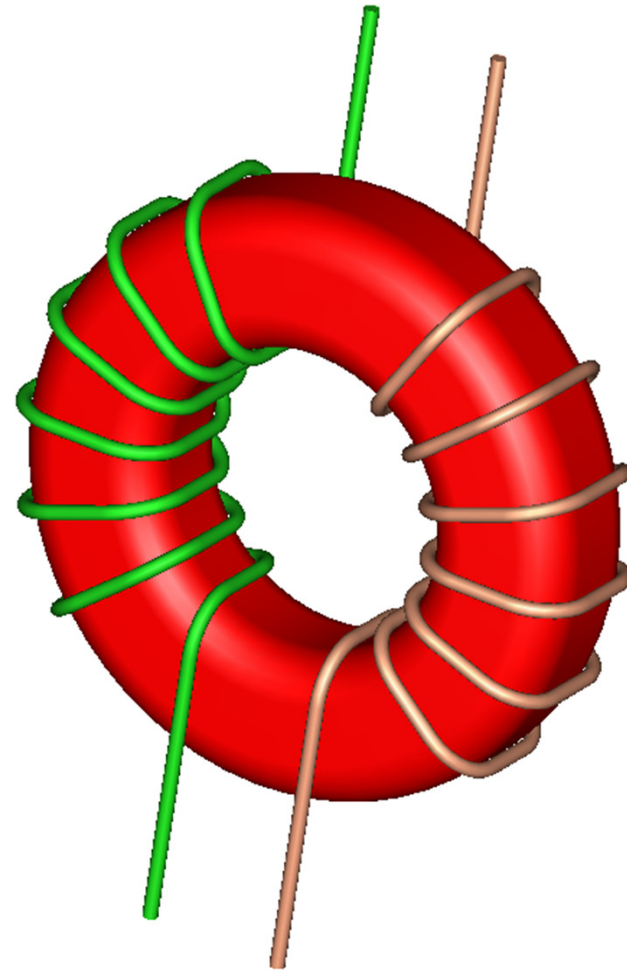
Common Mode Choke Construction



bifilar



sectional



Common mode choke - construction



Bifilar

- Less differential impedance
- High capacitive coupling
- Less leakage inductance

- Data lines
→ USB, Fire-wire, CAN, etc.
- Power supply
- Measuring lines
- Sensor lines



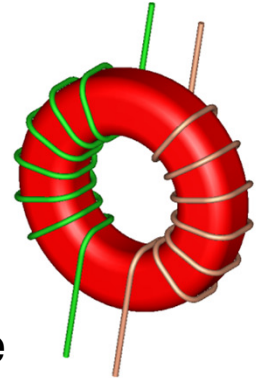
• WE-CNSW



• WE-SLM

Sectional

- Low capacitive coupling
- High leakage inductance
- High differential impedance

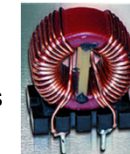


- Power supply input /output filter
→ CMC for mains power
- High voltage application
- Measuring lines
- Switching power supply decoupling



• WE-LF

• WE-SLx-Series

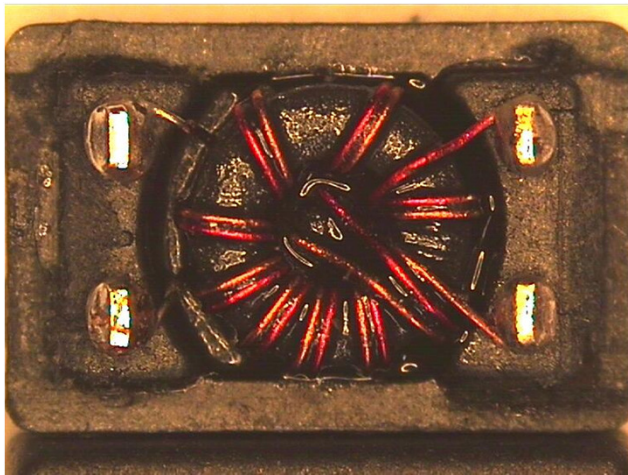


• WE-CMB

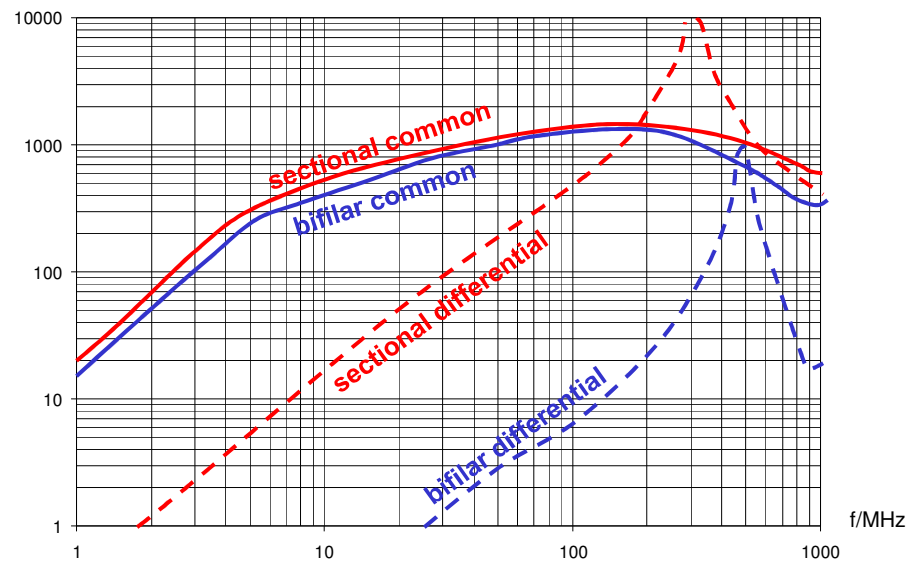
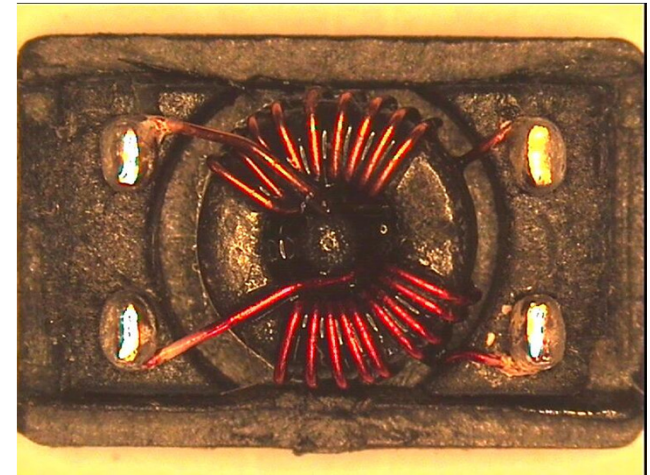
• WE-VB / VB2

Common mode choke - construction

bifilar winding

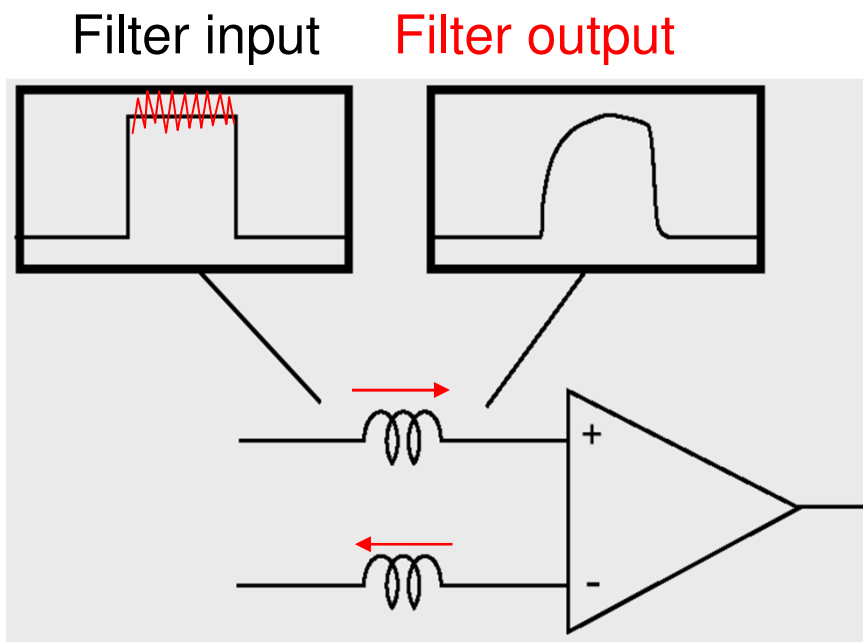


sectional winding

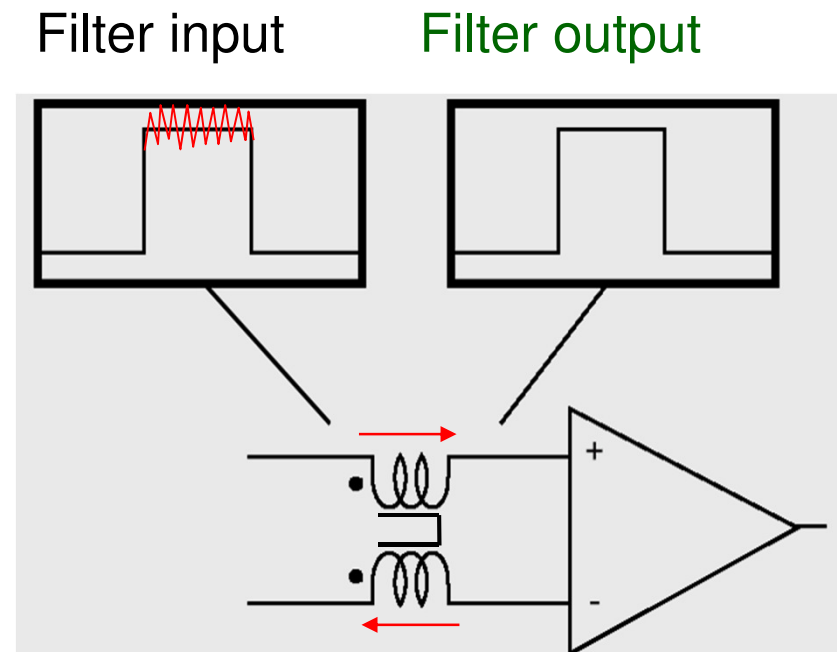


Common mode choke - advantages

Filter with two inductors

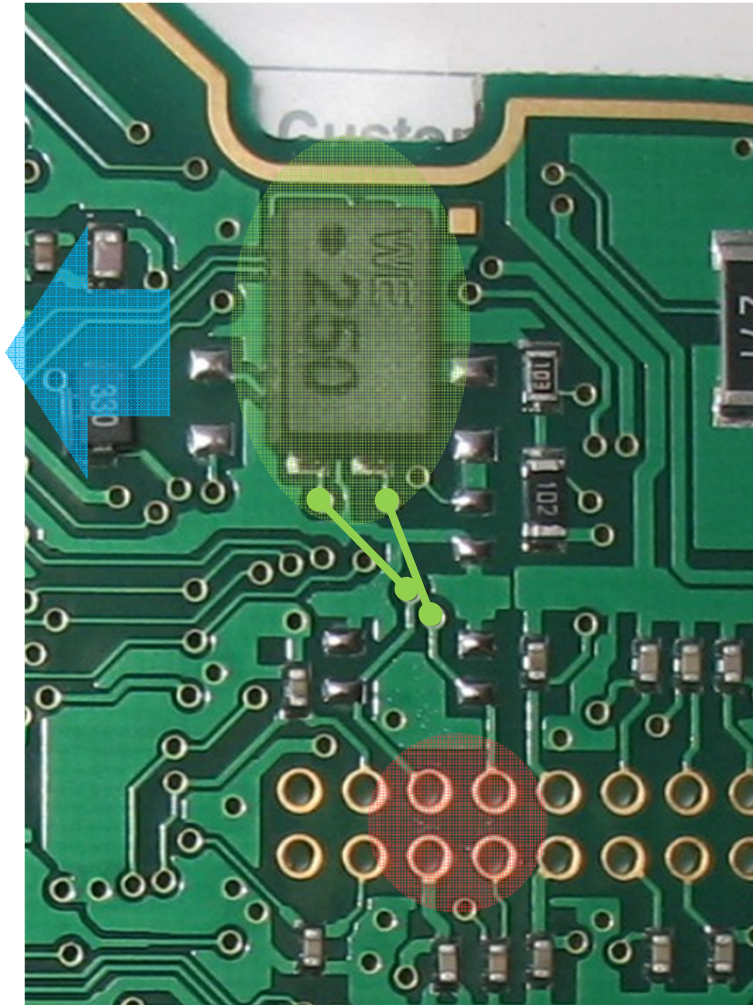


Filter with CMC



- no signal distortion – means signal will be not affected

Common mode choke - advantages



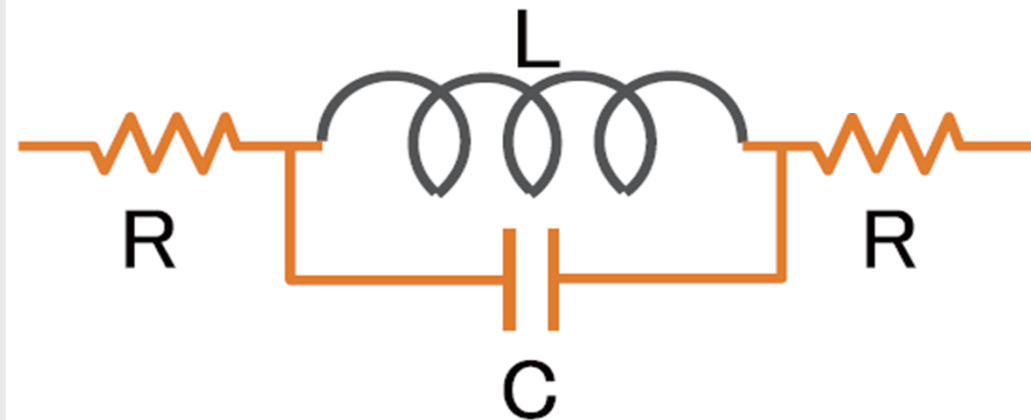
Digital logic device

Common Mode Choke

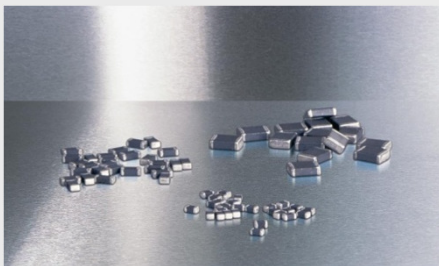
Interconnection → Cable/Connector

CHIP BEAD FERRITES

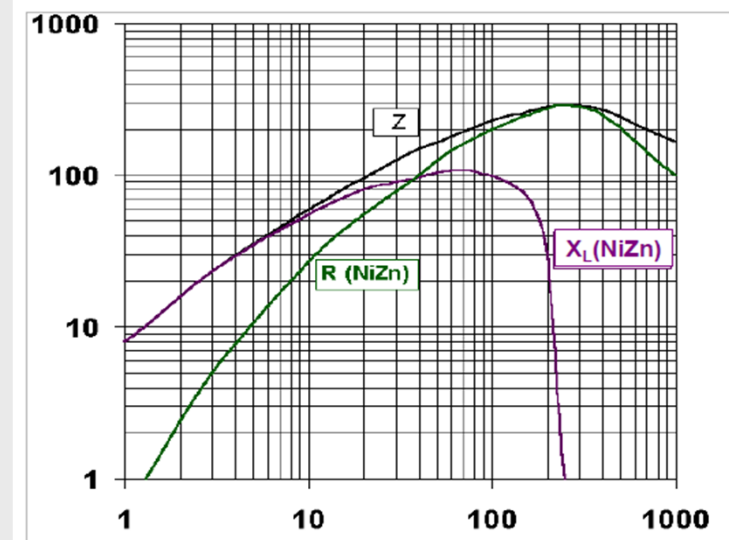
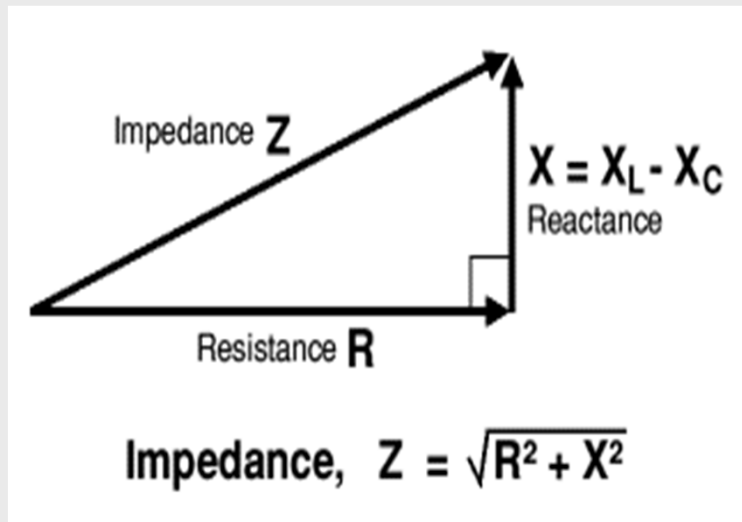
Chip Bead Ferrites



- Inductor
 - Used to filter unwanted noise
 - Supply voltage lines, ground planes, and data signals
- Frequency Dependant
 - Also know as a frequency dependent resistor



Chip Bead Ferrites Impedance

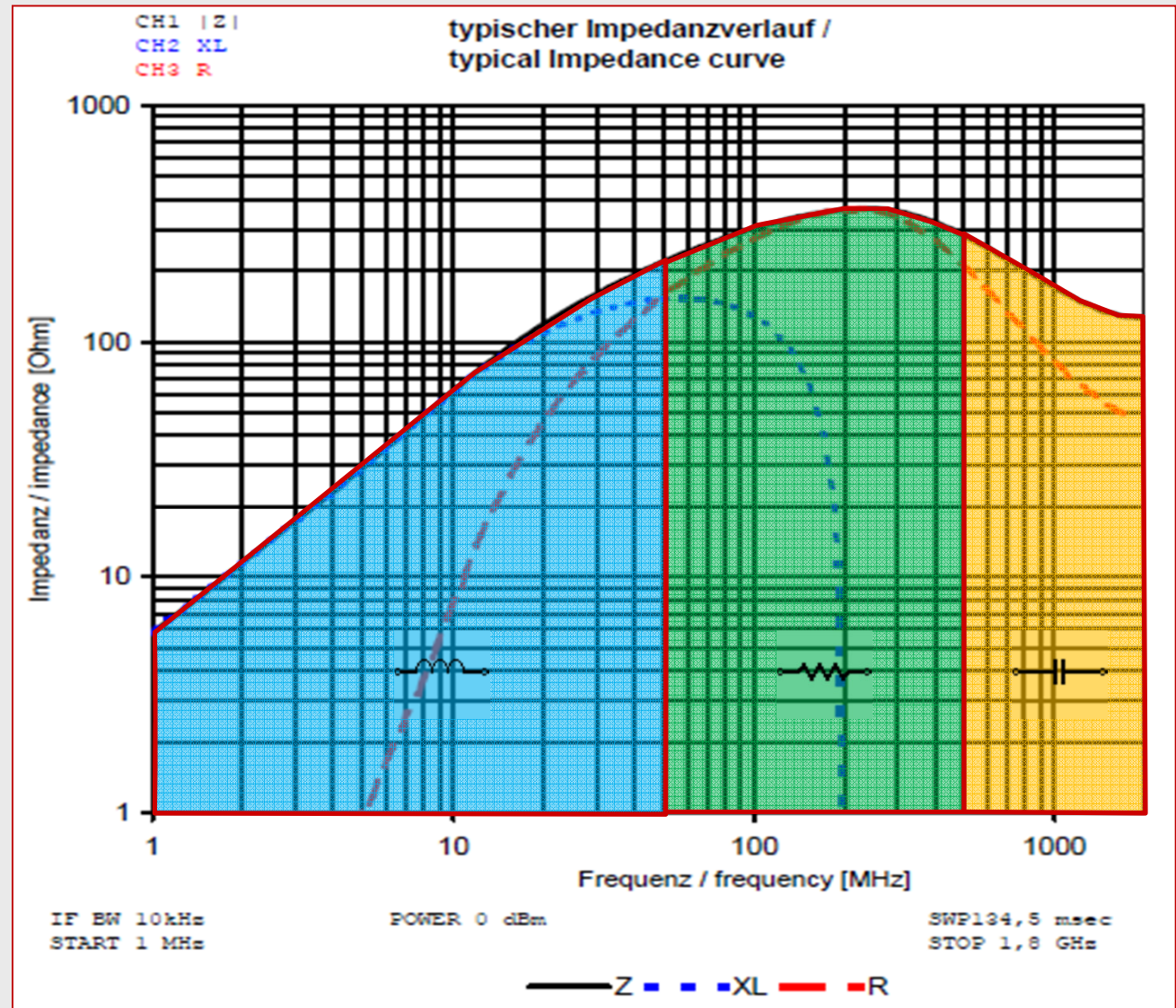


- AC resistance equivalent
 - Impedance can be split into two parts
 - Resistance R (Constant regardless of frequency)
 - Reactance X (Varies with frequency)

Chip Bead Ferrites

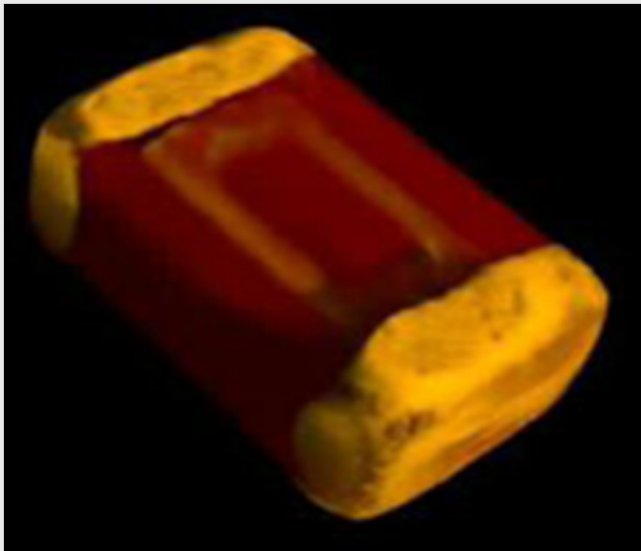
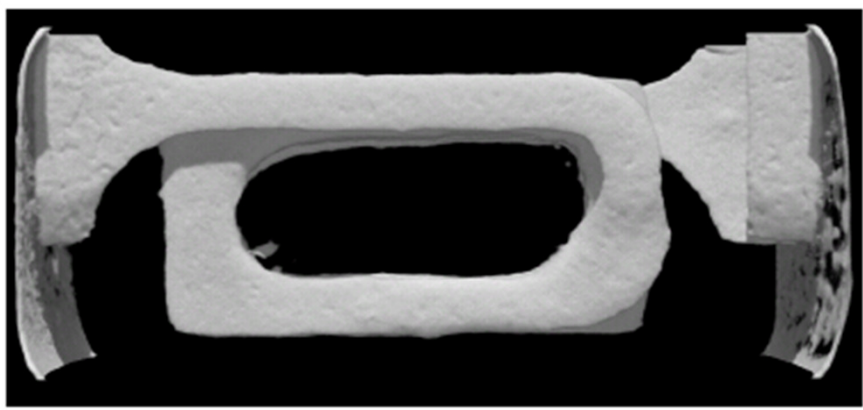
- **Resistive**
 - Impedes the noise energy and absorbs it
- **Inductive**
 - Energy storage
- **Capacitive**
 - Passes AC, blocks DC

SRF $X_L = X_C$

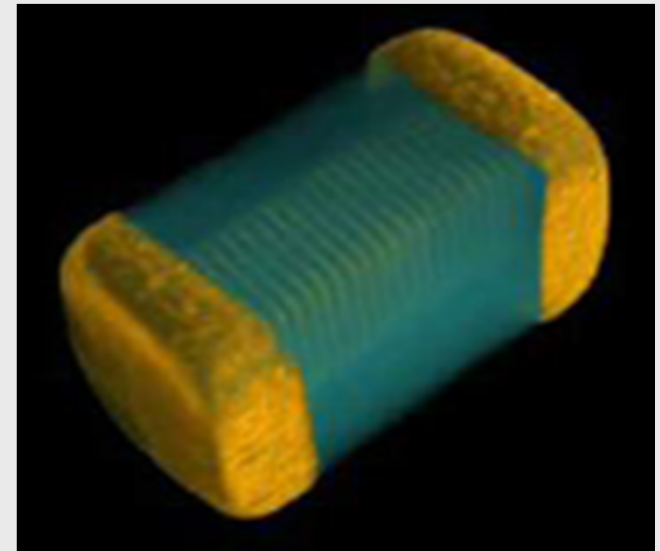
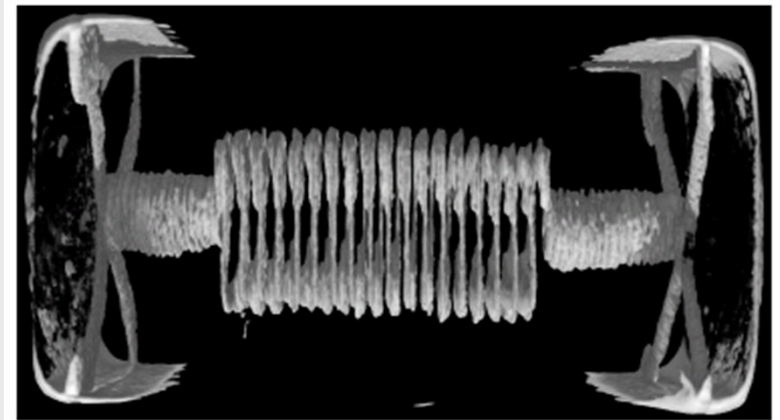


Chip Bead Ferrites

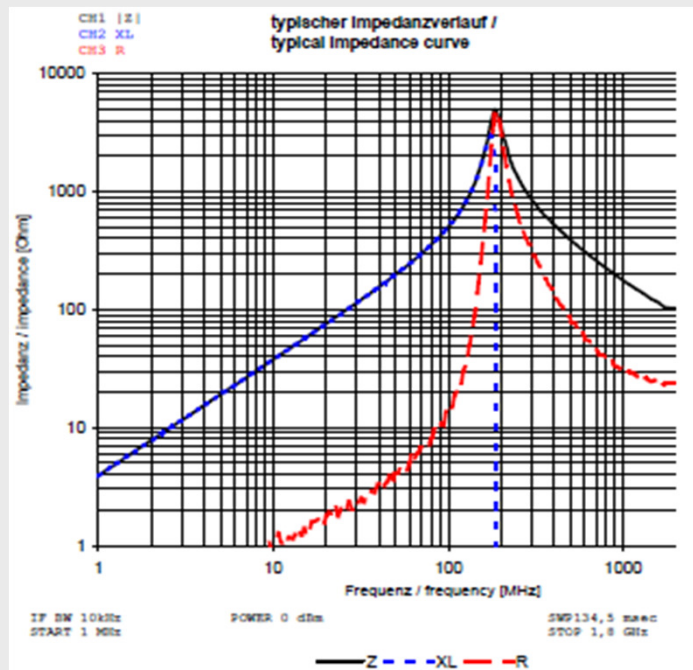
High Current



High Frequency

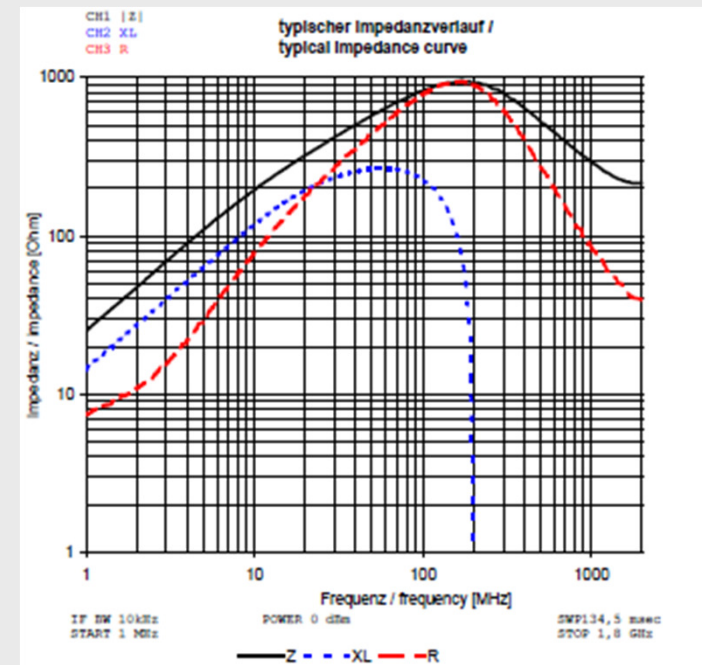


Chip Bead Ferrites



High speed

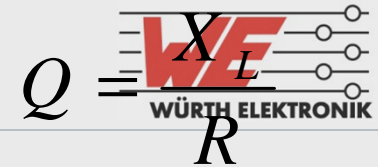
- Lower impedance in lower frequency
 - Low attenuation for fast signals
- Applications: USB 2.0, IEEE 1394, LVDS



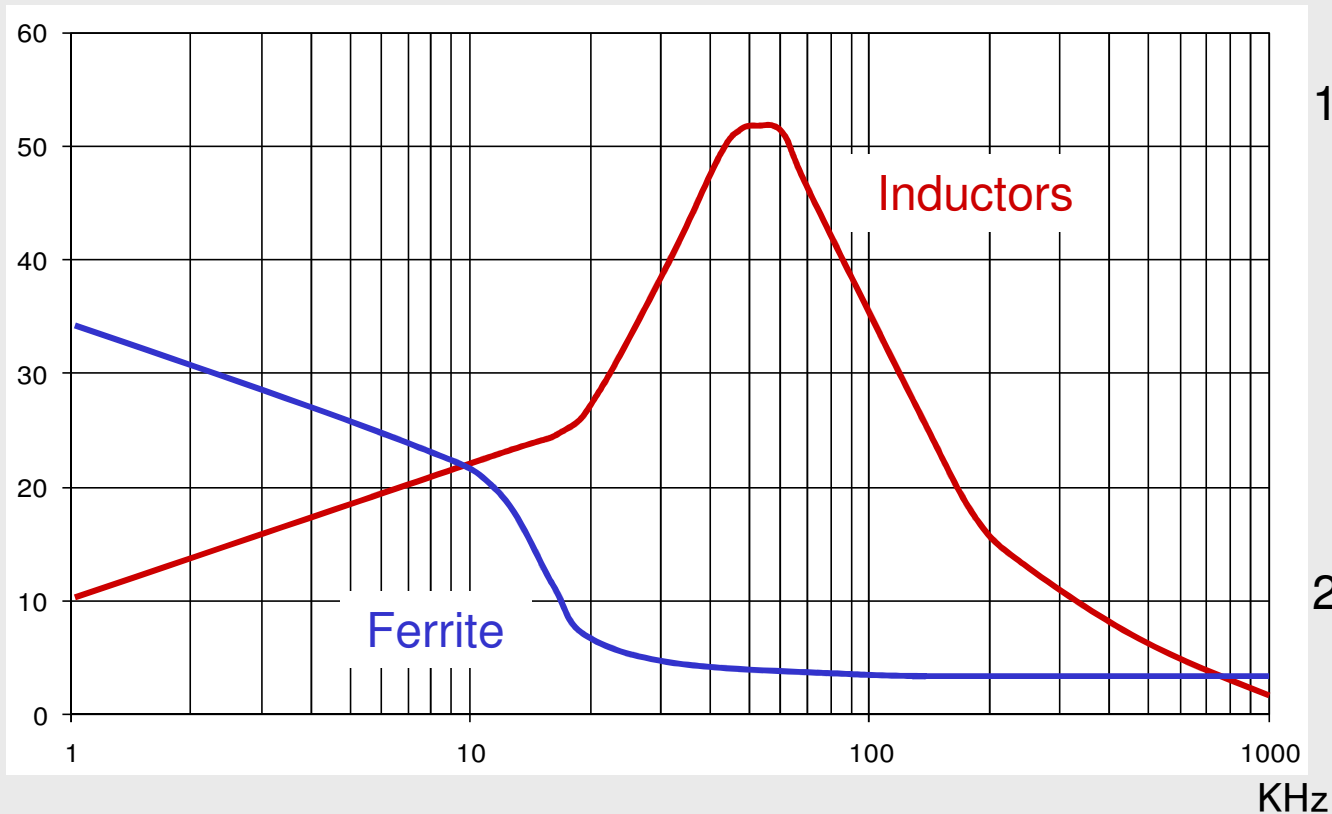
Wide band

- High impedance in low frequency
- Wide band through the entire spectrum
 - Applications: DC/DC converters

Q factor Application



Q



1. Application: **Storage inductor**

1. Request: **Lowest possible core losses at switching frequency (HIGH Q)**

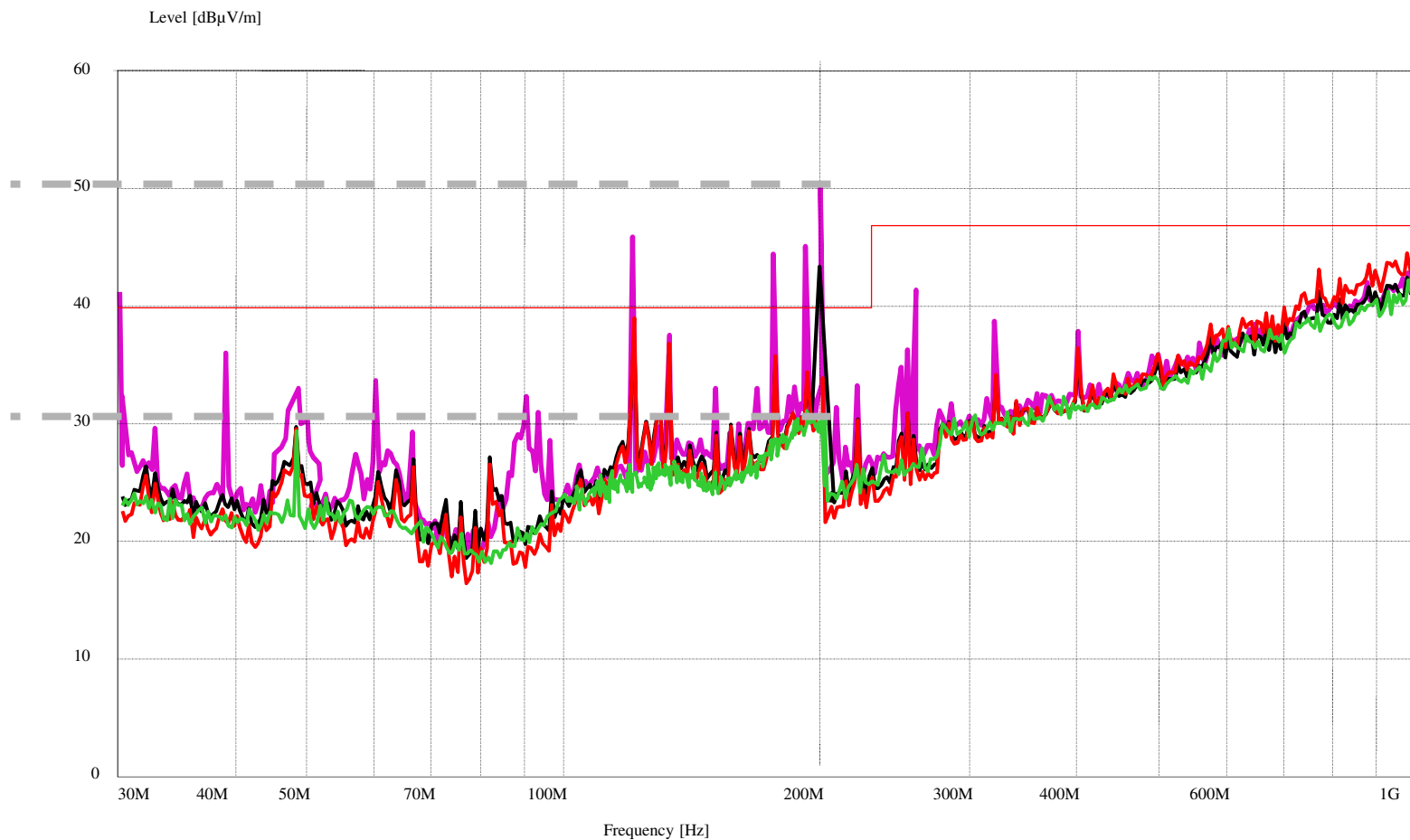
2. Application: **Absorber / Filter**

1. Request: **Highest possible core losses at application frequency (LOW Q)**

Insertion loss - example

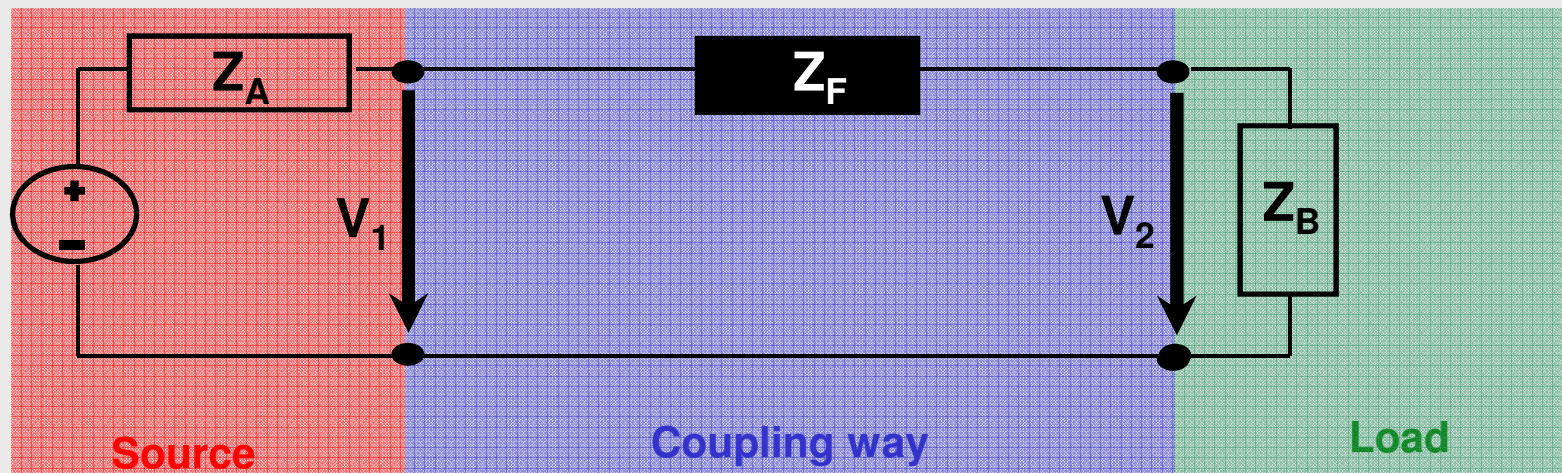
- check the results

→ measuring the emission and compare the attenuation



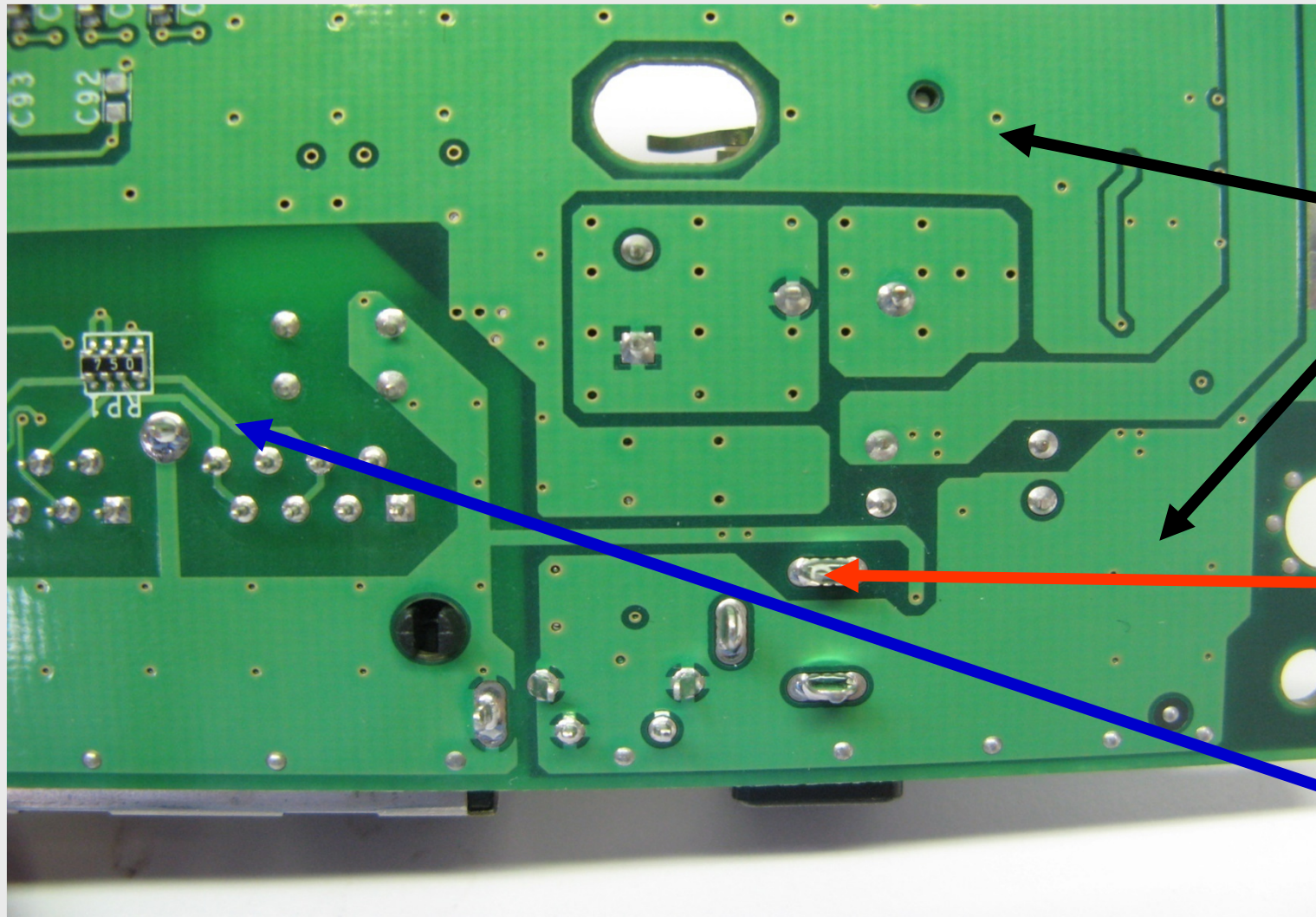
Insertion loss calculation

The logarithmic ratio of input power to output power, which, describes the signal attenuation along a defined transmission patch.



- System attenuation
$$A = 20 \cdot \log \frac{Z_A + Z_F + Z_B}{Z_A + Z_B} \text{ in (dB)}$$
- Impedance
$$Z_F = \left[10^{\frac{A}{20}} \cdot (Z_A + Z_B) \right] - (Z_A + Z_B) \text{ in } (\Omega)$$

Insertion loss calculation



1 Ω
Ground plane

10 Ω
Supply voltage line

50 Ω - 90 Ω
Data signal lines

Insertion loss calculation

$$Z_F = \left[10^{\frac{A}{20}} \cdot (Z_A + Z_B) \right] - (Z_A + Z_B)$$

$$Z_F = \left[10^{\frac{12}{20}} \cdot (10_A + 10_B) \right] - (10_A + 10_B)$$

$$Z_F = 59.6 \Omega$$

$$A = 20 \log \frac{Z_A + Z_F + Z_B}{Z_A + Z_B}$$

$$A = 20 \log \frac{10_A + 59.6_F + 10_B}{10_A + 10_B}$$

$$A = 20 \log 3.98$$

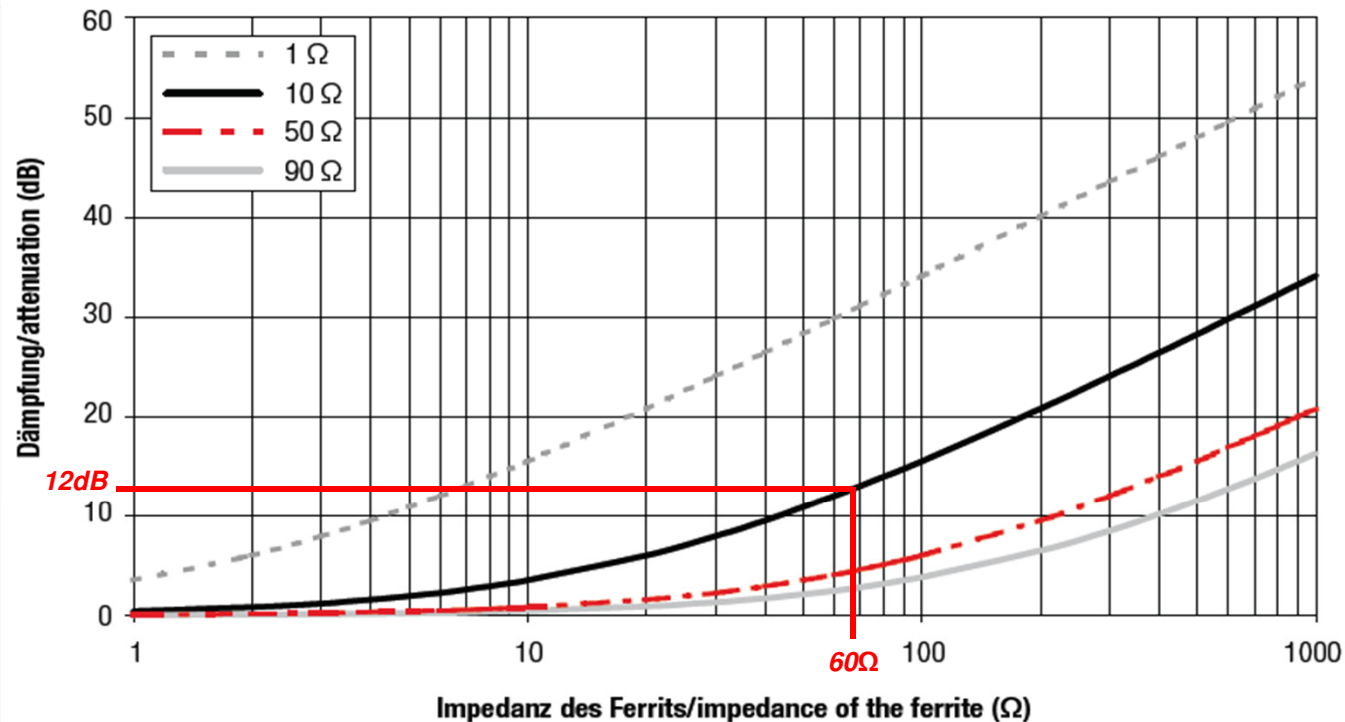
$$A = 11.99 \text{ dB}$$

1. Require 12dB of attenuation at 125 MHz
2. Know that it is a power cable
3. Power port has 10 Ω impedance
4. Result is a impedance of 60 Ω

Application overview

| Assumend practical system impedance | Application |
|-------------------------------------|---|
| 1 Ω | GND (Ground Planes) |
| 10 Ω | V _{cc} (Supply Voltage lines) |
| 50 Ω – 90 Ω | Datasignal Lines/Clock/ Video Signal/USB |
| 90 Ω – 150 Ω | Long Datasignal Lines |

Insertion loss calculation



Application overview

| Assumed practical system impedance | Application |
|------------------------------------|---|
| 1 Ω | GND (Ground Planes) |
| 10 Ω | V _{cc} (Supply Voltage lines) |
| 50 Ω – 90 Ω | Datasignal Lines/Clock/Video Signal/USB |
| 90 Ω – 150 Ω | Long Datasignal Lines |

1. Require 12dB of attenuation at 125 MHz
2. Know that it is a power cable
3. Power port has 10 Ω impedance
4. Result is a impedance of 60 Ω

Errors are generally due to wrong system impedance estimation

→ Too much attenuation. Reduce the impedance of ferrite

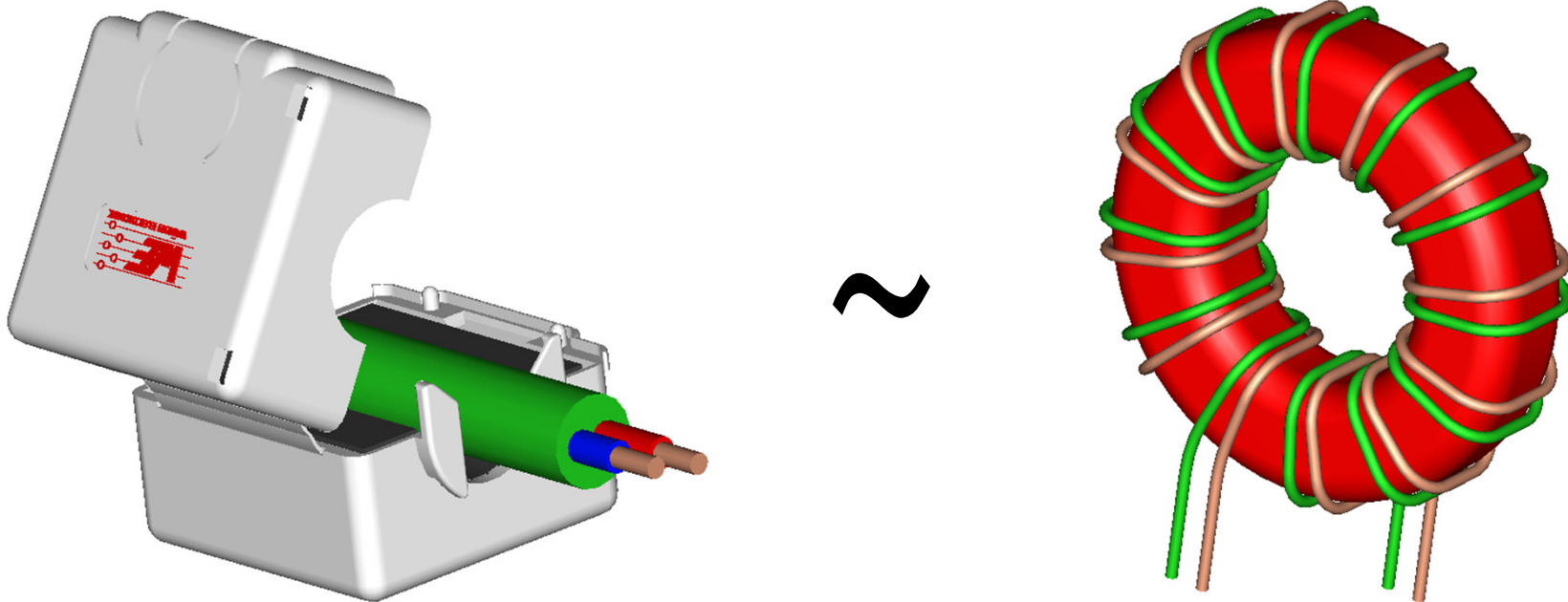
→ Too little attenuation. Increase the impedance of ferrite

CLAMP-ON FERRITES

Clamp On Ferrites



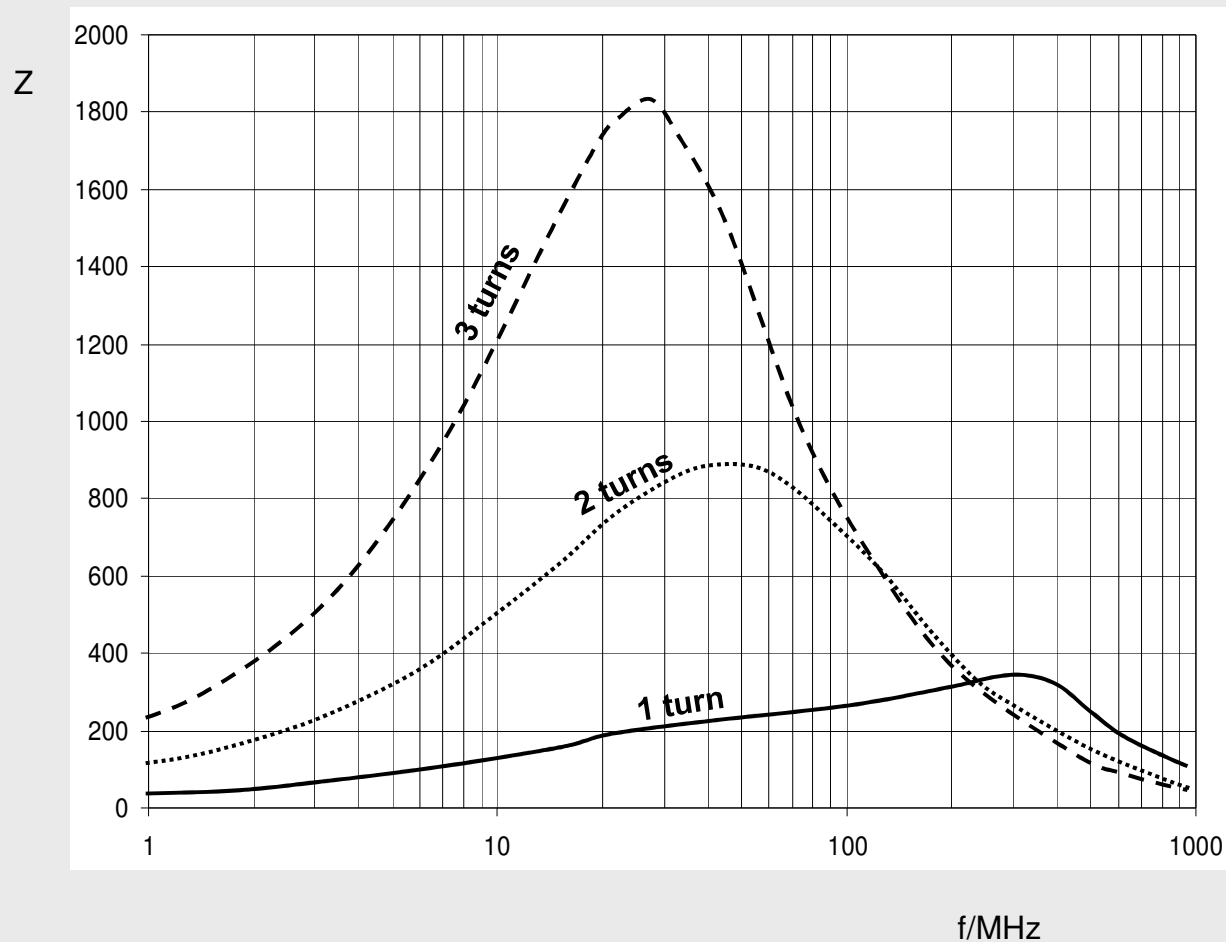
Comparable Performance to a bifilar winding Common Mode Choke



- Both will absorb Common Mode interferences
- Clamp on Ferrite is a Common Mode Choke with ONE winding

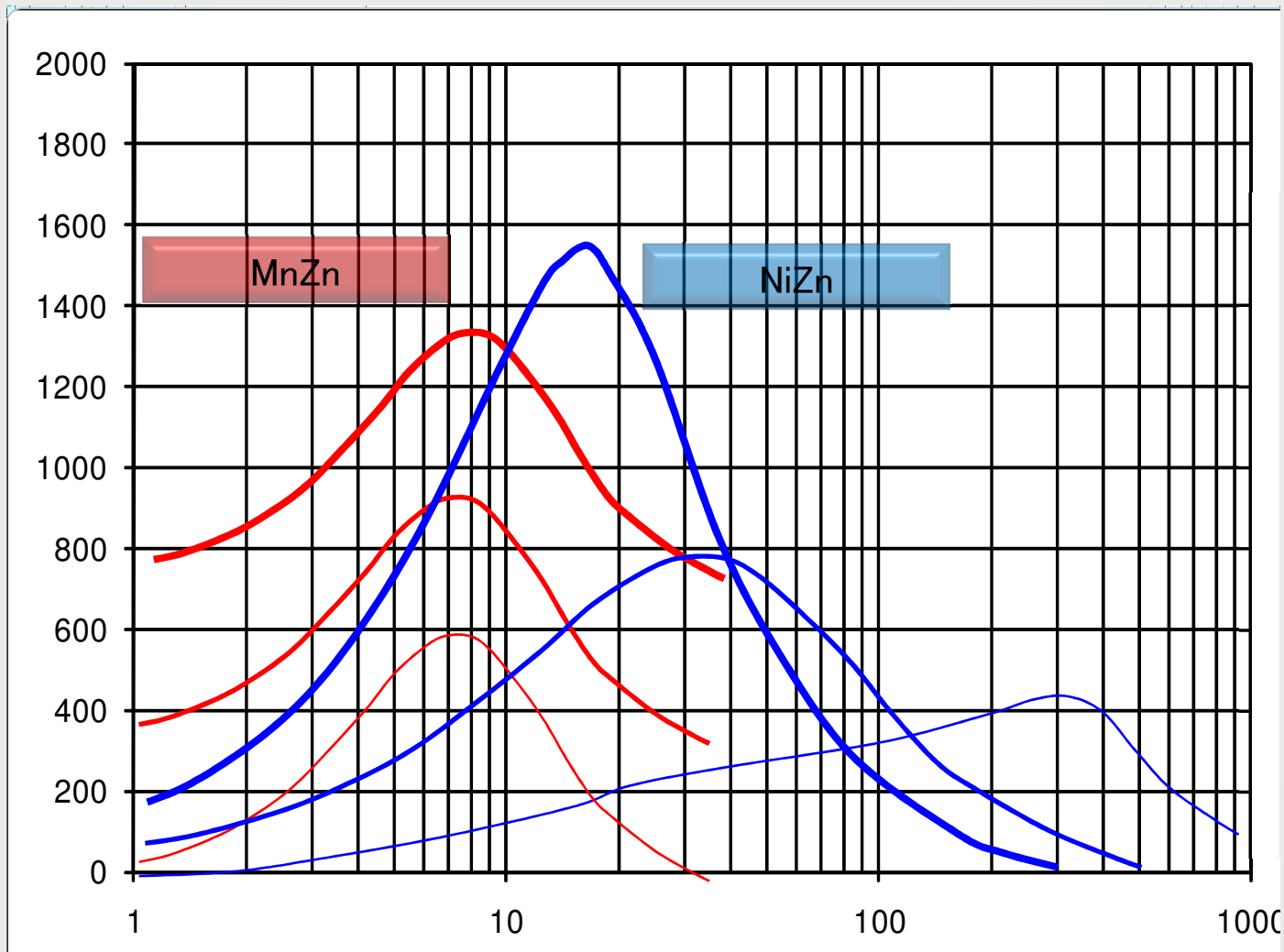
Clamp on Ferrites – Differential mode noise

- Can also be used as a differential mode filter.
 - In this example, it is single wire.



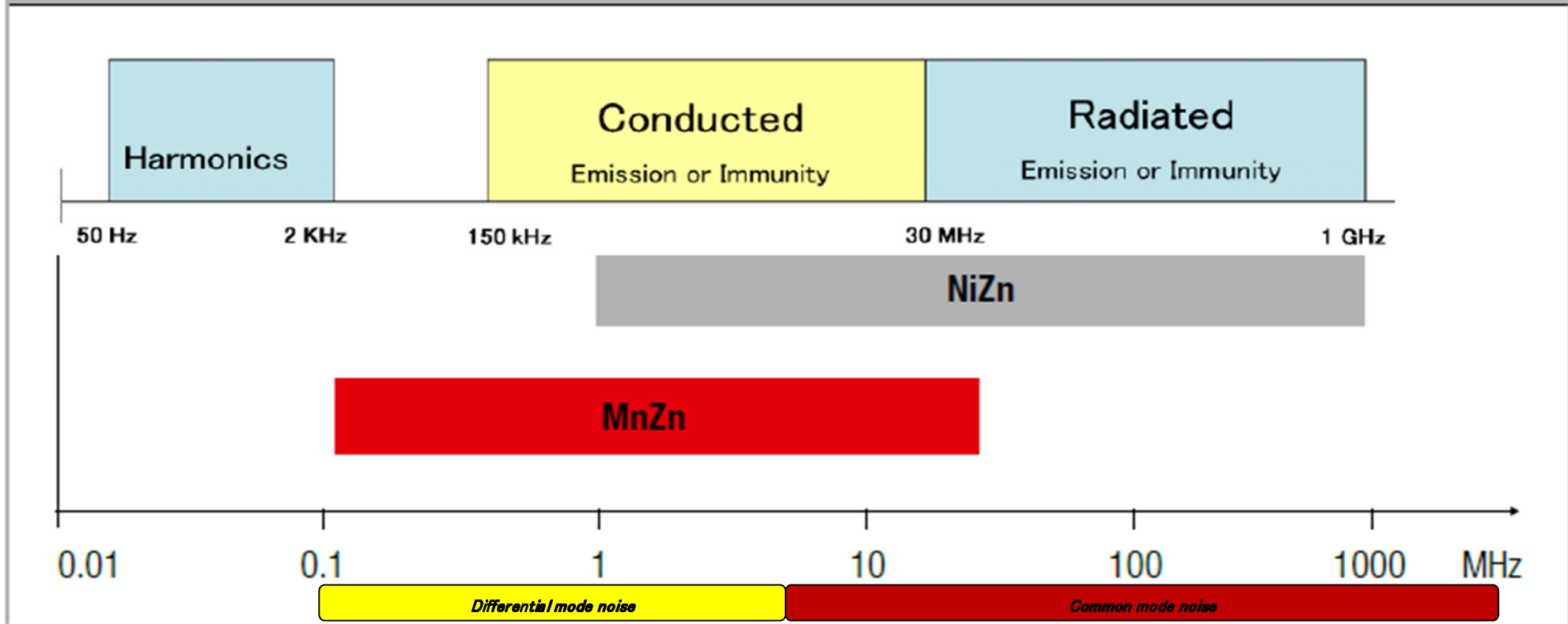
■ 2 turns

Clamp on Ferrites. Material Comparison



Clamp on Ferrites

Material characteristics



How can we find out what interference we have?

Common mode or differential mode?

Take a Snap Ferrite and fix it on the cable
(both lines e.g. VCC and GND)

if noise is reduced or
noise immunity
is increased

Common Mode interferences

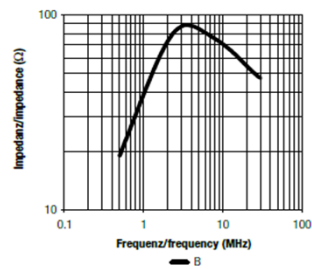
if not

Differential Mode interferences

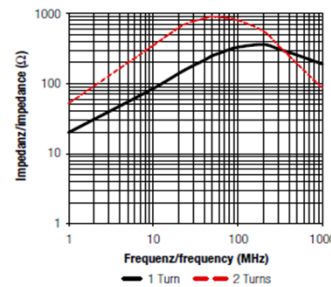
Clamp on Ferrites



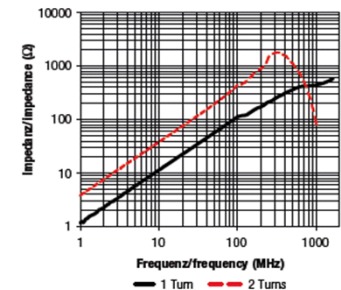
Impedance vs. Frequency



742 711 12 / 742 711 12S



742 716 33 / 742 716 33S



FIX – LFS

TEC / RING / FIX

GAP

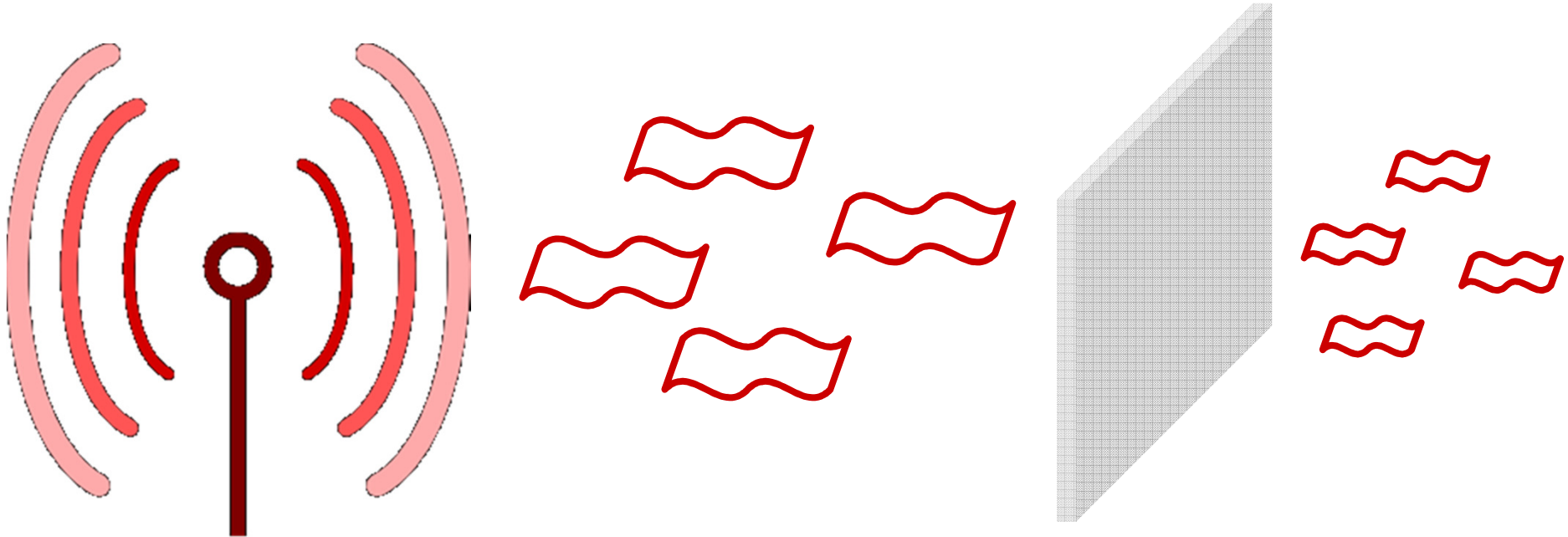
• STAR-GAP

- Developed to increase impedance and to lower the suppression effect in the high frequency range with DC bias.
- Best performance with two turns.
- Very helpful in lowering EMI problems in frequency range of 100MHz up to 2.5GHz.

SHIELDS

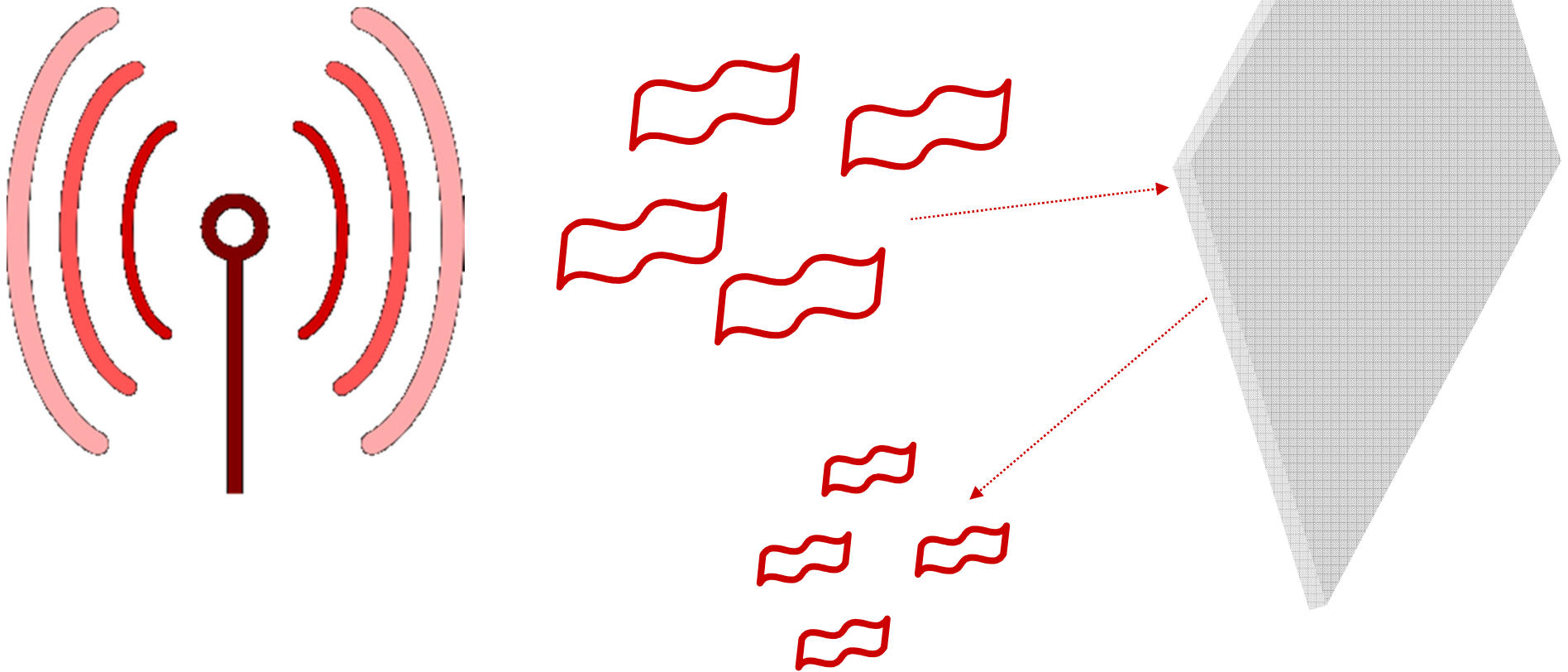
Radiated noise modes

Transmission of radiated noise



Radiated noise modes

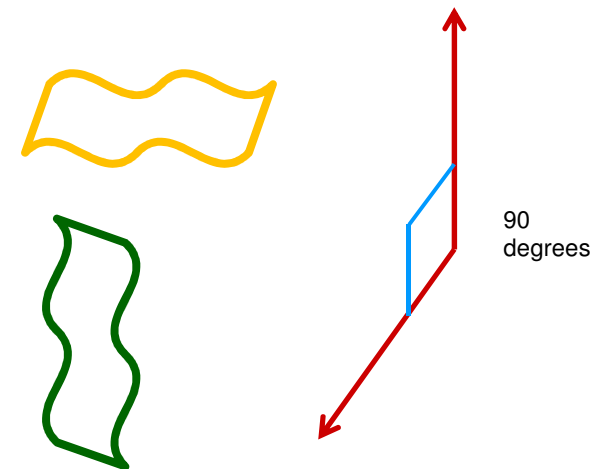
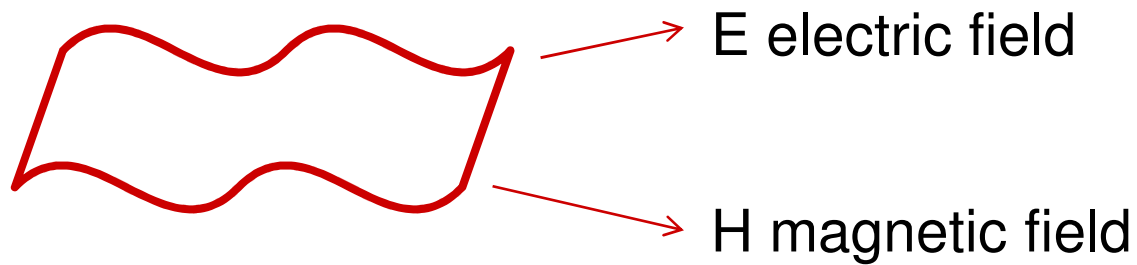
Reflection of radiated noise



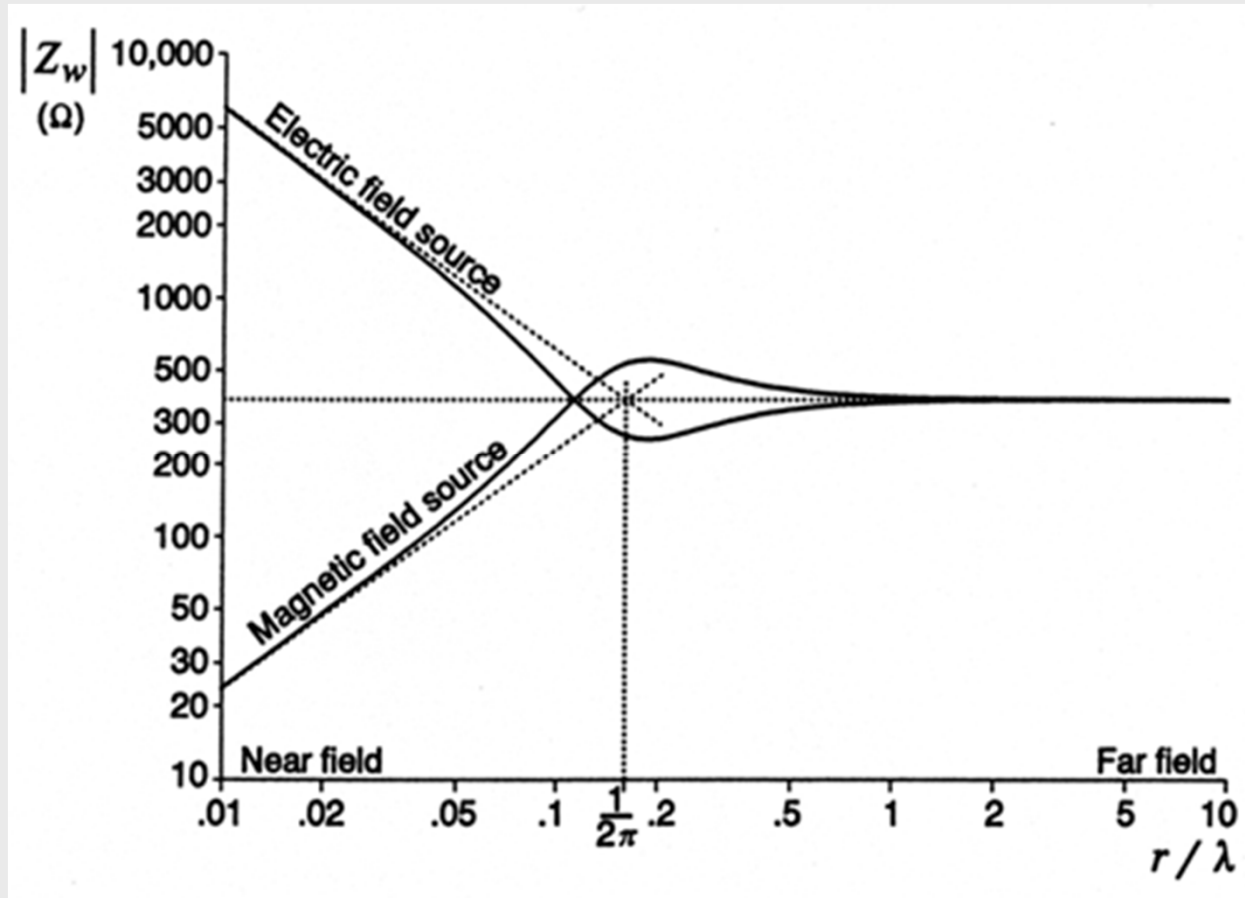
Electromagnetic wave



Components from a magnetic wave



Near Field and Far Field



Enclosures

- Products can be built with two materials regarding RF:

1. Metallic – Aluminium, steel, brass, copper.

Natural shielding materials



2. Non metallic – Plastic, Nylon, Poliestyrene, PVC.

Materials transparent to RF



RF shielding types

Principal two type: **Metallic** (Cu, Al, Steel) and **Ferrite** (NiZn, MnZn or similar compounds)



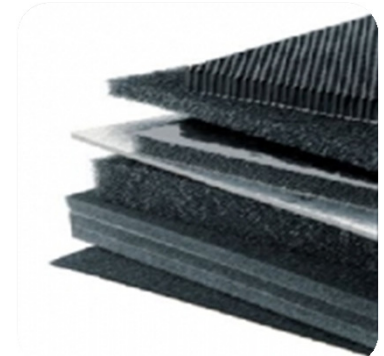
metallic shielding
WE-SHC



Absorber sheet
WE-FAS



Conductive gasket
WE-LTS



Conductive foam
WE-LS



Conductive gasket
WE-LT

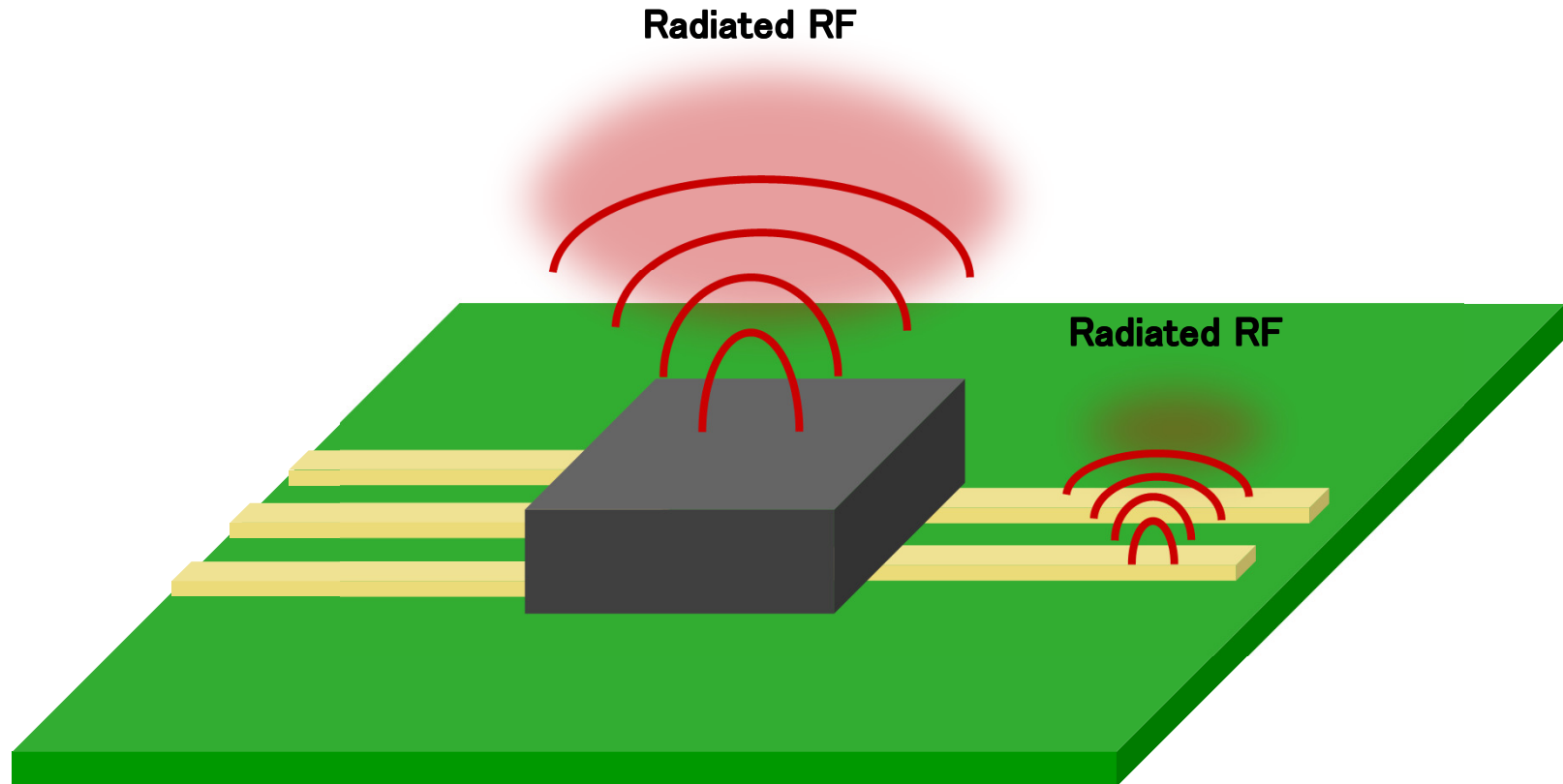


Conductive tapes
WE-CF/TS



Brooke Shields

Basic problem to be solved



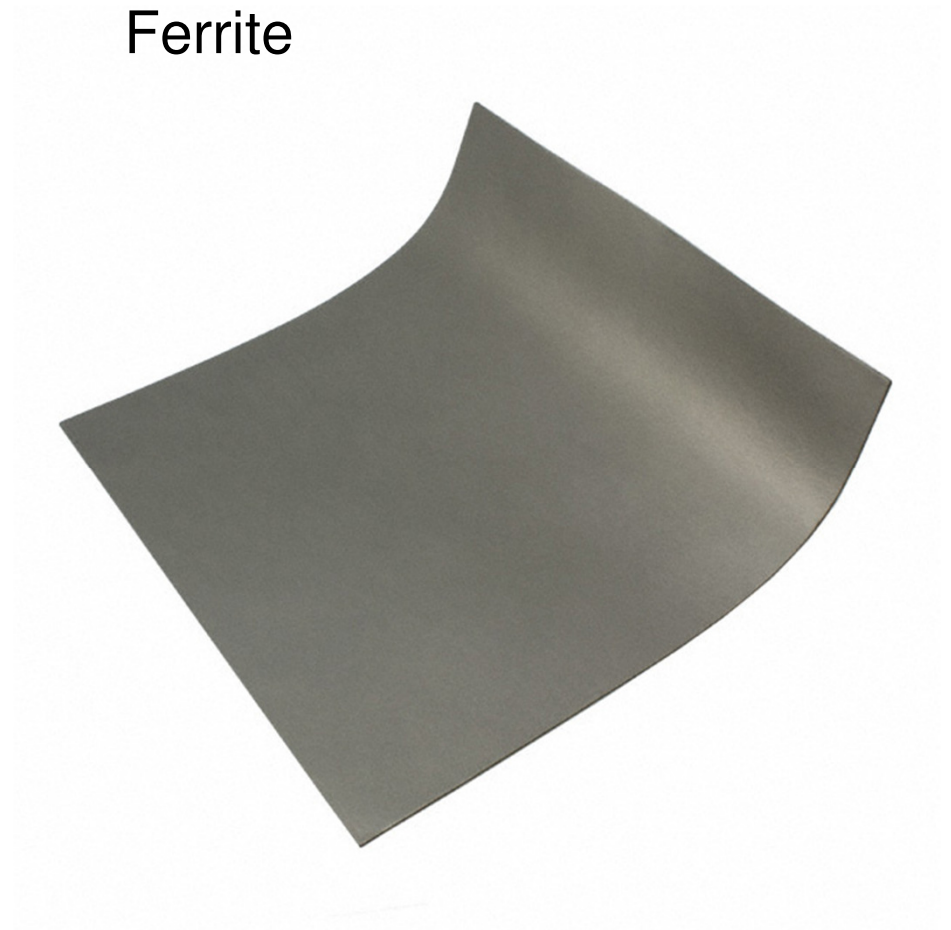
Applications (Shielding)



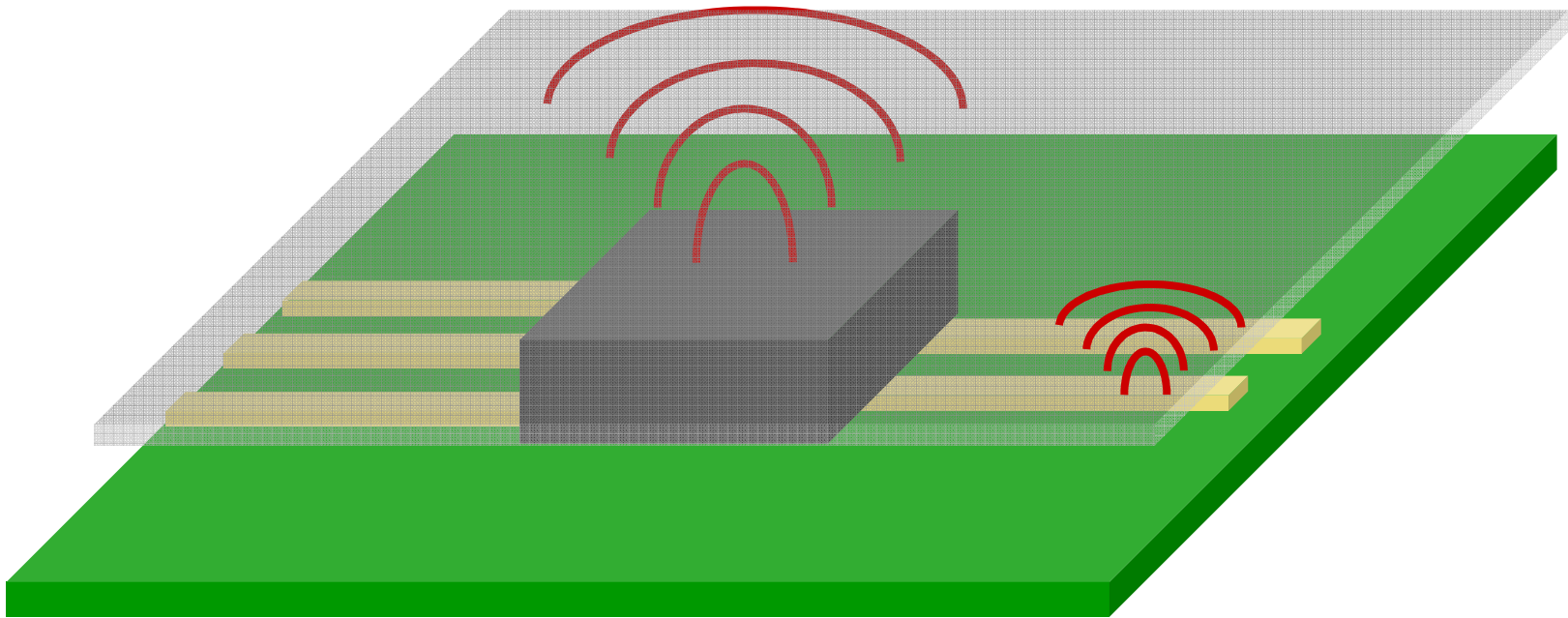
metallic



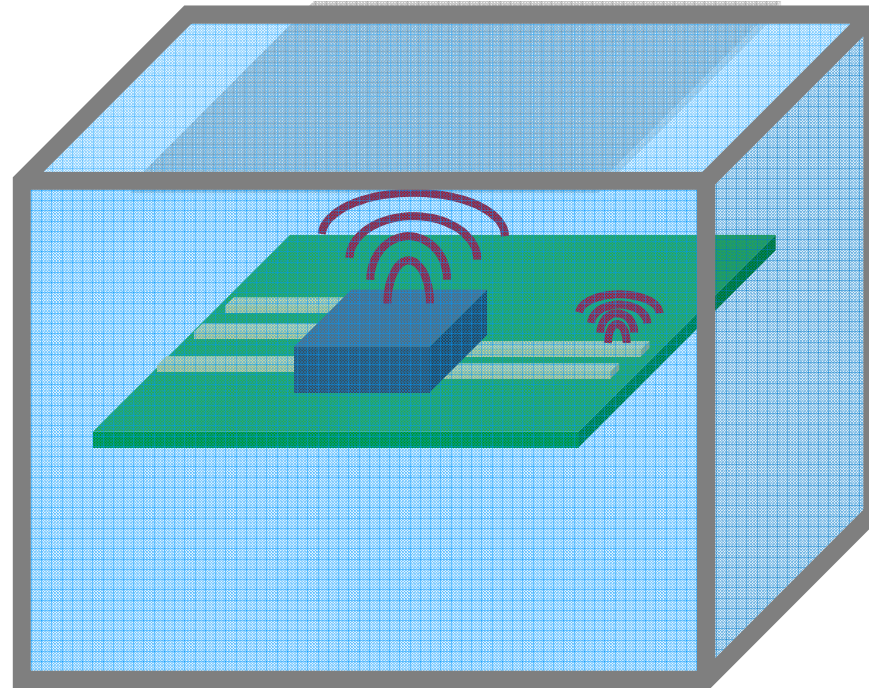
Ferrite



Application (Shielding components)



Application (shielding enclosures)



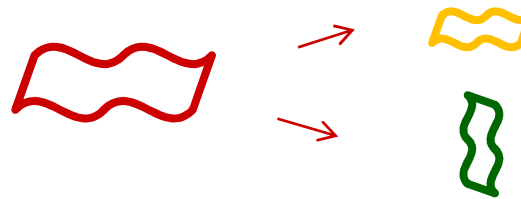
Revision



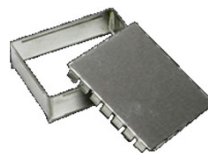
Two paths: transmission and reflection



Two components E field and H Field



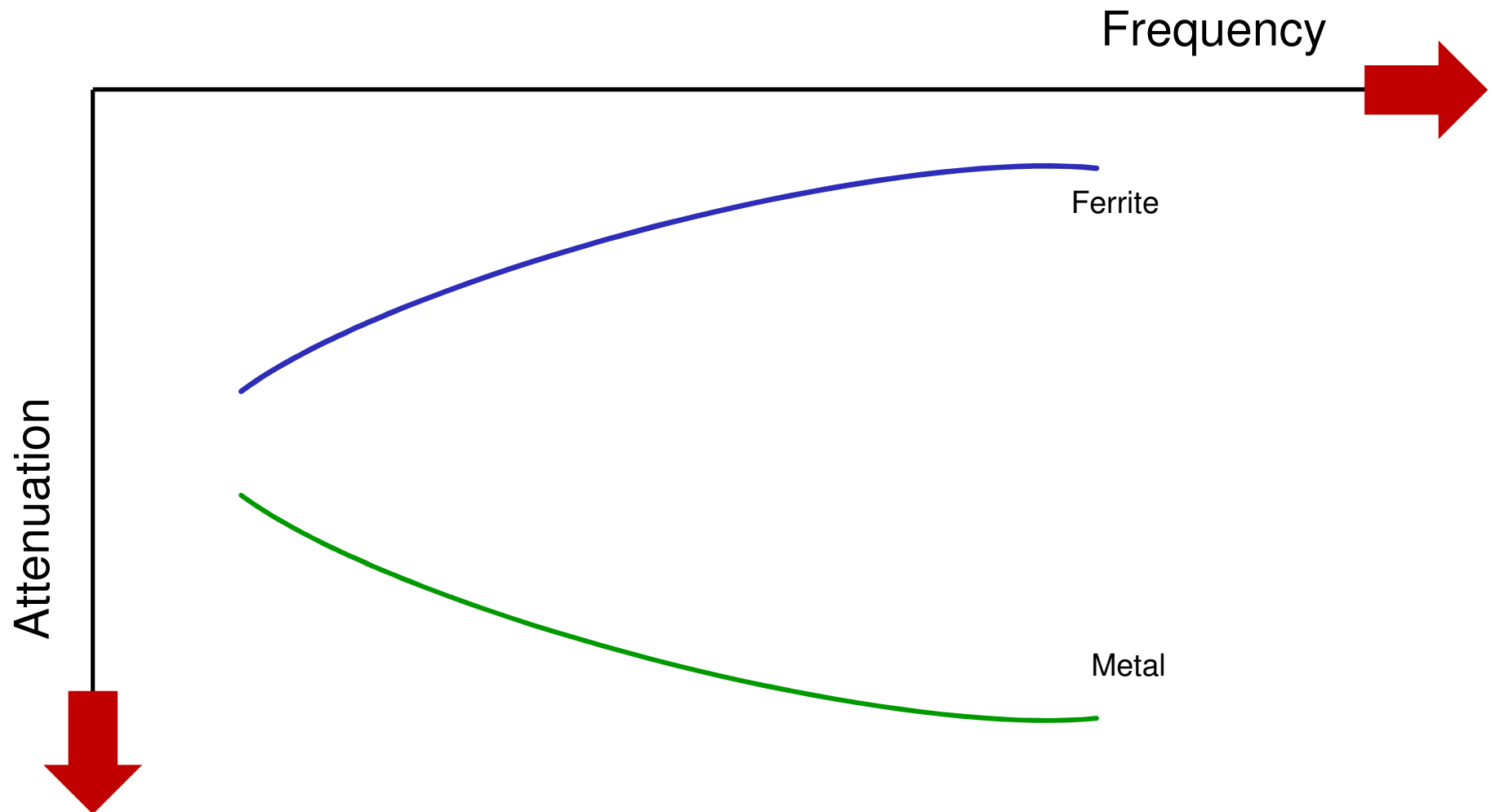
Two types of material



Material Behavior with Radiated Signals



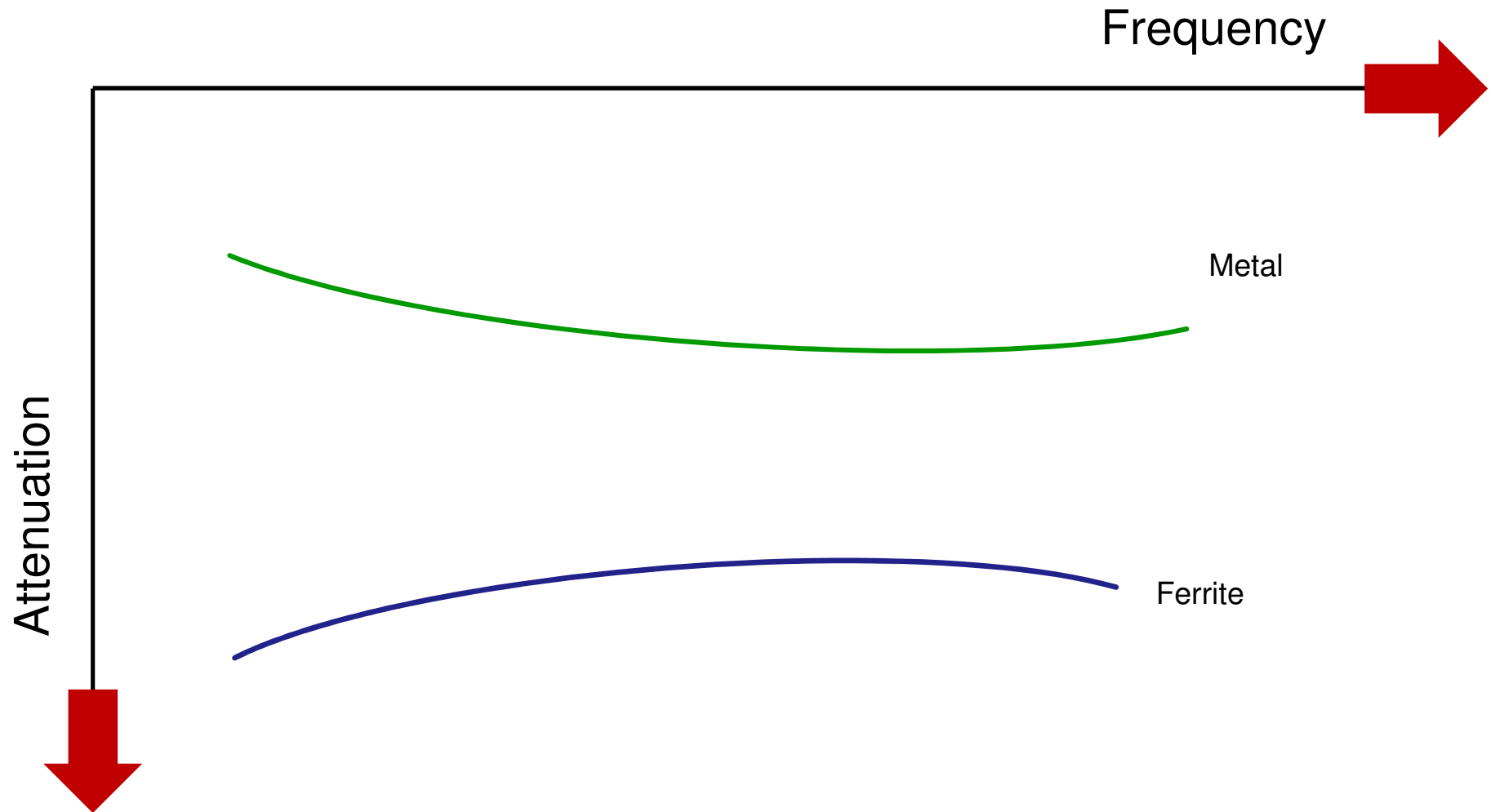
Metal vs Ferrite when signals are transmitted



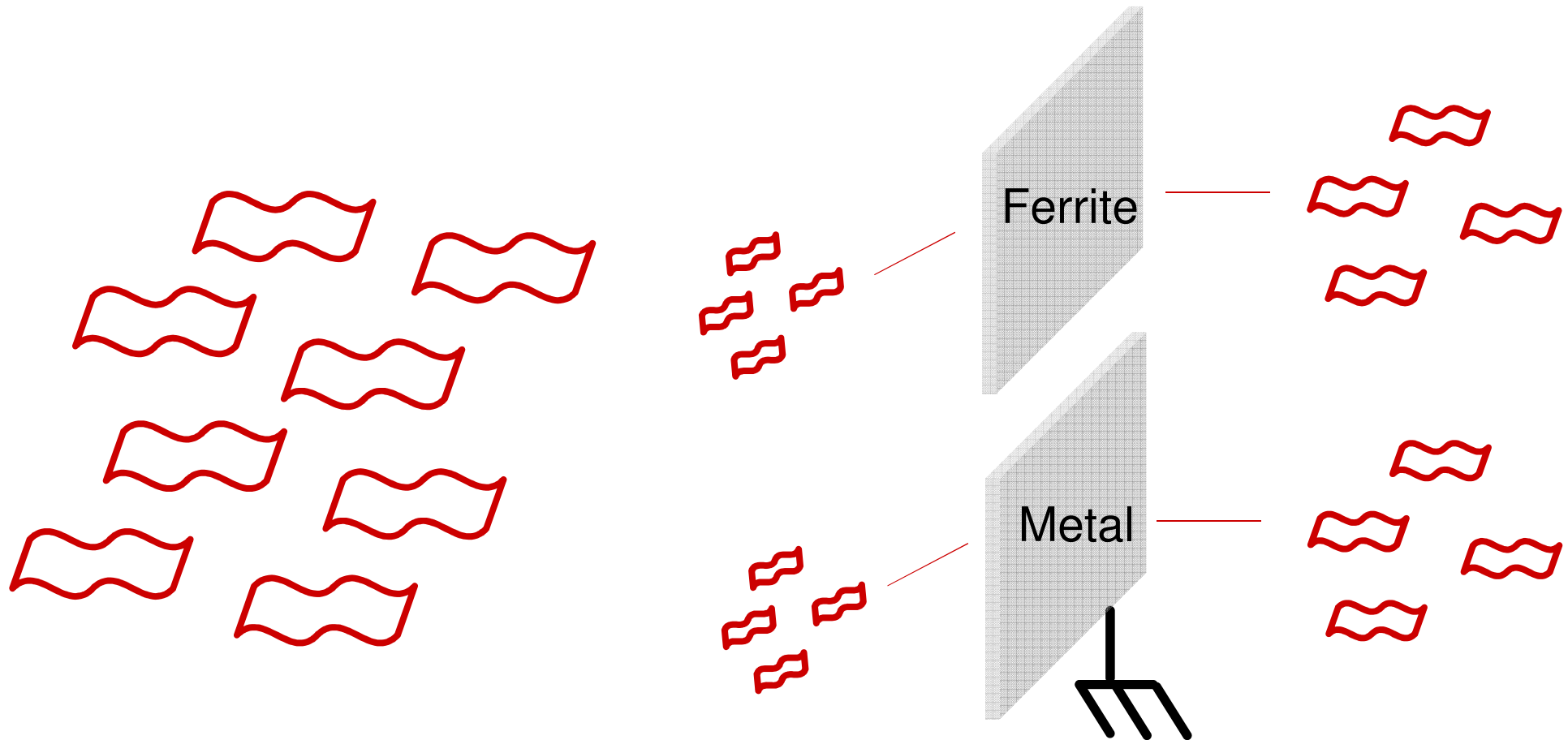
Material Behavior with Radiated Signals



Metal vs Ferrite when signals are reflected



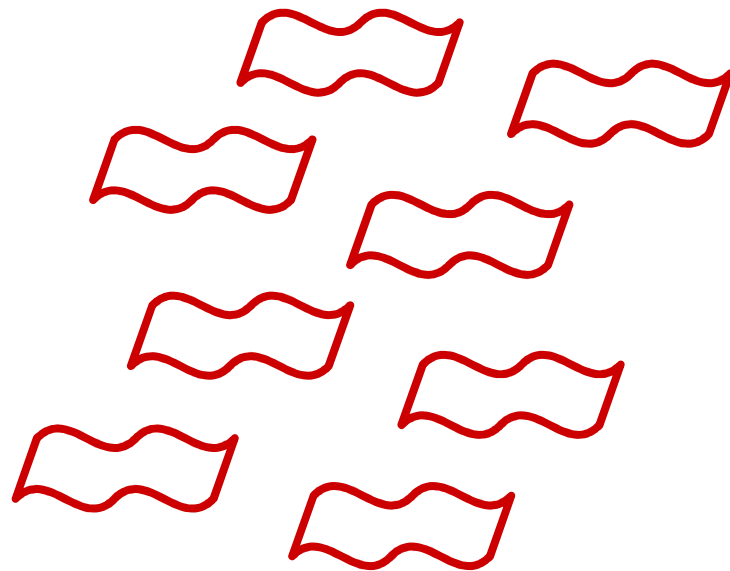
Advantages Metal vs Ferrite



Metal has to be grounded! But a ferrite does not.

Metal is cheaper than ferrite and easy to find.

Shield Effect on Components of the EM Wave

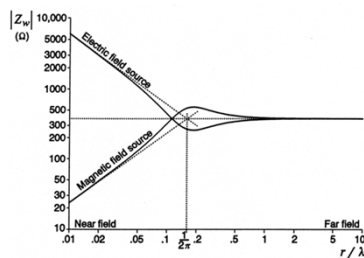


Ferrite

Metal

Ferrite works better at short distances and it affects more significantly the magnetic field

Metal works at any distance and it affects more significantly the electric field



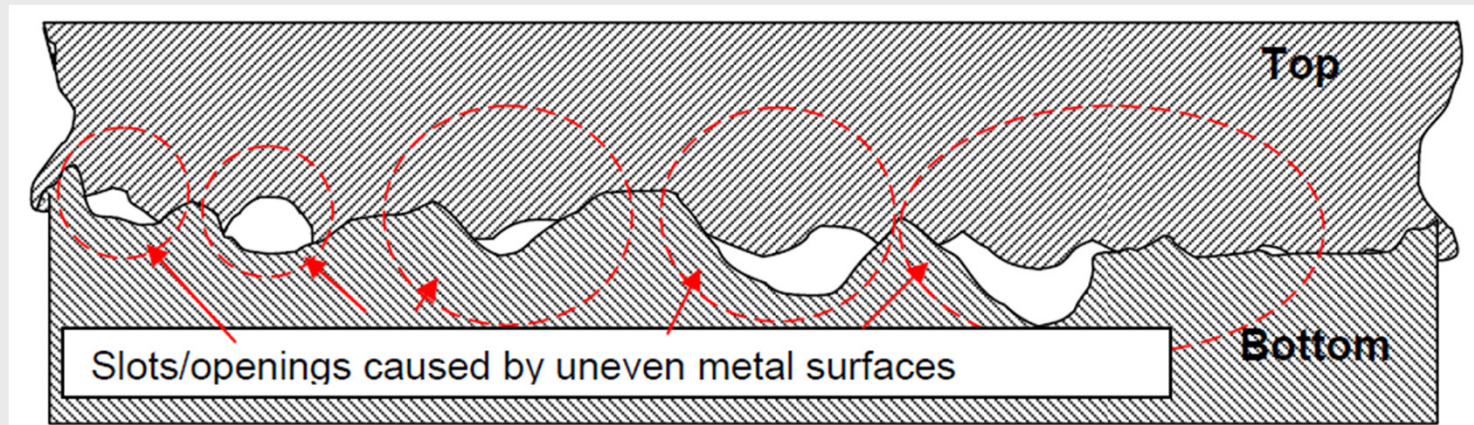
Applications (scenario)



Metallic enclosure



Problems with metallic enclosures



- Gaps allow RF noises pass through.
- Slots can work as antennas

Antena WLAN 2,4GHz

$$\lambda = \frac{3 \times 10^8 \cdot m \cdot s}{2,4 \times 10^9 \cdot s}$$

$$\lambda = 0,125m$$



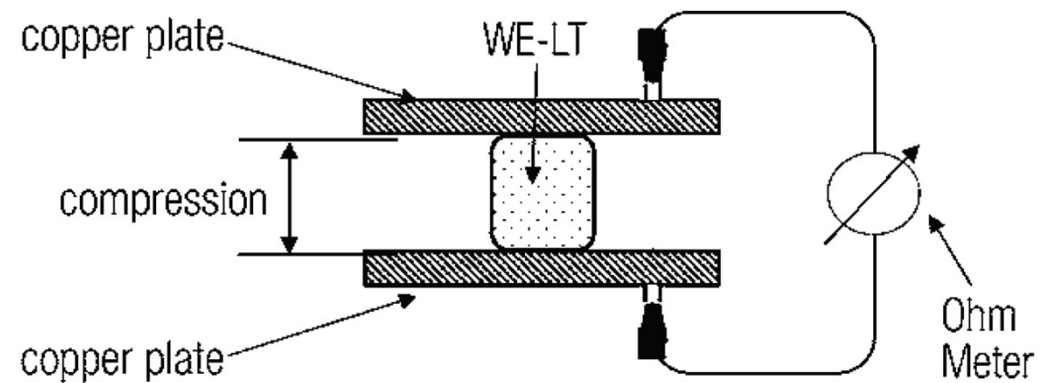
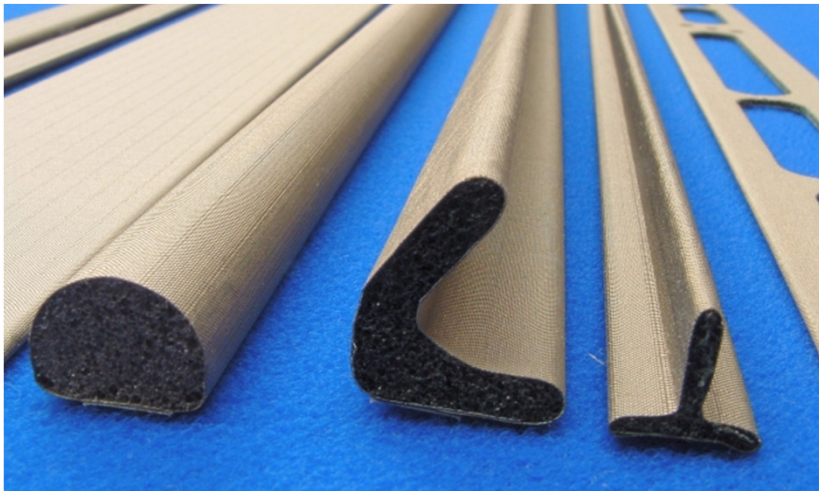
$$\lambda / 4 = 3,125cm$$



Metallic enclosures correctly grounded



Gaskets and conductive tapes

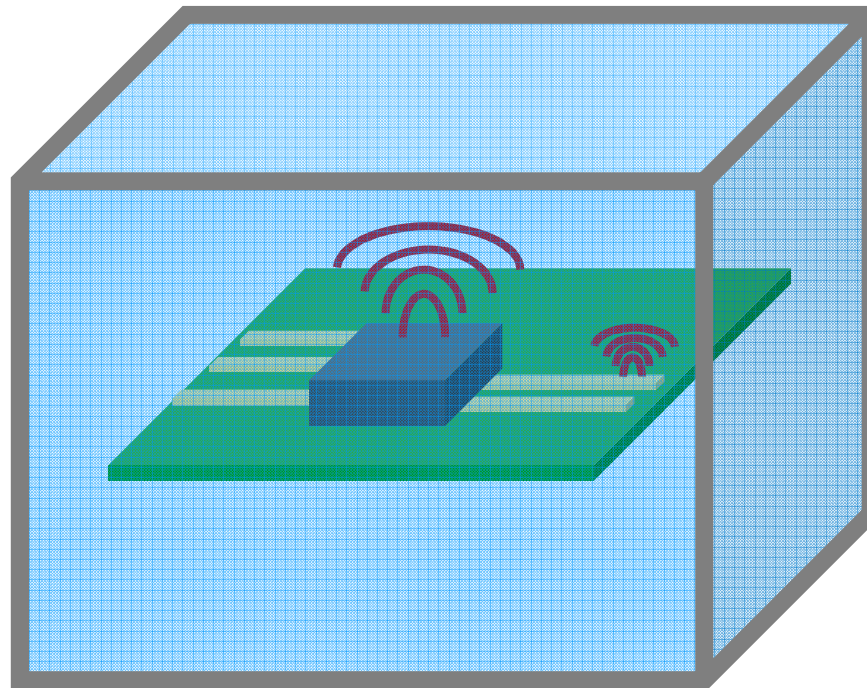


Best gasket performance is when compressed by 50% to lowest impedance.

Applications (Gaskets)



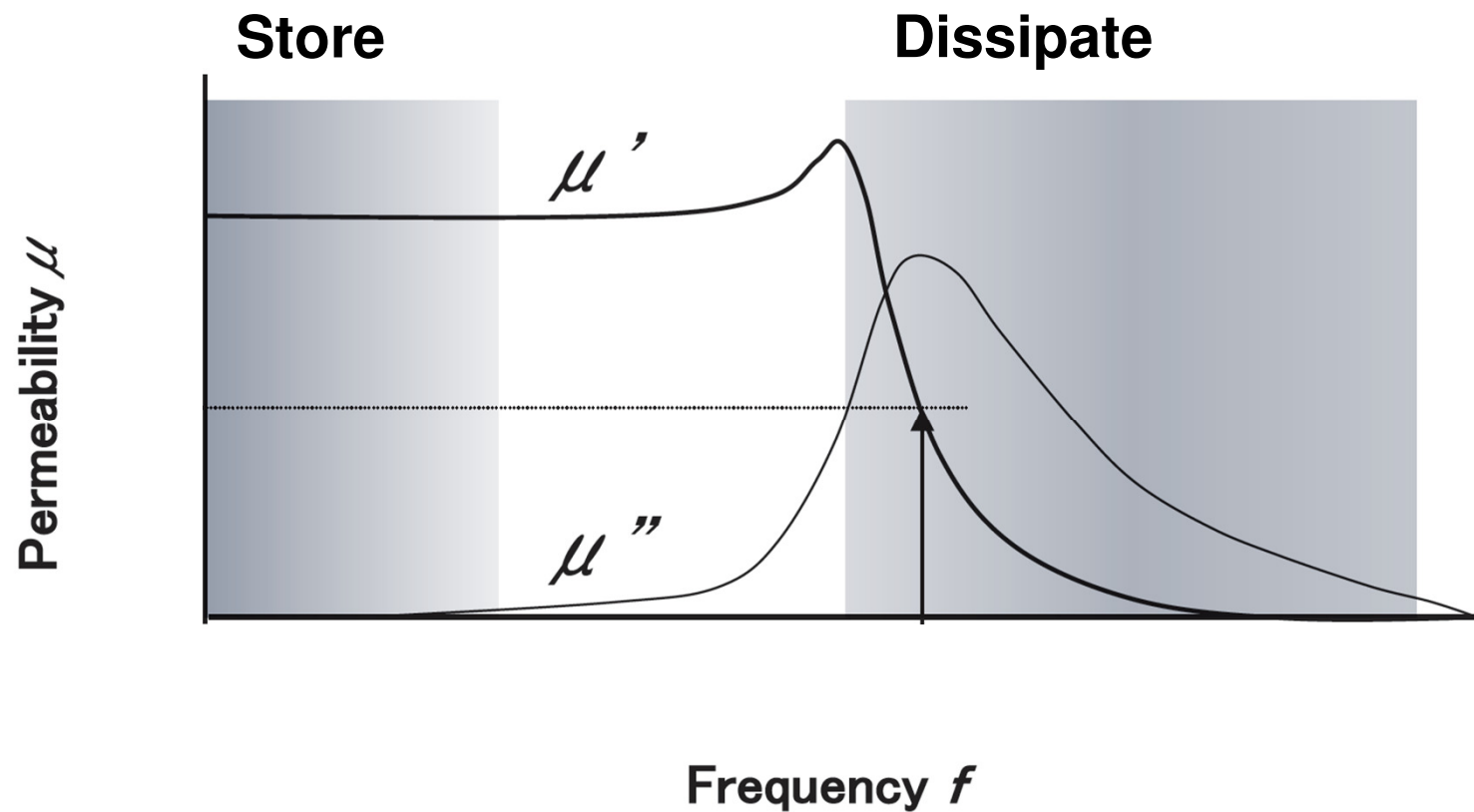
Gaskets fill the gap between metallic walls from enclosure.



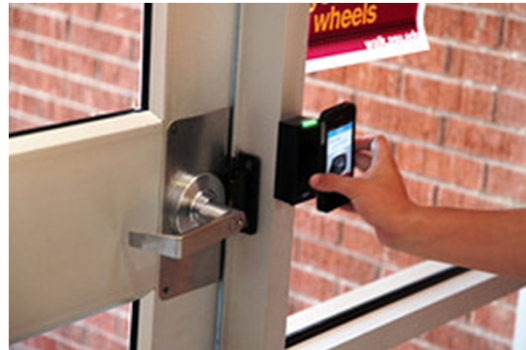
Ferrites characteristics:



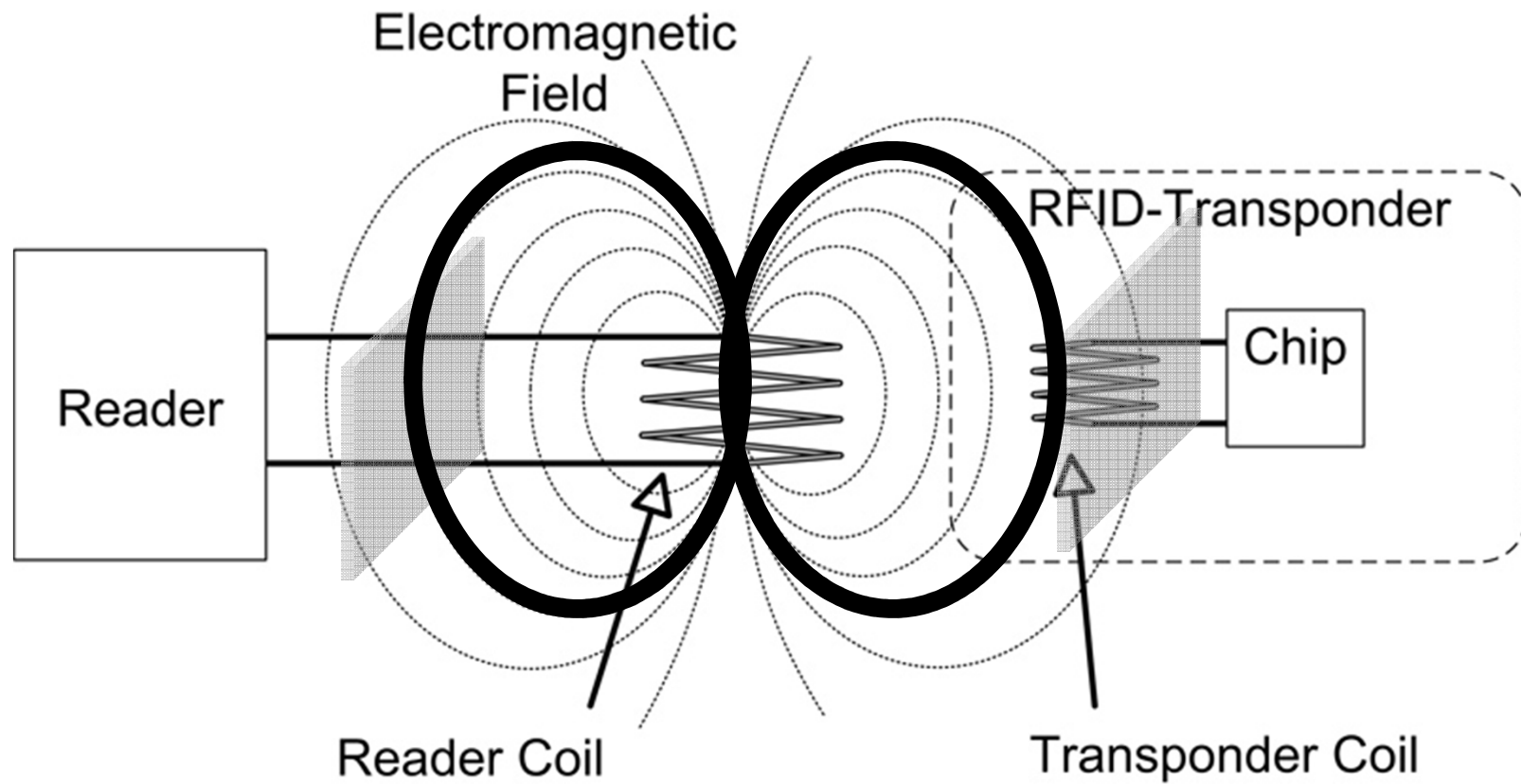
Ferrites can store and dissipate



Application (Enhance RFID 13.56MHz)



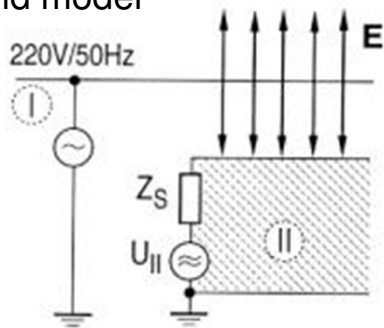
Applications (Enhance RFID 13.56MHz)



What is best? Metal or Ferrite?

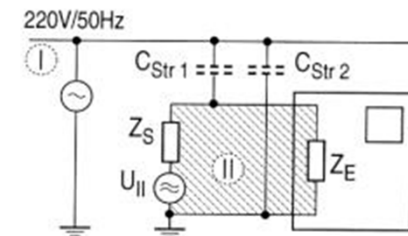
Capacitive coupling

Field model



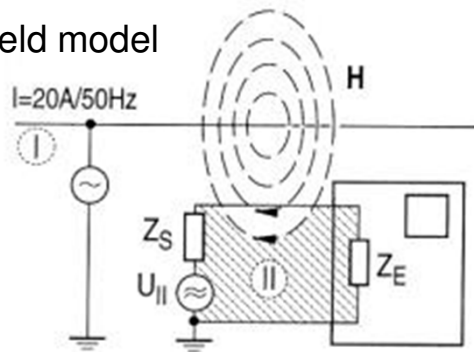
None

Network model



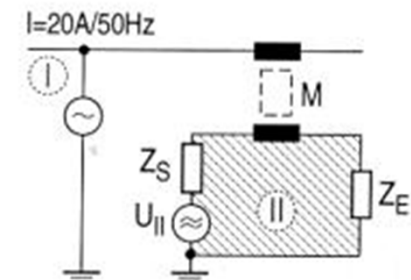
Inductive coupling

Field model



Ferrite

Network model



Q&A

<http://www.we-online.com/>

http://www.we-online.com/web/en/passive_bauelemente_-_standard/toolbox_pbs/Toolbox.php

Wurth Electronics Midcom Inc. Headquarters

Phone: (605) 886-4385

Fax: (605) 886-4486

E-Mail: midcom@we-online.com

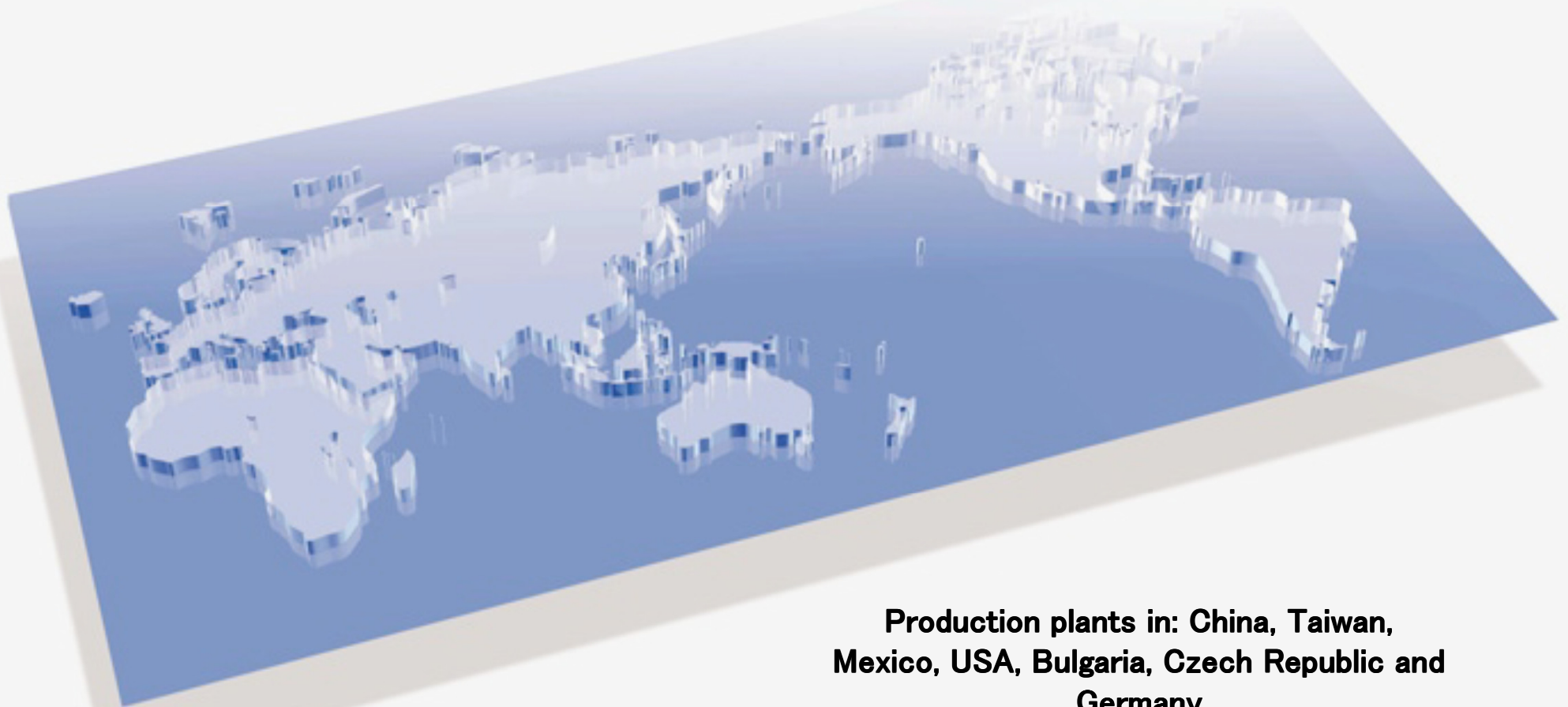
121 Airport Drive

Watertown, SD 57201

United States

Headquarters in Germany.

Offices: USA, Mexico, Brazil, Canada, UK, Sweden, France, Ireland, Italy, Austria, Spain, Switzerland, Netherlands, Czech Rep., Hungary, Singapore, China and Taiwan



Production plants in: China, Taiwan, Mexico, USA, Bulgaria, Czech Republic and Germany

APPENDIX

EMC Overview

Conductive coupling

- Coupling path between source and victim is formed by direct contact.
Example a transmission line, wire, cable, PCB trace or metal enclosure.

Capacitive coupling

- Electric field coupling.
- Occurs when a varying electrical field exists between two adjacent conductors.

Inductive coupling

- Magnetic field coupling
- Occurs when a varying magnetic field exists between two parallel conductors.
Coupling between conductors causes parasitic induced voltages.

Radiative

- Source is the “transmitter” and victim is the “receiver”.

The Magnetic Field



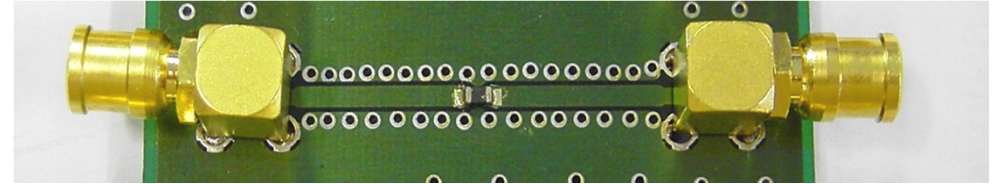
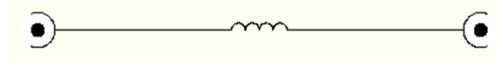
The H field corresponds to what is called the magnetic field strength. It is measured in amps / meter (A/m).

In free space or in air the B field represents magnetic flux density which is given in units of Tesla by $B = \mu_0 \cdot H$ where μ_0 is the absolute magnetic permeability of free space $\mu_0 = 4\pi \times 10^{-7} \frac{N}{A^2}$

More magnetic flux can be produced by the same H value in certain (magnetic) materials, notably iron, and this is accounted by introducing another factor, the relative permeability μ_r , giving $B = \mu_0 \mu_r H$ for magnetic materials.

Filter topologies – L-Filter

- L-Filter

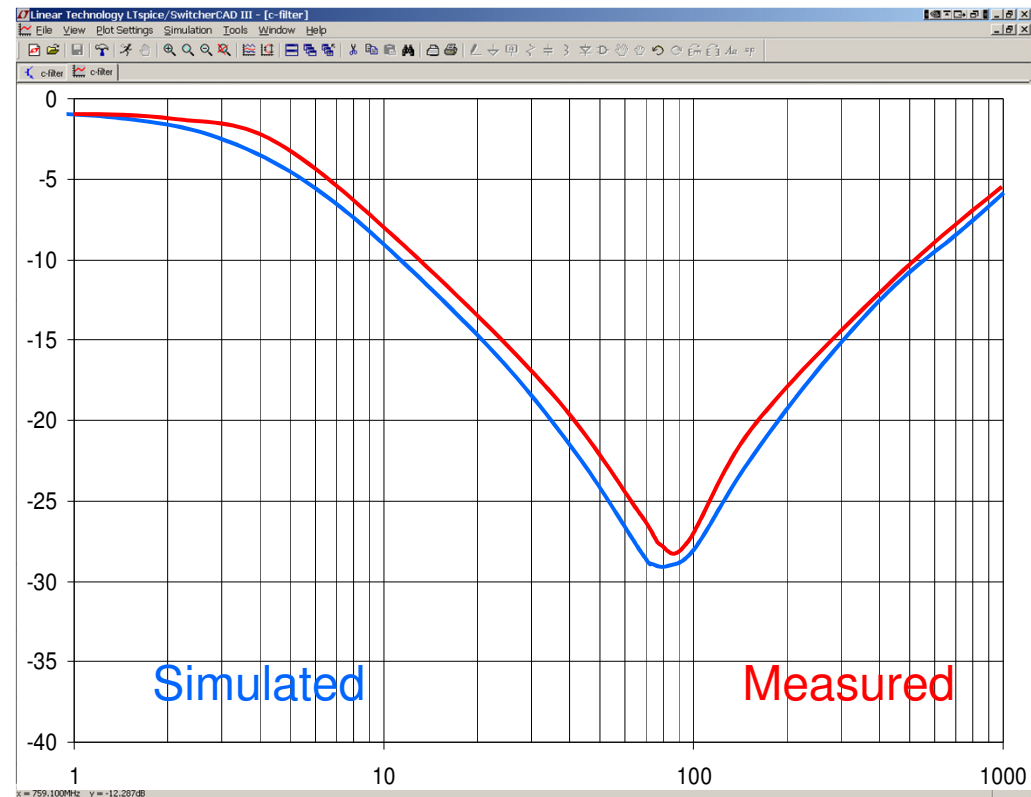


- instead of inductor use a SMD-Ferrite

- WE-CBF 742 792 093

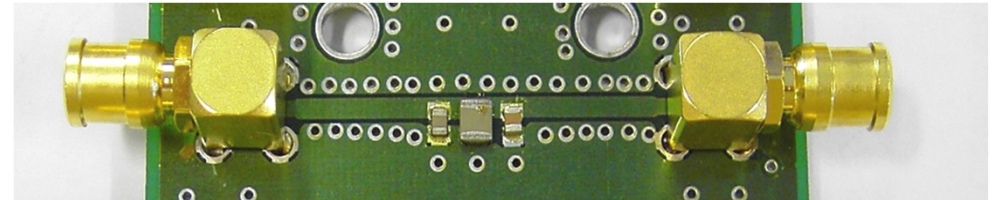
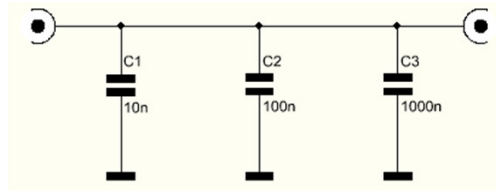
$$Z_{\max} = 3000 \, \Omega \text{ @ } 80 \text{ MHz}$$

$$A_F = -29 \text{ dB @ } 80 \text{ MHz}$$

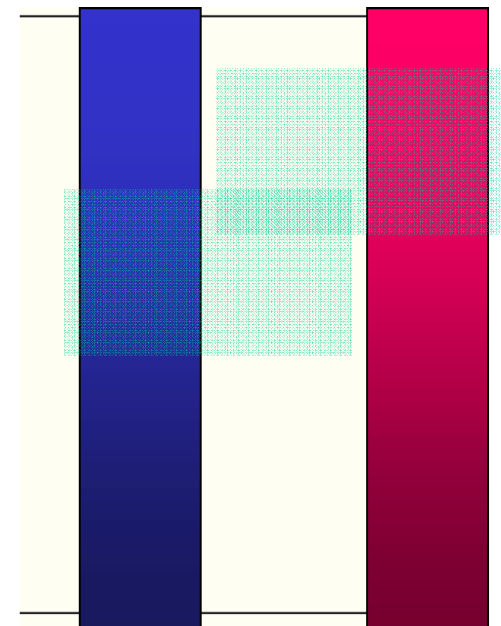
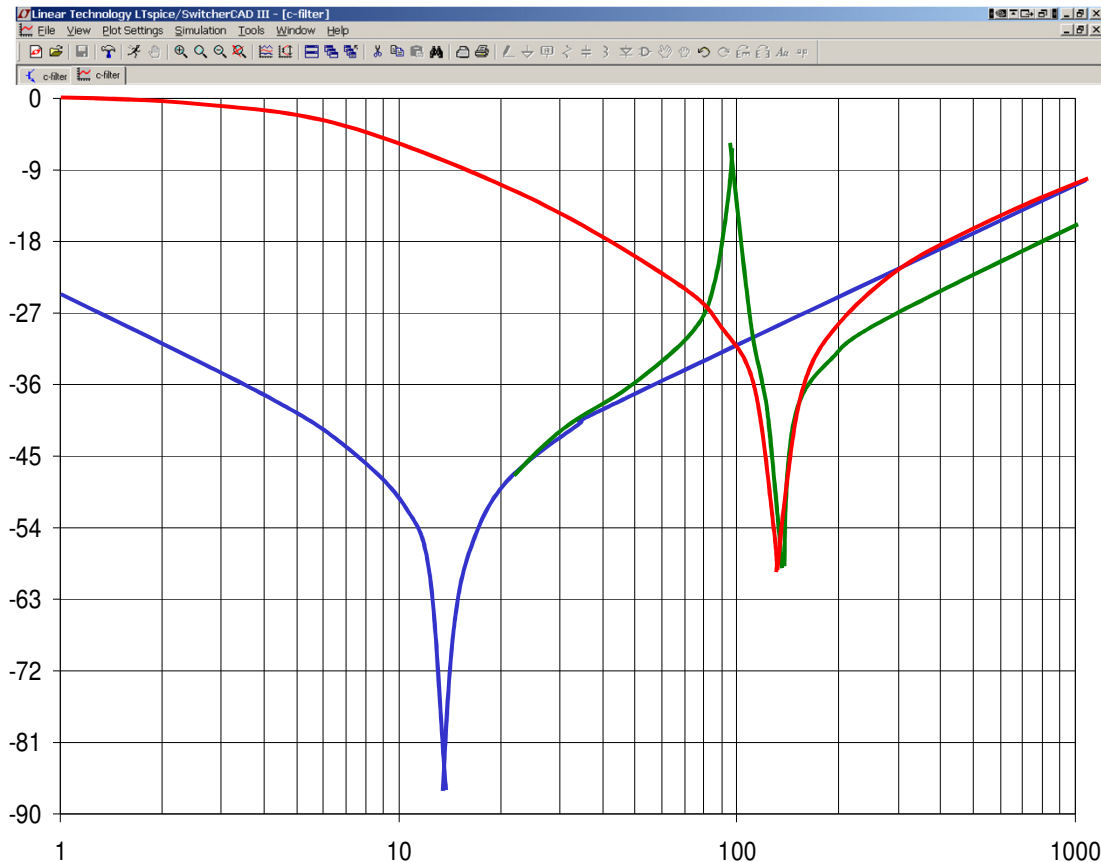


Filter topologies – Parallel-C-Filter

- Parallel-C-Filter

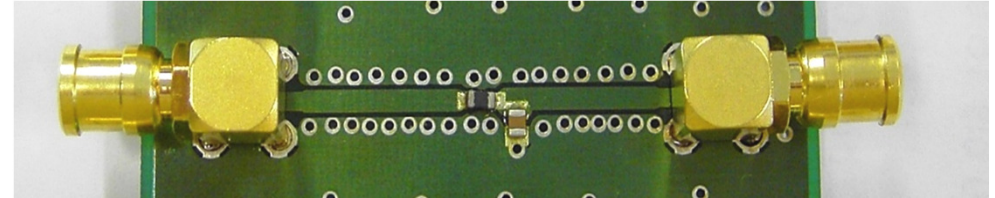
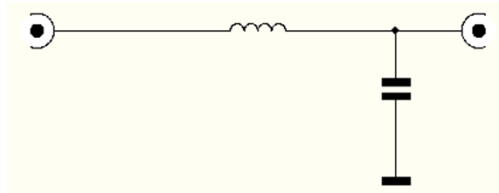


- resonance points



Filter topologies – LC-Filter

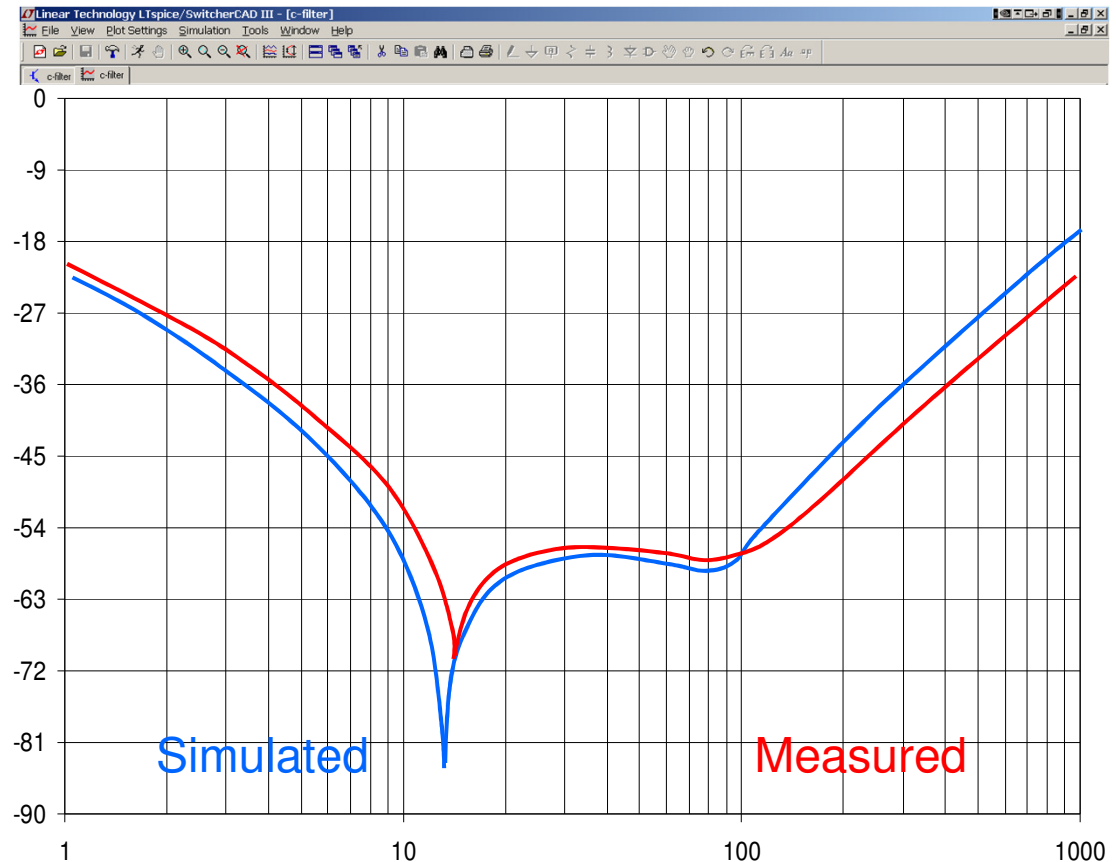
- LC-Filter



- Comparison simulated vs. measured

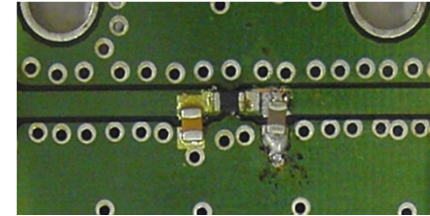
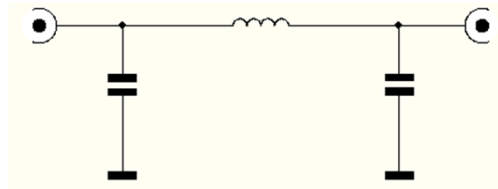
WE-CBF 742 792 093

$C=100\text{nF}$



Filter topologies – PI-Filter

- π -Filter

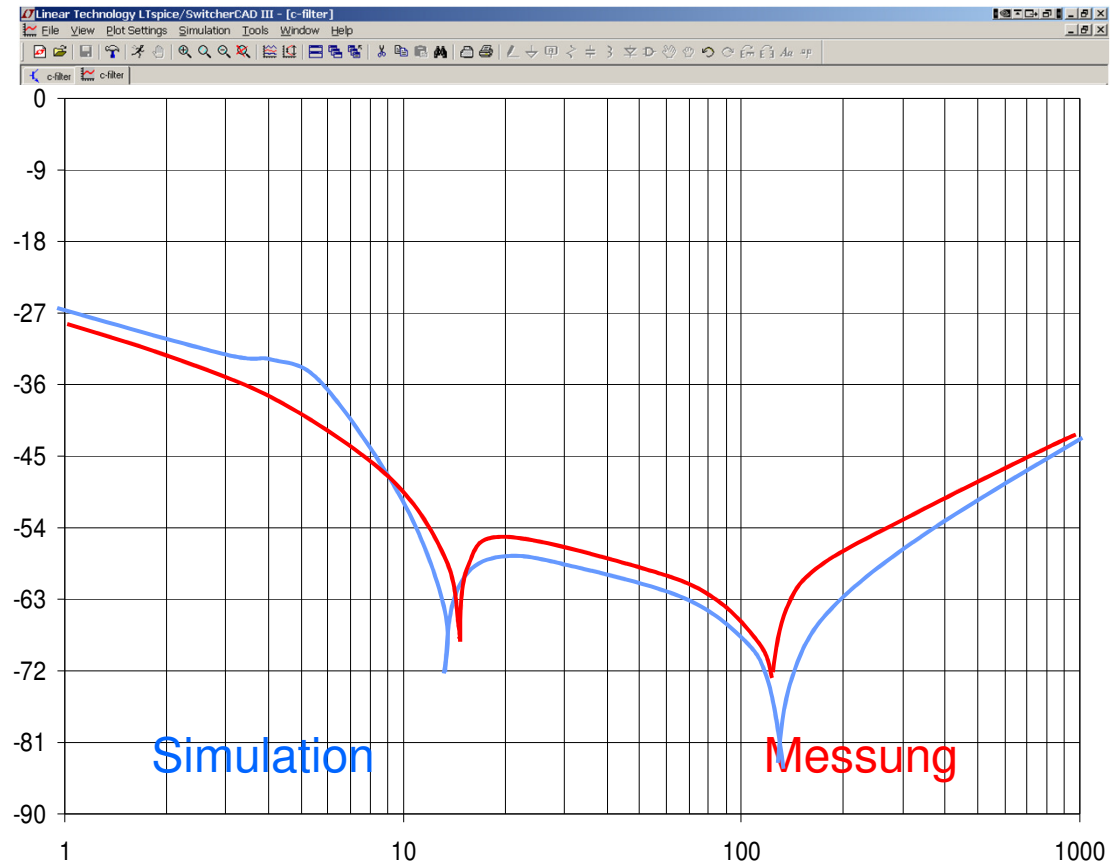


- Comparison Measurement - Simulation

WE-CBF 742 792 093

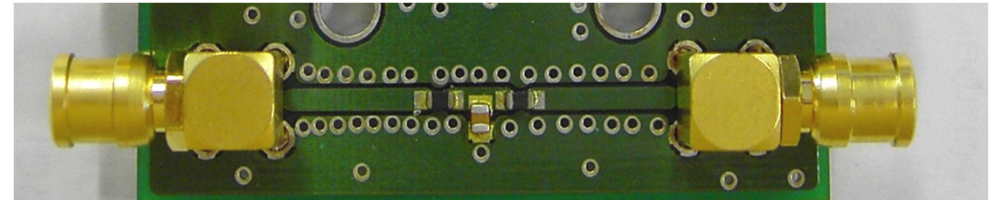
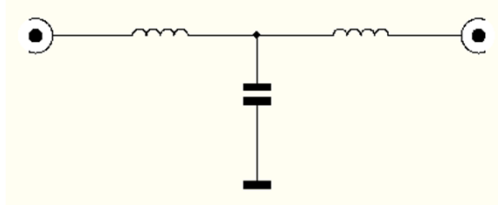
$C_1 = 1\text{nF}$

$C_2 = 100\text{nF}$



Filter topologies – T-Filter

- T-Filter

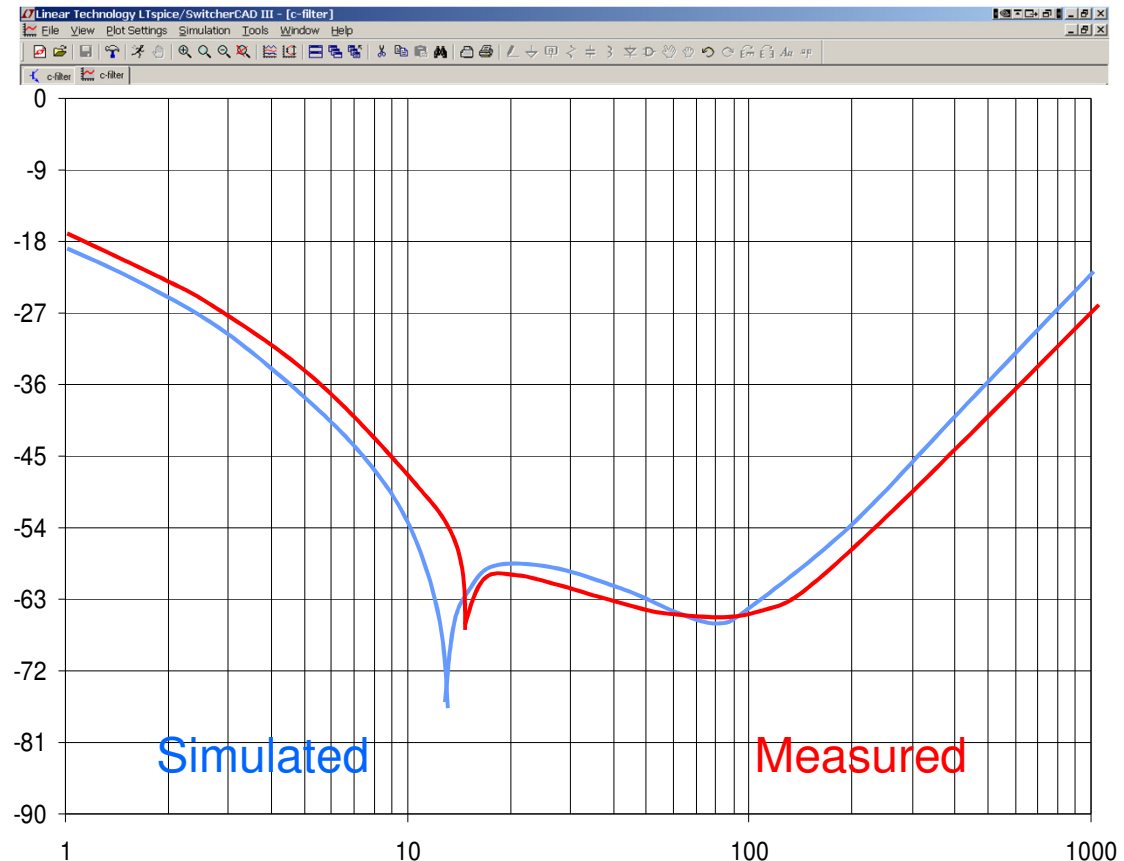


- Comparison simulated vs. measured

$C=100\text{nF}$

$L_1=742\ 792\ 040$

$L_2=742\ 792\ 093$



Soft ferrites (Typical permeability values)

- Iron power / Superflux : 50 ~ 150
- Nickel Zinc : 40 ~ 1500
- Manganese Zinc : 300 ~ 20000

The Magnetic Field



The H field corresponds to what is called the magnetic field strength. It is measured in amps / meter (A/m).

In free space or in air the B field represents magnetic flux density which is given in units of Tesla by $B = \mu_0 \cdot H$ where μ_0 is the absolute magnetic permeability of free space $\mu_0 = 4\pi \times 10^{-7} \frac{N}{A^2}$

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Thank you